Designed primarily for linear large signal output stages in the 2.0 to 100 MHz frequency range.

N–Channel enhancement mode MOSFET

- Specified 50 volts, 30 MHz characteristics
  - Output power = 600 watts
  - Power gain = 17 dB (typ.)
  - Efficiency = 45% (typ.)

MAXIMUM RATINGS

<table>
<thead>
<tr>
<th>Rating</th>
<th>Symbol</th>
<th>Value</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Drain–Source Voltage</td>
<td>$V_{DSS}$</td>
<td>125</td>
<td>Vdc</td>
</tr>
<tr>
<td>Drain–Gate Voltage</td>
<td>$V_{DGO}$</td>
<td>125</td>
<td>Vdc</td>
</tr>
<tr>
<td>Gate–Source Voltage</td>
<td>$V_{GS}$</td>
<td>±40</td>
<td>Vdc</td>
</tr>
<tr>
<td>Drain Current — Continuous</td>
<td>$I_D$</td>
<td>60</td>
<td>Adc</td>
</tr>
<tr>
<td>Total Device Dissipation @ $T_C = 25^\circ C$</td>
<td>$P_D$</td>
<td>1350</td>
<td>Watts</td>
</tr>
<tr>
<td>Derate above 25°C</td>
<td></td>
<td>7.7</td>
<td>W/°C</td>
</tr>
<tr>
<td>Storage Temperature Range</td>
<td>$T_{slg}$</td>
<td>−65 to +150</td>
<td>°C</td>
</tr>
<tr>
<td>Operating Junction Temperature</td>
<td>$T_J$</td>
<td>200</td>
<td>°C</td>
</tr>
</tbody>
</table>

THERMAL CHARACTERISTICS

<table>
<thead>
<tr>
<th>Characteristic</th>
<th>Symbol</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Thermal Resistance, Junction to Case</td>
<td>$R_{BJC}$</td>
<td>0.13</td>
<td>°C/W</td>
</tr>
</tbody>
</table>

Handling and Packaging — MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.
## ELECTRICAL CHARACTERISTICS (T_C = 25°C unless otherwise noted)

<table>
<thead>
<tr>
<th>Characteristic</th>
<th>Symbol</th>
<th>Min</th>
<th>Typ</th>
<th>Max</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>OFF CHARACTERISTICS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Drain–Source Breakdown Voltage (V_GS = 0, I_D = 100 mA)</td>
<td>V(BR)DSS</td>
<td>125</td>
<td></td>
<td></td>
<td>Vdc</td>
</tr>
<tr>
<td>Zero Gate Voltage Drain Current (V_DS = 50 V, V_GS = 0)</td>
<td>I_DSS</td>
<td></td>
<td></td>
<td></td>
<td>mAdc</td>
</tr>
<tr>
<td>Gate–Body Leakage Current (V_GS = 20 V, V_DS = 0)</td>
<td>I_GSS</td>
<td></td>
<td></td>
<td></td>
<td>μAdc</td>
</tr>
<tr>
<td><strong>ON CHARACTERISTICS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Gate Threshold Voltage (V_DS = 10 V, I_D = 100 mA)</td>
<td>V_GS(th)</td>
<td>1.0</td>
<td>3.0</td>
<td>5.0</td>
<td>Vdc</td>
</tr>
<tr>
<td>Drain–Source On–Voltage (V_GS = 10 V, I_D = 40 A)</td>
<td>V_DS(on)</td>
<td>1.0</td>
<td>3.0</td>
<td>5.0</td>
<td>Vdc</td>
</tr>
<tr>
<td>Forward Transconductance (V_DS = 10 V, I_D = 20 A)</td>
<td>g_fs</td>
<td>16</td>
<td>20</td>
<td></td>
<td>mhos</td>
</tr>
<tr>
<td><strong>DYNAMIC CHARACTERISTICS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Input Capacitance (V_DS = 50 V, V_GS = 0, f = 1.0 MHz)</td>
<td>C_GSS</td>
<td></td>
<td>1600</td>
<td></td>
<td>pF</td>
</tr>
<tr>
<td>Output Capacitance (V_DS = 50 V, V_GS = 0, f = 1.0 MHz)</td>
<td>C_GSS</td>
<td></td>
<td>950</td>
<td></td>
<td>pF</td>
</tr>
<tr>
<td>Reverse Transfer Capacitance (V_DS = 50 V, V_GS = 0, f = 1.0 MHz)</td>
<td>C_RSS</td>
<td></td>
<td>175</td>
<td></td>
<td>pF</td>
</tr>
<tr>
<td><strong>FUNCTIONAL TESTS</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Common Source Amplifier Power Gain</td>
<td>G_ps</td>
<td></td>
<td>17</td>
<td></td>
<td>dB</td>
</tr>
<tr>
<td>(V_DD = 50 V, P_out = 600 W, I_DQ = 800 mA, f = 30 MHz)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Drain Efficiency</td>
<td>η</td>
<td></td>
<td>45</td>
<td></td>
<td>%</td>
</tr>
<tr>
<td>(V_DD = 50 V, P_out = 600 W, I_DQ = 800 mA, f = 30 MHz)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Intermodulation Distortion</td>
<td>IMD(d3)</td>
<td></td>
<td>−25</td>
<td></td>
<td>dB</td>
</tr>
<tr>
<td>(V_DD = 50 V, P_out = 600 W (PEP), f1 = 30 MHz, f2 = 30.001 MHz, I_DQ = 800 mA)</td>
<td></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
</tbody>
</table>
MRF154

