

## 3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver with 2x2 Crosspoint Switch

Rev V3

### Applications

- 3G/HD/SD-SDI switchers and routers
- SMPTE 259M, 292M, 344M, 424M, DVB ASI 270Mb/s

### Features

- Dual FR4 equalizer and output de-emphasis
- Robust operation up to 3.2Gbps
- Input equalization for up to 40" of FR4 + 2 connectors
- Output de-emphasis for up to 40" of FR4 + 2 connectors
- 2x2 crosspoint switch

- Integrated 50Ω input termination
- Loss of Signal detection at the input
- Very low power consumption (38 mW per channel @1.2V)
- On-chip regulators for operation from 1.2V to 3.3V DC supply
- Universal DC coupling at the input and output with integrated level shifter
- Industrial operating temperature range of -40°C to 85°C
- 4mm x 4mm, 24-pin QFN package

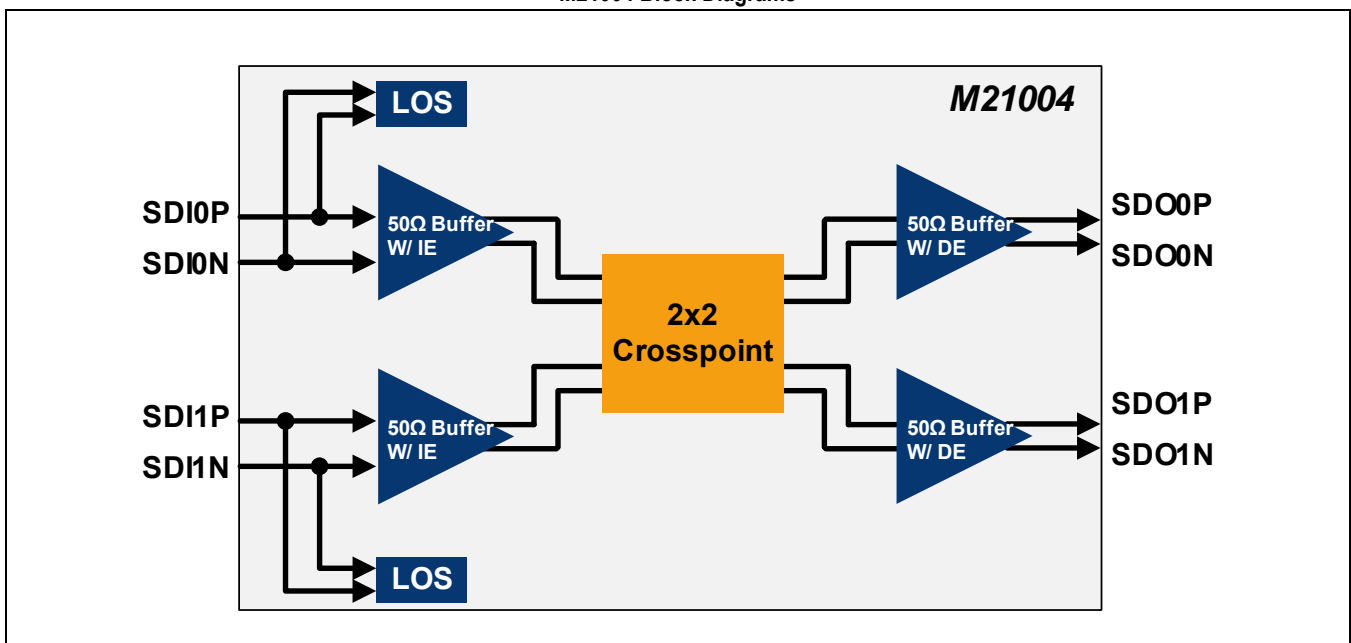
The M21004 is a very low power, highly integrated, dual backplane equalizer and redriver with optimized power and performance for Serial Digital Interface (SDI) video applications. It can also be used in non-SDI systems for data rates up to 3.2Gbps.

Each of the two independent channels has a 50Ω input buffer with configurable input equalizer, capable of compensating for losses across 40" of FR4 and two connectors. Each channel also includes a 50Ω output buffer with configurable de-emphasis to aid transmission of the signal across an additional 40" of FR4 trace and two connectors. In addition, the M21004 features a non-blocking 2x2 crosspoint switch. The switch allows either input to be routed to any or both of the outputs.

The device has integrated internal supply regulators, allowing it to be powered from a single 1.2V, 1.8V, 2.5V, or 3.3V supply voltage. The power rails for the input and output circuitry are electrically independent from each other and the core supply and thus may be connected to a different voltage rail on the board. This feature enables the M21004 to be DC coupled to any upstream and downstream device in the 1.2V to 3.3V range without level shifting.

The M21004 is offered in a green and RoHS compliant 24-pin QFN package.

M21004 Block Diagrams



1

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## 3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver with 2x2 Crosspoint Switch

Rev V3

### Ordering Information

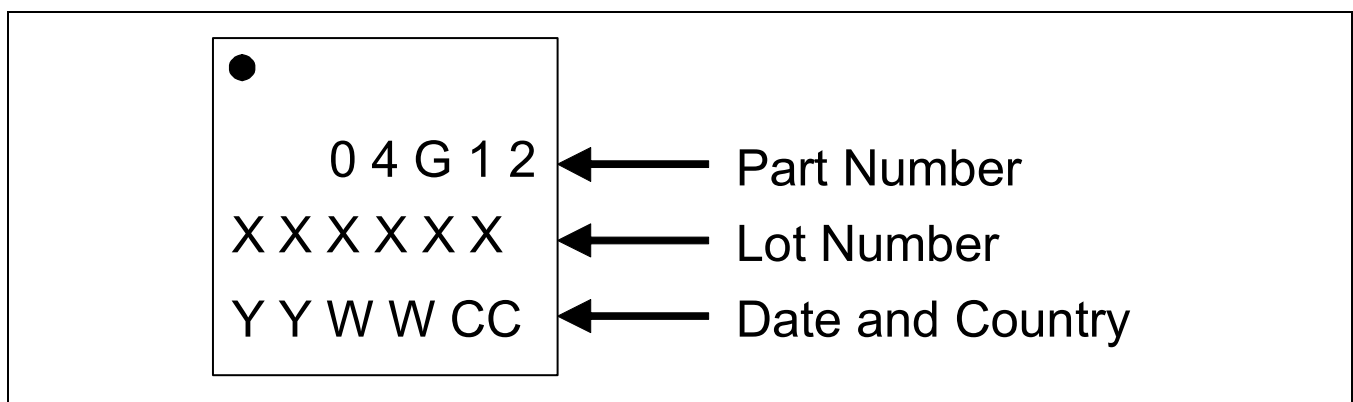
Part Number	Package	Operating Temperature
M21004G-12*	4 mm, 24-pin QFN (RoHS compliant)	-40 °C to 85 °C

\* The letter "G" designator after the part number indicates that the device is RoHS compliant. The RoHS compliant devices are backwards compatible with 225 °C reflow profiles.

### Revision History

Revision	Level	Date	Description
V3	Release	October 2017	Added <a href="#">Tables 1-5</a> Logic output Characteristic Updated <a href="#">Tables 4-5</a> XPT Control Updated <a href="#">Tables 1-4</a> Output Characteristic
V2	Release	December 2015	Updated Package Drawing, <a href="#">Figure 3-9</a> and <a href="#">Figure 3-10</a> . Package effective as of July 2014.
C (V1)	Release	March 2010	Added Marking Diagram. Added final characterization figures to <a href="#">Tables 1-3</a> , <a href="#">1-4</a> , <a href="#">1-5</a> . Added <a href="#">Figures 3-2</a> to <a href="#">3-8</a> . Added $\theta_{JA}$ to <a href="#">Table 1-3</a> . Added recommended 10 $\mu$ F input caps in <a href="#">Section 4.1</a> and <a href="#">4.3</a> . Added <a href="#">Figures 2-3</a> , <a href="#">2-4</a> and <a href="#">Section 4.2</a> and <a href="#">4.3</a> .
B (V2A)	Advance	November 2009	Updated power, added jitter, DCD rise/fall time figures. Add <a href="#">Section 4.6.4</a> , <a href="#">4.6.5</a> , <a href="#">4.6.6</a> .
A (V1A)	Advance	July 2009	Initial Release

### Marking Diagram



## Table of Contents

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Ordering Information .....	2
Revision History .....	2
Table of Contents .....	3
1.0 Electrical Characteristics .....	4
2.0 Typical Performance Characteristics .....	7
3.0 Pinout Diagram, Pin Descriptions, and Packaging Outline Drawing .....	8
3.1 Package Drawings and Surface Mount Details .....	16
4.0 Functional Description .....	19
4.1 High Speed Input Description .....	20
4.2 Input Circuit Power .....	21
4.2.1 AC Coupled Configuration .....	21
4.2.2 DC Coupled Configuration .....	22
4.2.3 Self Biased Configuration .....	23
4.3 High-Speed Output Description .....	23
4.4 Power Supply Description .....	25
4.5 Power Up Sequence .....	27
4.6 Logic Control Signals .....	27
4.6.1 Input Equalizer Control .....	27
4.6.2 Output De-emphasis Control .....	28
4.6.3 Output Swing Control .....	28
4.6.4 LOS/Mute Control .....	28
4.6.5 XPT Control .....	29
4.6.6 Regulator Enable .....	29
4.7 Typical Application Circuit .....	30

# 1.0 Electrical Characteristics

Unless noted otherwise, specifications in this section apply to nominal power supply, 25 °C ambient temperature, 800 mVpp input data swing, default output data swing, PRBS 2<sup>15</sup> – 1 test pattern, RL = 50Ω. voltages are referenced to AV<sub>SS</sub>.

**Table 1-1. Absolute Maximum Ratings**

Symbol	Parameter	Note	Minimum	Typical	Maximum	Unit
AV <sub>DD0,1</sub>	Analog Core power supply voltage	1	-0.5	—	1.5	V
AV <sub>DDOUT</sub>	Analog power Output supply voltage	1	-0.5	—	3.6	V
AV <sub>DDIN</sub>	Analog power Input supply voltage	1	-0.5	—	3.6	V
V <sub>IN,PCML</sub>	DC input voltage (PCML)	1	-0.5	—	AV <sub>DDOUT</sub> + 0.5	V
V <sub>IN,CMOS</sub>	DC input voltage (CMOS)	1	-0.5	—	AV <sub>DDOUT</sub> + 0.5	V
T <sub>STORE</sub>	Storage temperature	1	-65	—	150	°C
T <sub>JUNC</sub>	Junction temperature	1	-40	—	125	°C
V <sub>ESD,HBM</sub>	Electrostatic discharge voltage (HBM)	1, 2	—	—	4	kV
V <sub>ESD,CDM</sub>	Electrostatic discharge voltage (CDM)	1, 2	—	—	500	V

**NOTES:**

- Exposure of the device beyond the minimum/maximum limits may cause permanent damage. Limits listed in the above table are stress limits only, and do not imply functional operation within these limits.
- HBM and CDM per JEDEC Class 2 (JESD22-A114-B).

**Table 1-2. Recommended Operating Conditions**

Symbol	Parameter	Note	Minimum	Typical	Maximum	Unit
AV <sub>DD0,1</sub>	Analog Core supply voltage	—	1.14	1.2	1.26	V
AV <sub>DDOUT</sub>	Analog Output supply voltage	—	1.14	1.2/1.8/2.5/3.3	3.47	V
AV <sub>DDIN</sub>	Analog Input supply voltage	—	1.14	1.2/1.8/2.5/3.3	3.47	V
T <sub>CASE</sub>	Operating temperature	1	-40	—	85	°C

**NOTES:**

- Case temperature.

## 3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver with 2x2 Crosspoint Switch

Rev V3

**Table 1-3. Power Consumption Specifications**

Symbol	Parameter	Note	Minimum	Typical	Maximum	Unit
$I_{DD}CORE$	Core current consumption	1	—	35	50	mA
$I_{DD}OUT$	Output current consumption	1	—	30	40	mA
$I_{DD}IN$	Input current consumption	1, 5	—	0.7	1	mA
$I_{DD}TOTAL$	Total current consumption	1, 5	—	65	86	mA
$P_{TOTAL}$	Power consumption	1, 5	—	75	108	mW
$I_{DD}CORE$	Core Current consumption	2	—	35	50	mA
$I_{DD}OUT$	Output Current consumption	2	—	40	50	mA
$I_{DD}IN$	Input current consumption	2, 5	—	0.7	1	mA
$I_{DD}TOTAL$	Total current consumption	2, 5	—	75	101	mA
$P_{TOTAL}$	Power consumption	2, 5	—	90	128	mW
$I_{DD}CORE$	Core Current consumption	3	—	40	60	mA
$I_{DD}OUT$	Output Current consumption	3	—	60	80	mA
$I_{DD}IN$	Input current consumption	3, 5	—	8	10	mA
$I_{DD}TOTAL$	Total current consumption	3, 5	—	108	150	mA
$P_{TOTAL}$	Power consumption	3, 5	—	272	388	mW
$I_{DD}OUT$	Output Current consumption	4	—	100	140	mA
$I_{DD}IN$	Input current consumption	4, 5	—	8	10	mA
$I_{DD}TOTAL$	Total current consumption	4, 5	—	108	150	mA
$P_{TOTAL}$	Power consumption	4, 5	—	356	520	mW
$\theta_{JA}$	Junction to ambient thermal resistance	6	—	60	—	°C/W

**NOTES:**

1.  $AV_{DD}CORE = 1.2V$ ,  $AV_{DD}IN$ ,  $AV_{DD}OUT = 1.2V$  and low output swing setting.
2.  $AV_{DD}CORE = 1.2V$ ,  $AV_{DD}IN$ ,  $AV_{DD}OUT = 1.2V$  and med output swing setting.
3.  $AV_{DD}CORE = 1.2V$ ,  $AV_{DD}IN$ ,  $AV_{DD}OUT = 3.3V$  and high output swing setting.
4.  $AV_{DD}IN$ ,  $AV_{DD}OUT = 3.3V$ , Regulator Enabled and high output swing setting.
5. See Section 4.3 for additional current drawn by the input termination.
6. Airflow = 0 m/s.

## 3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver with 2x2 Crosspoint Switch

Rev V3

**Table 1-4. PCML Input/Output Electrical Characteristics**

Symbol	Parameter	Note	Minimum	Typical	Maximum	Unit
DR	NRZ data rate	—	143	—	3200	Mbps
V <sub>IN</sub>	Differential Input Voltage	1	250	800	1600	mVppd
IE	Input equalization	—	—	6, 4, 0	—	dB
R <sub>IN</sub>	Input termination resistance	2	—	50	—	Ω
V <sub>LOSA</sub>	LOS level, assert	1	70	75	85	mVpp
V <sub>LOSD</sub>	LOS level, deassert	1	—	145	160	mVpp
V <sub>OUT</sub>	PCML differential output swing - low	5	485	600	720	mVppd
	PCML differential output swing - med	5, 6	680	800	960	mVppd
	PCML differential output swing - high	4, 5	1000	1200	1440	mVppd
R <sub>OUT</sub>	Output termination resistance	3	—	50	—	Ω
DE	Output de-emphasis settings	7	—	6, 4, 0	—	dB
J <sub>OUT</sub>	Total Output Jitter	8, 9	—	95	170	mUI
DCD <sub>O</sub>	Output Duty Cycle distortion	8	—	—	16	ps
tr/tf	Rise/Fall Time	—	—	80	135	ps