Broadband RF Power MOSFET
600W, to 80MHz, 50V

C1, C3, C8 — Arco 469
C2 — 330 pF
C4 — 880 pF
C5, C19, C20 — 0.47 μF, RMC Type 2225C
C6, C7, C14, C15, C16 — 0.1 μF
C9, C10, C11 — 470 pF
C12 — 1000 pF
C13 — Two Unencapsulated 1000 pF Mica, in Series
C17, C18 — 0.039 μF
C21 — 10 μF/100 V Electrolytic
L1 — 2 Turns #16 AWG, 1/2” OD, 3/8” Long
L2, L3 — Ferrite Beads, Fair–Rite Products Corp. #2673000801

R1, R2 — 10 Ohms/20 W Carbon
T1 — RF Transformer, 1:25 impedance Ratio. See M/A-COM Application Note AN749, Figure 4 for details.
Ferrite Material: 2 Each, Fair–Rite Products Corp. #2667540301

All capacitors ATC type 100/200 chips or equivalent unless otherwise noted.

**Figure 1. 30 MHz Test Circuit**

**Figure 2. Power Gain versus Frequency**

**Figure 3. Output Power versus Input Power**
Broadband RF Power MOSFET
600W, to 80MHz, 50V

Figure 4. DC Safe Operating Area

Figure 5. Capacitance versus Drain Voltage

Figure 6. Gate Voltage versus Drain Current

Figure 7. Common Source Unity Gain Frequency versus Drain Current
Figure 8. Series Equivalent Impedance
Figure 9. 20–80 MHz 1.0 kW Broadband Amplifier
RF POWER MOSFET CONSIDERATIONS

The physical structure of a MOSFET results in capacitors between the terminals. The metal oxide gate structure determines the capacitors from gate–to–drain (Cgd), and gate–to–source (Cgs). The PN junction formed during the fabrication of the RF MOSFET results in a junction capacitance from drain–to–source (Cds).