**NOTES:**

- Value specified at the device pins.
- Internal termination to AV<sub>DD</sub>IN.
- Internal termination to AV<sub>DD</sub>OUT.
- To achieve high swing; AV<sub>DD</sub>OUT must be > =1.8V.
- Measured into 50Ω load.
- Default output swing level.
- Measured with 16 ones and 16 zeros pattern.
- Measured at 3.2 Gbps
- Measured point blank, BER = 10<sup>-12</sup>

**Table 1-5. Control/Interface Logic Input/Output Characteristics**

Symbol	Parameter	Note	Minimum	Typical	Maximum	Unit
V <sub>IH</sub>	Input logic high		0.85 x AV <sub>DD</sub> OUT	—	AV <sub>DD</sub> OUT	V
V <sub>IF</sub>	Input logic float		0.25 x AV <sub>DD</sub> OUT	—	0.75 x AV <sub>DD</sub> OUT	V
V <sub>IL</sub>	Input logic low		0	—	0.15 x AV <sub>DD</sub> OUT	V
V <sub>OH</sub>	Output logic high		0.8 x AV <sub>DD</sub> OUT	AV <sub>DD</sub> OUT	—	V
V <sub>OL</sub>	Output logic low		—	0	0.2 x AV <sub>DD</sub> OUT	V
I <sub>IL</sub>	Input Current logic low		100	—	—	uA
I <sub>IH</sub>	Input Current logic high		—	—	-100	uA

6

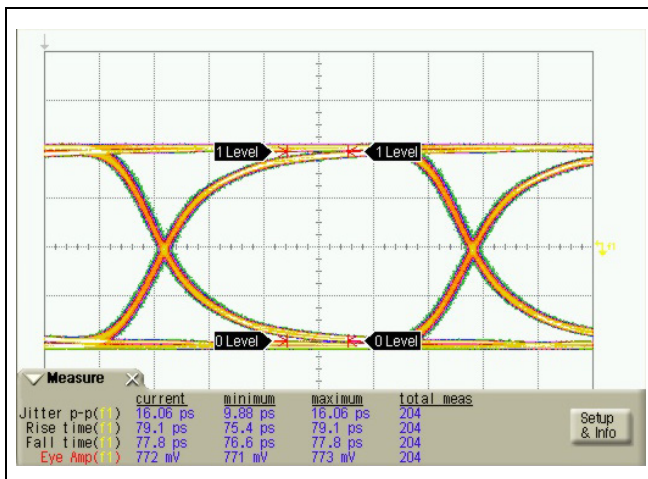
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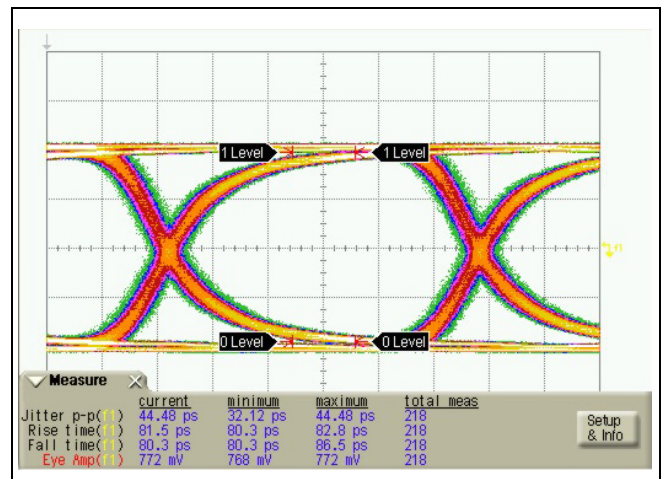
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# 2.0 Typical Performance Characteristics

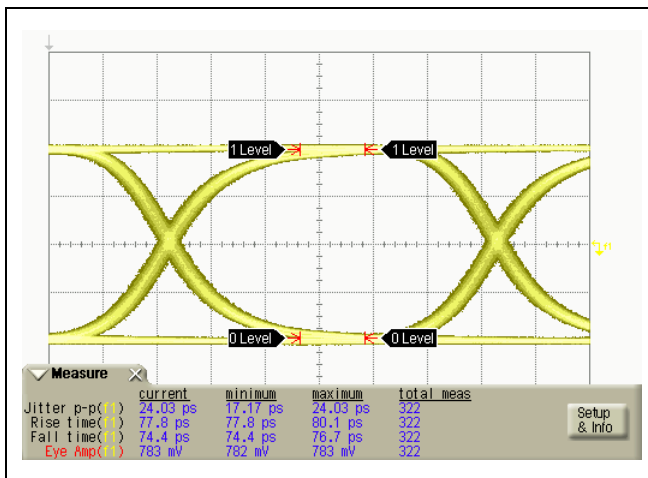
**Figure 2-1. Eye Diagram @ 0" FR4, 3.2 Gbps,  $AV_{DD\_CORE} = AV_{DD\_IN} = AV_{DD\_OUT} = 1.2V$ , Medium Swing**



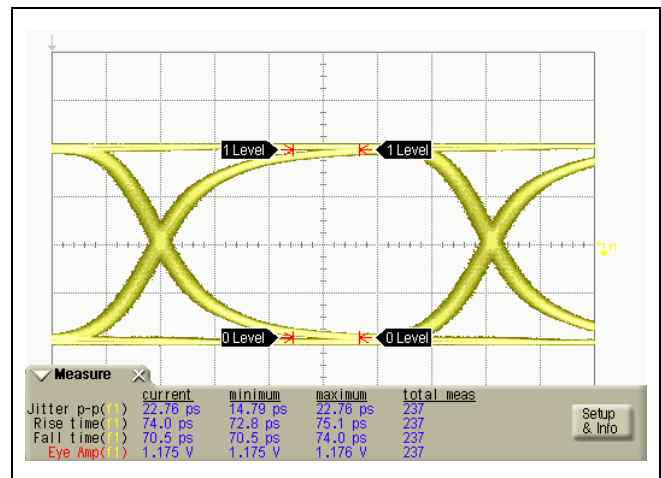
**Figure 2-2. Eye Diagram @ 40" FR4, 3.2 Gbps,  $AV_{DD\_CORE} = AV_{DD\_IN} = AV_{DD\_OUT} = 1.2V$ , Medium Swing**



**Figure 2-3. Eye Diagram @ 0" FR4, 3.2 Gbps,  $AV_{DD\_CORE} = AV_{DD\_IN} = 1.2V$ ,  $AV_{DD\_OUT} = 1.8V$ , Medium Swing**

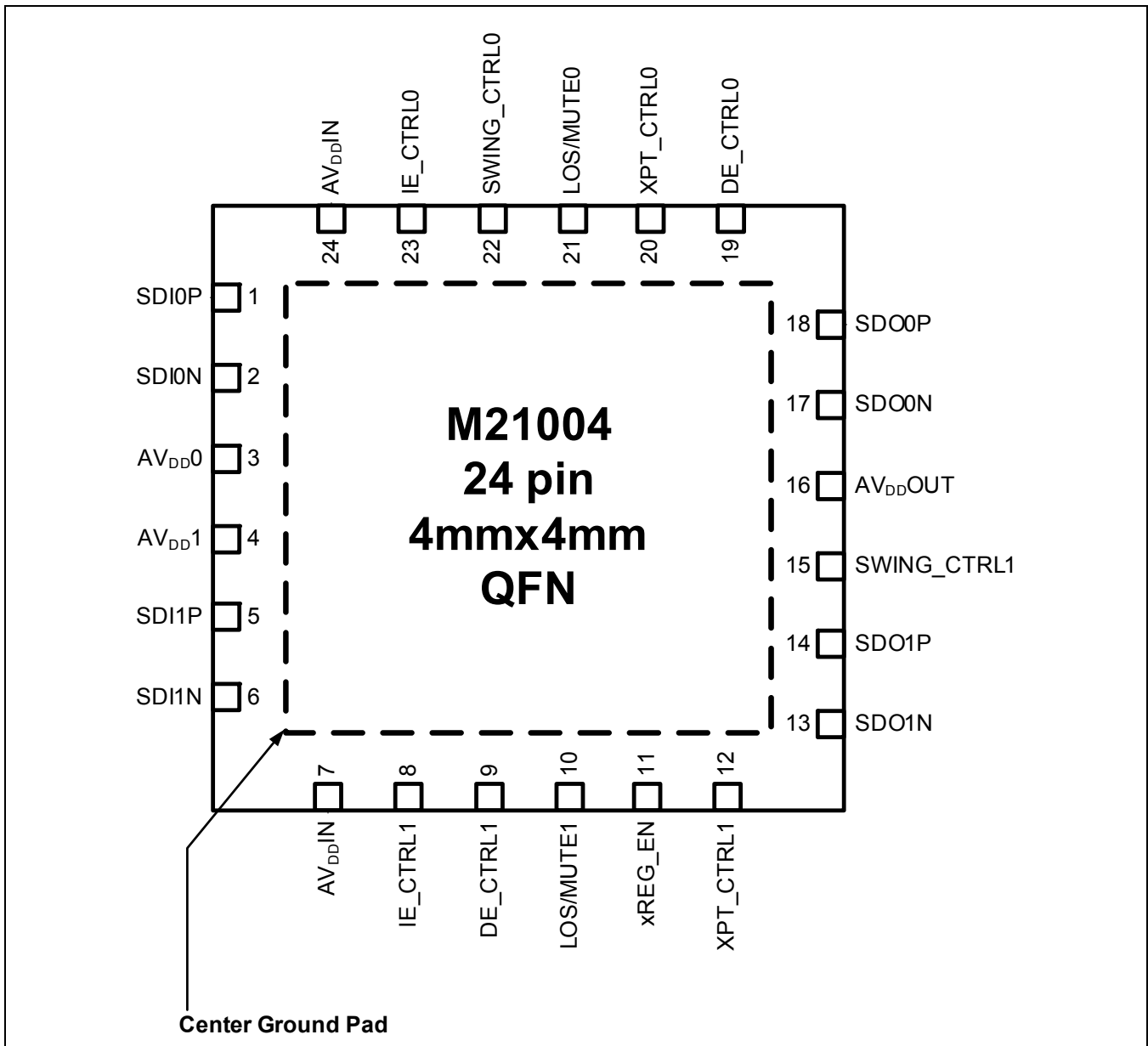


**Figure 2-4. Eye Diagram @ 0" FR4, 3.2 Gbps,  $AV_{DD\_CORE} = AV_{DD\_IN} = 1.2V$ ,  $AV_{DD\_OUT} = 1.8V$ , High Swing**



## 3.0 Pinout Diagram, Pin Descriptions, and Packaging Outline Drawing

Figure 3-1. M21004 Pinout Diagram (Top View)





**3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver  
with 2x2 Crosspoint Switch**

Rev V3

**Table 3-1. M21004 Pin Descriptions (1 of 2)**

Pin Name	Pin Number(s)	Type	Description
AV <sub>SS</sub>	Center Pad	Power	Ground
AV <sub>DD0</sub>	3	Power	Analog Core positive supply for channel 0
AV <sub>DD1</sub>	4	Power	Analog Core positive supply for channel 1
AV <sub>DDOUT</sub>	16	Power	Analog positive supply for output circuitry
AV <sub>DDIN</sub>	24, 7	Power	Analog positive supply for input circuitry
xREG_EN	11	I-Digital with pull up	Internal regulator disable L = Enable integrated regulator H = disable integrated regulator (default)
XPT_CTRL[1:0]	12, 20	I-Digital: XPT_CTRL1 with pull up, XPT_CTRL0 with pull down	Input Crosspoint Control L L = Broadcast SDI0; SDI0 to SDO0, SDI0 to SDO1 L H = Crossover; SDI1 to SDO0, SDI0 to SDO1 H L = Feedthrough; SDI0 to SDO0, SDI1 to SDI1 (default) H H = Broadcast SDI1; SDI1 to SDO0, SDI1 to SDI1
SWING_CTRL0, 1	22, 15	3-state/ I-Digital	Output swing control for channel 0 and channel 1 L = Low F = Medium (default) H = High
DE_CTRL0, 1	19, 9	3-state/ I-Digital	Output de-emphasis control for channel 0 and channel 1 L = DE off F = Medium DE (default) H = High DE
IE_CTRL0, 1	23, 8	3-state/ I-Digital	Input Equalization control for channel 0 and channel 1 L = IE off F = Medium IE (default) H = High IE
LOS/MUTE0, 1	21, 10	O-Digital/ I-Digital	Configured as output (> 50 k $\Omega$ resistive load): LOS alarm output (active high) for channel 0 and channel 1 Configured as input (driven with R < 0.25 k $\Omega$ ) L = never mute the output H = force mute the output
SDI0P	1	I-Analog	Serial Data video input0, true
SDI0N	2	I-Analog	Serial Data video input0, complement
SDI1P	5	I-Analog	Serial Data video input1, true
SDI1N	6	I-Analog	Serial Data video input1, complement
SDO0P	18	O-Analog	Serial Data output0, true

9

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DC-0017546

## 3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver with 2x2 Crosspoint Switch

Rev V3

**Table 3-1. M21004 Pin Descriptions (2 of 2)**

Pin Name	Pin Number(s)	Type	Description
SDO0N	17	O-Analog	Serial Data output0, complement
SDO1P	14	O-Analog	Serial Data output1, true
SDO1N	13	O-Analog	Serial Data output1, complement

**NOTE:** The default state is controlled by pull up/pull down resistors of 100kΩ to AV<sub>DD</sub>OUT or AV<sub>SS</sub>.