These capacitances are characterized as input (Ciss), output (Coss) and reverse transfer (Crss) capacitances on data sheets. The relationships between the inter–terminal capacitances and those given on data sheets are shown below. The Ciss can be specified in two ways:

1. Drain shorted to source and positive voltage at the gate.
2. Positive voltage of the drain in respect to source and zero volts at the gate. In the latter case the numbers are lower. However, neither method represents the actual operating conditions in RF applications.

\[
\begin{align*}
\text{C}_{\text{iss}} &= \text{C}_{\text{gd}} + \text{C}_{\text{gs}} \\
\text{C}_{\text{oss}} &= \text{C}_{\text{gd}} + \text{C}_{\text{gs}} \\
\text{C}_{\text{rss}} &= \text{C}_{\text{gd}}
\end{align*}
\]

LINEARITY AND GAIN CHARACTERISTICS

In addition to the typical IMD and power gain data presented, Figure 5 may give the designer additional information on the capabilities of this device. The graph represents the small signal unity current gain frequency at a given drain current level. This is equivalent to fT for bipolar transistors. Since this test is performed at a fast sweep speed, heating of the device does not occur. Thus, in normal use, the higher temperatures may degrade these characteristics to some extent.

GATE CHARACTERISTICS

The gate of the RF MOSFET is a polysilicon material, and is electrically isolated from the source by a layer of oxide. The input resistance is very high — on the order of 109 ohms

— resulting in a leakage current of a few nanoamperes.

Gate control is achieved by applying a positive voltage slightly in excess of the gate–to–source threshold voltage, VGS(th).

Gate Voltage Rating — Never exceed the gate voltage rating. Exceeding the rated VGS can result in permanent damage to the oxide layer in the gate region.

Gate Termination — The gates of these devices are essentially capacitors. Circuits that leave the gate open–circuited or floating should be avoided. These conditions can result in turn–on of the devices due to voltage build–up on the input capacitor due to leakage currents or pickup.

Gate Protection — These devices do not have an internal monolithic zener diode from gate–to–source. If gate protection is required, an external zener diode is recommended.

MOUNTING OF HIGH POWER RF POWER TRANSISTORS

The package of this device is designed for conduction cooling. It is extremely important to minimize the thermal resistance between the device flange and the heat dissipator.

Since the device mounting flange is made of soft copper, it may be deformed during various stages of handling or during transportation. It is recommended that the user makes a final inspection on this before the device installation. ≥0.0005, is considered sufficient for the flange bottom. The same applies to the heat dissipator in the device mounting area. If copper heatsink is not used, a copper head spreader is strongly recommended between the device mounting surfaces and the main heatsink. It should be at least 1/4, thick and extend at least one inch from the flange edges. A thin layer of thermal compound in all interfaces is, of course, essential. The recommended torque on the 4–40 mounting screws should be in the area of 4–5 lbs.–inch, and spring type lock washers along with flat washers are recommended.

For die temperature calculations, the Δ temperature from a corner mounting screw area to the bottom center of the flange is approximately 5°C and 10°C under normal operating conditions (dissipation 150 W and 300 W respectively). The main heat dissipater must be sufficiently large and have low Rθ for moderate air velocity, unless liquid cooling is employed.
CIRCUIT CONSIDERATIONS
At high power levels (500 W and up), the circuit layout becomes critical due to the low impedance levels and high RF currents associated with the output matching. Some of the components, such as capacitors and inductors must also withstand these currents. The component losses are directly proportional to the operating frequency. The manufacturers specifications on capacitor ratings should be consulted on these aspects prior to design. Push–pull circuits are less critical in general, since the ground referenced RF loops are practically eliminated, and the impedance levels are higher for a given power output. High power broadband transformers are also easier to design than comparable LC matching networks.

EQUIVALENT TRANSISTOR PARAMETER TERMINOLOGY

Collector .............................................. Drain
Emitter .............................................. Source
Base .................................................. Gate
V(BR)CES ............................................ V(BR)DSS
VCEO .................................................. VDSO
IC ...................................................... ID
ICESS .................................................. IDSS
IEBO .................................................. IGS
VBE(on) ............................................... VGS(th)
VCE(sat) ............................................... VDS(on)
CIE ..................................................... CIESS
CBO ..................................................... COSS
gme ..................................................... gme

RCE(sat) = \frac{VCE(sat)}{IC}

RDS(on) = \frac{VDS(on)}{ID}
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