**Figure 3-2. I-Analog**

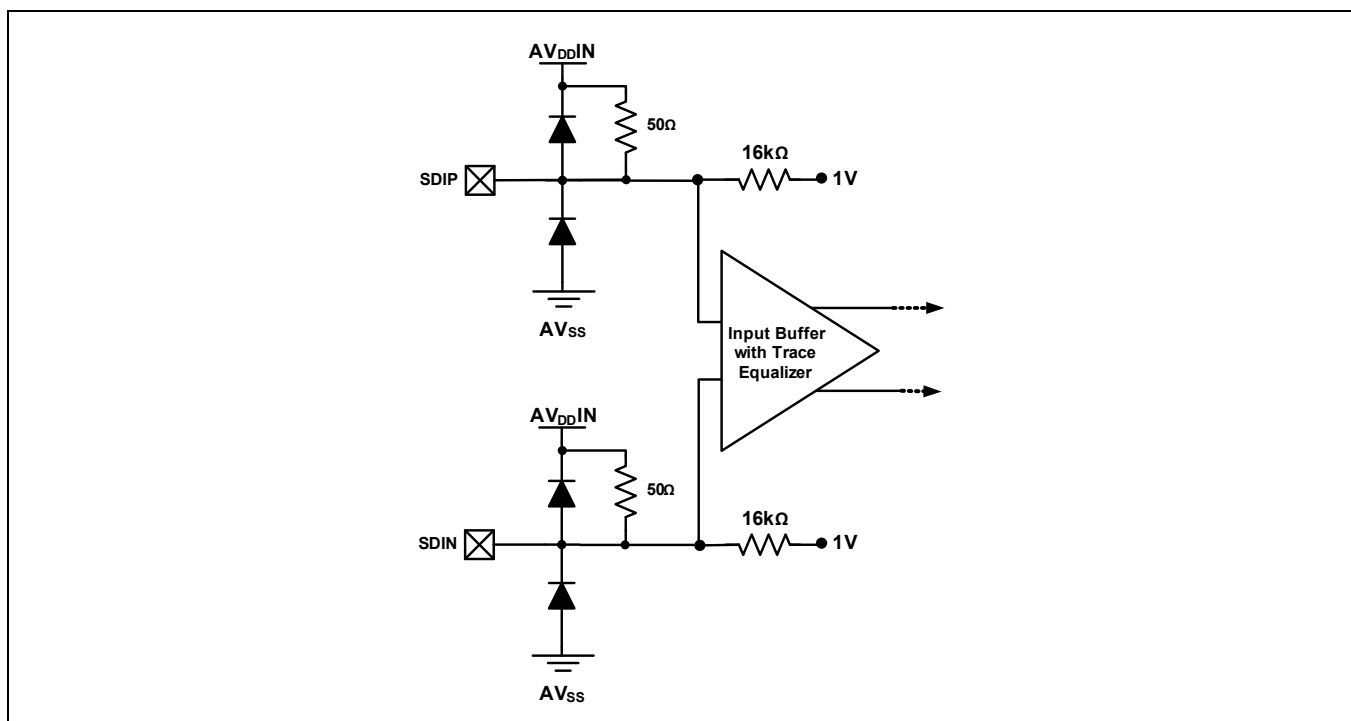


Figure 3-3. O-Analog

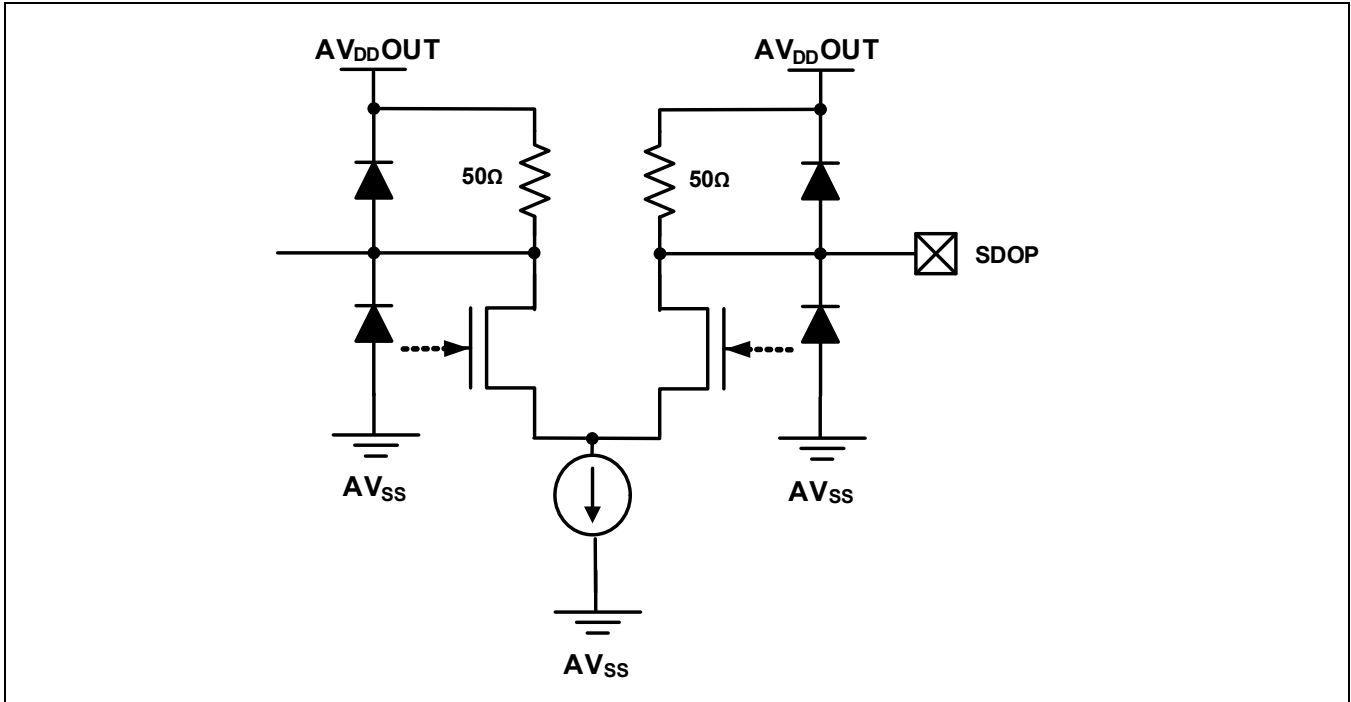


Figure 3-4. I-Digital With Pull-up

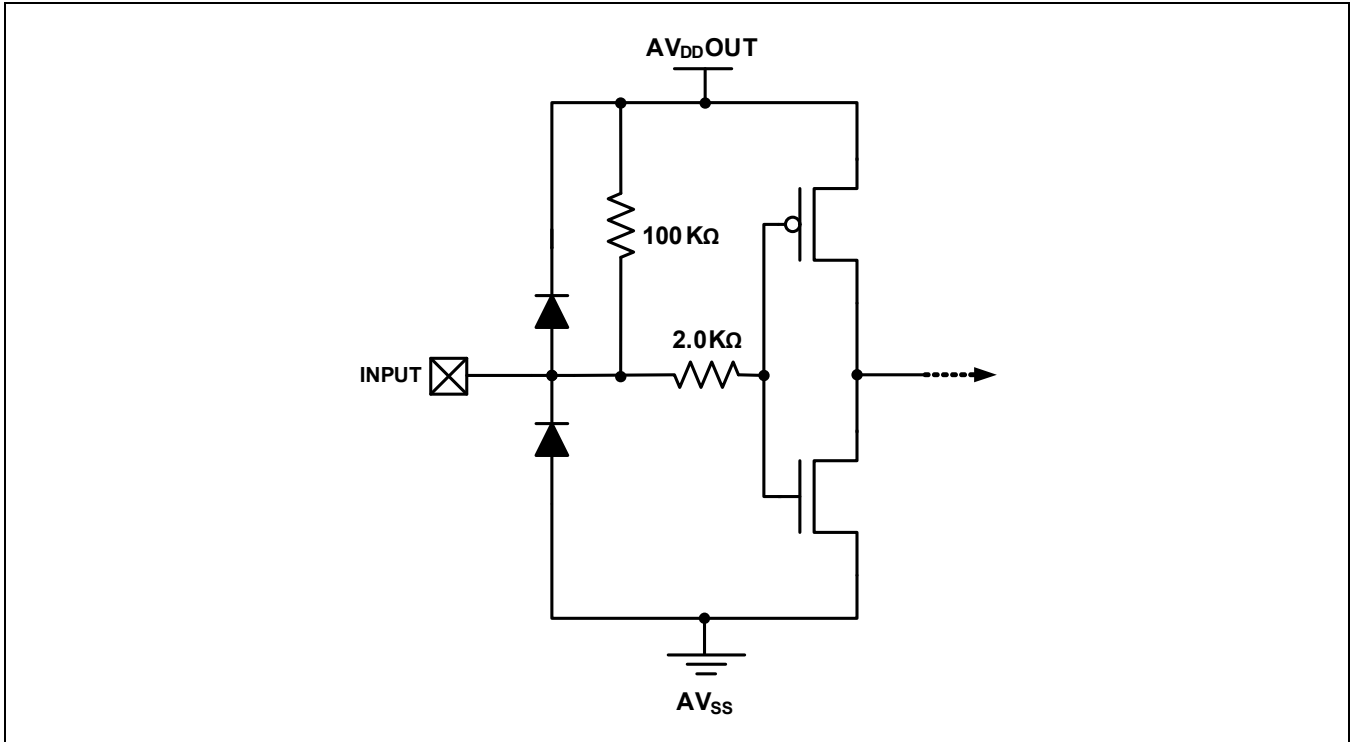


Figure 3-5. I-Digital With Pull-down

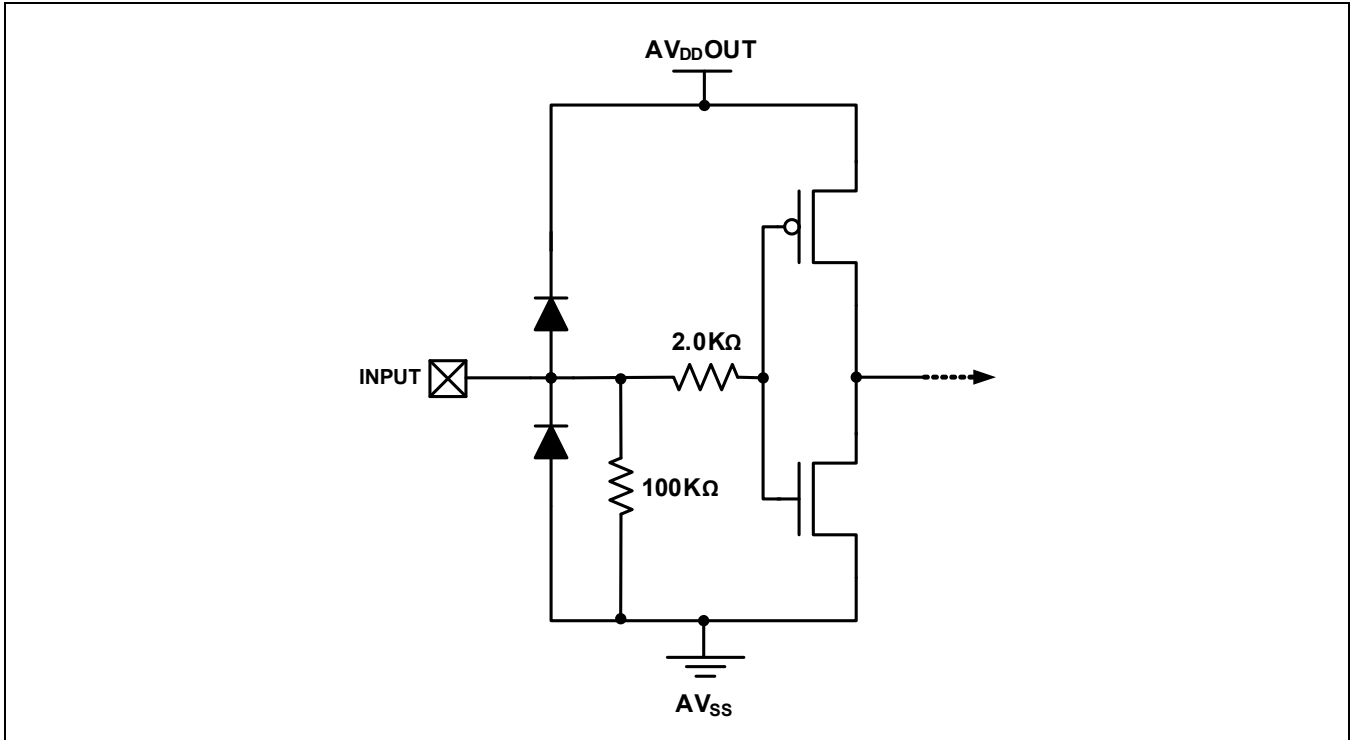


Figure 3-6. 3-State/I-Digital

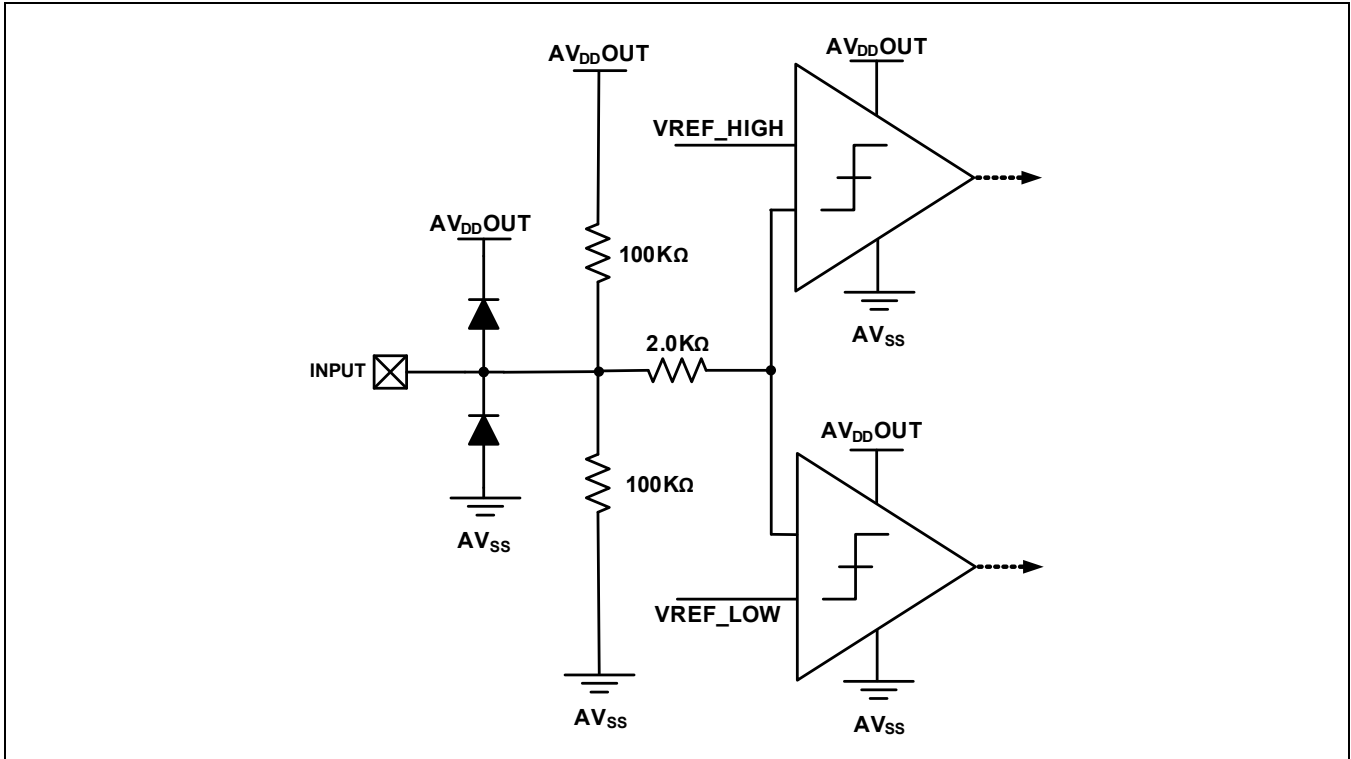
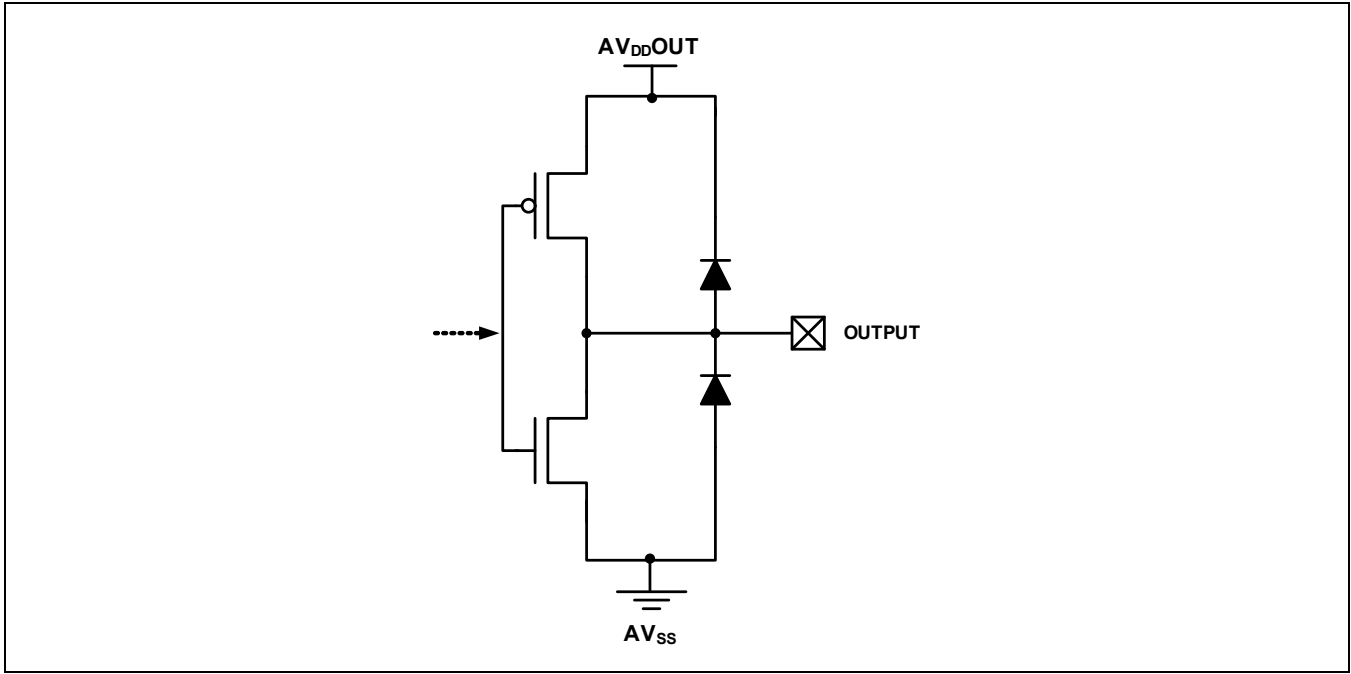


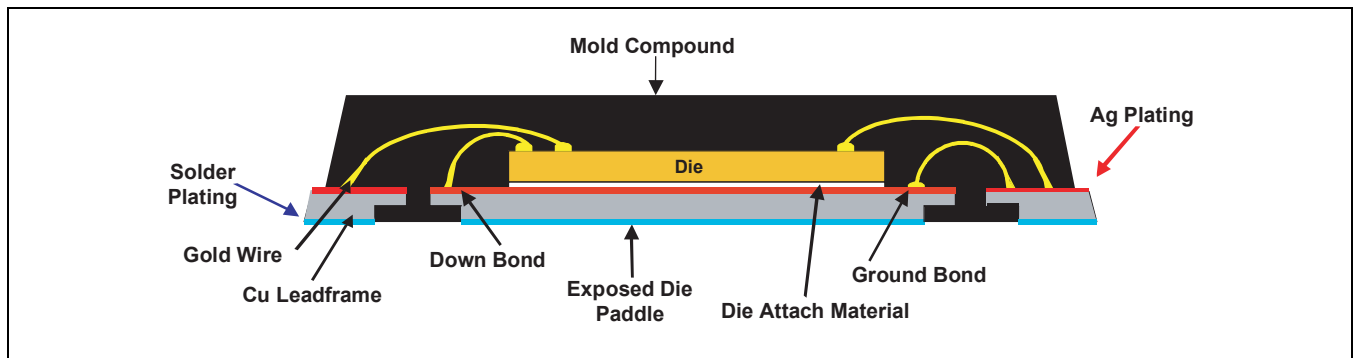
Figure 3-7. O-Digital



### 3.1 Package Drawings and Surface Mount Details

The M21004 is assembled in a 24-pin, 4 mm x 4 mm Quad Flat No-Lead (QFN) package. The exposed die paddle serves as the IC ground ( $AV_{SS}$ ), and the primary means of thermal dissipation. This die paddle should be soldered to the PCB ground. A cross-section of the QFN package can be found in [Figure 3-8](#).

**Figure 3-8. QFN Package Cross Section**

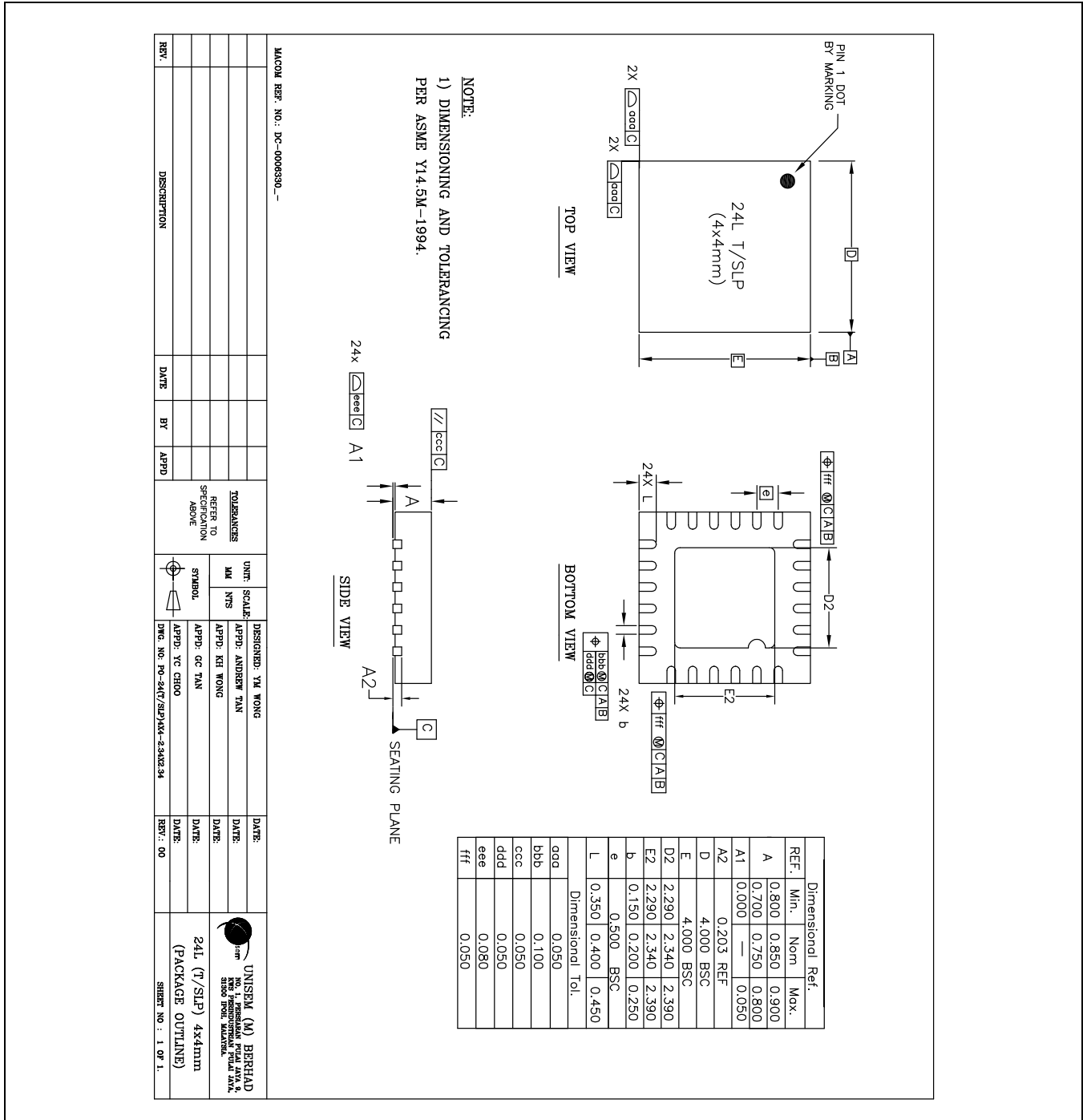




## 3G/HD/SD-SDI Low Power Backplane Equalizer and Redriver with 2x2 Crosspoint Switch

Rev V3

Figure 3-9. M21004 Package Drawing



- For dimension reference A, Min. is 0.800, Nom. is 0.850 and Max. is 0.900.
- New Unisem package and the old Amkor package have the same footprint.

Figure 3-10. M21004 24-Pin Package Dimensions

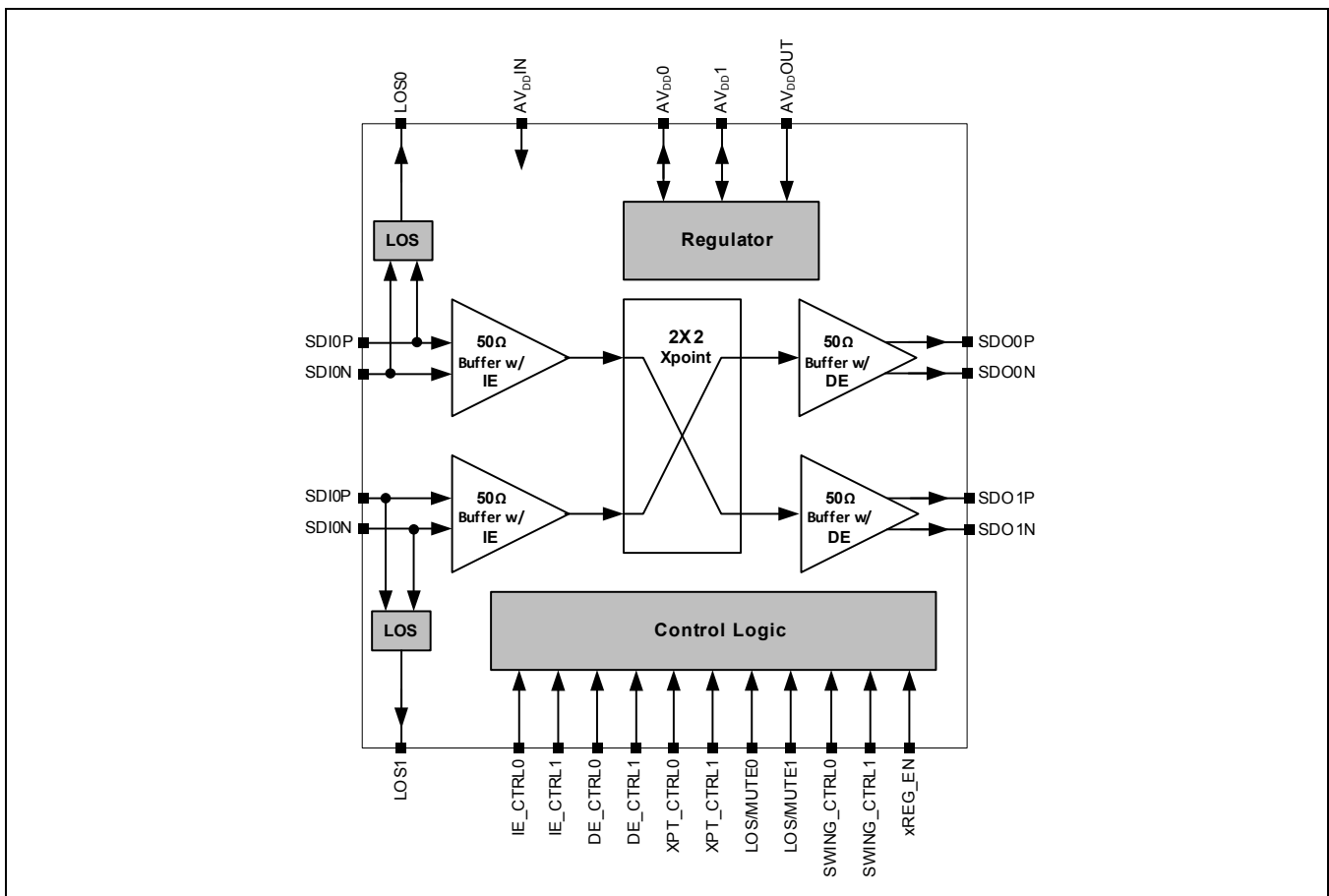
Dimensional Ref.			
REF.	Min.	Nom	Max.
A	0.800	0.850	0.900
	0.700	0.750	0.800
A1	0.000	—	0.050
A2	0.203 REF		
D	4.000 BSC		
E	4.000 BSC		
D2	2.290	2.340	2.390
E2	2.290	2.340	2.390
b	0.150	0.200	0.250
e	0.500 BSC		
L	0.350	0.400	0.450
Dimensional Tol.			
aaa	0.050		
bbb	0.100		
ccc	0.050		
ddd	0.050		
eee	0.080		
fff	0.050		

1. For dimension reference A, Min. is 0.800, Nom. is 0.850 and Max. is 0.900.
2. New Unisem package and the old Amkor package have the same footprint.

### 4.0 Functional Description

Figure 4-1 illustrates the functional block diagram of the M21004. The subsequent sections provide additional detail on the operation of the device.

Figure 4-1. M21004 Functional Block Diagram



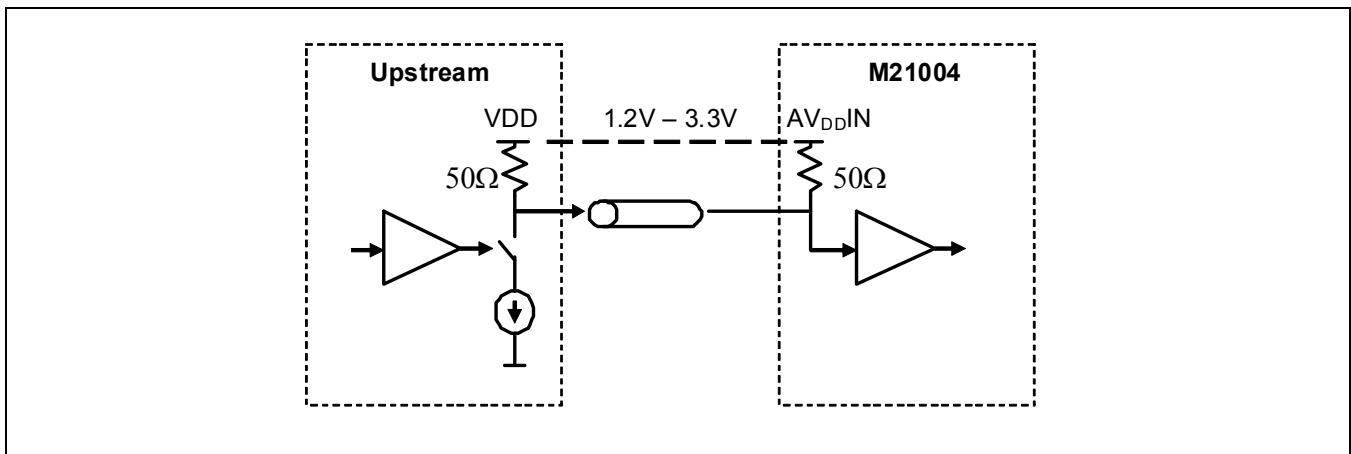
### 4.1 High Speed Input Description

The M21004 features two inputs with a 50Ω termination to AV<sub>DD</sub>IN. AV<sub>DD</sub>IN can be supplied from any voltage ranging from 1.2V to 3.3V.

In order to improve signal integrity when used in large systems, each input also comes equipped with programmable input equalization (IE) for FR4 trace. There are three settings for input equalization: 6 dB, 4 dB and 0 dB (or no equalization). The IE for each input channel is controlled through the corresponding three state control pin: IE\_CTRL0 or IE\_CTRL1.

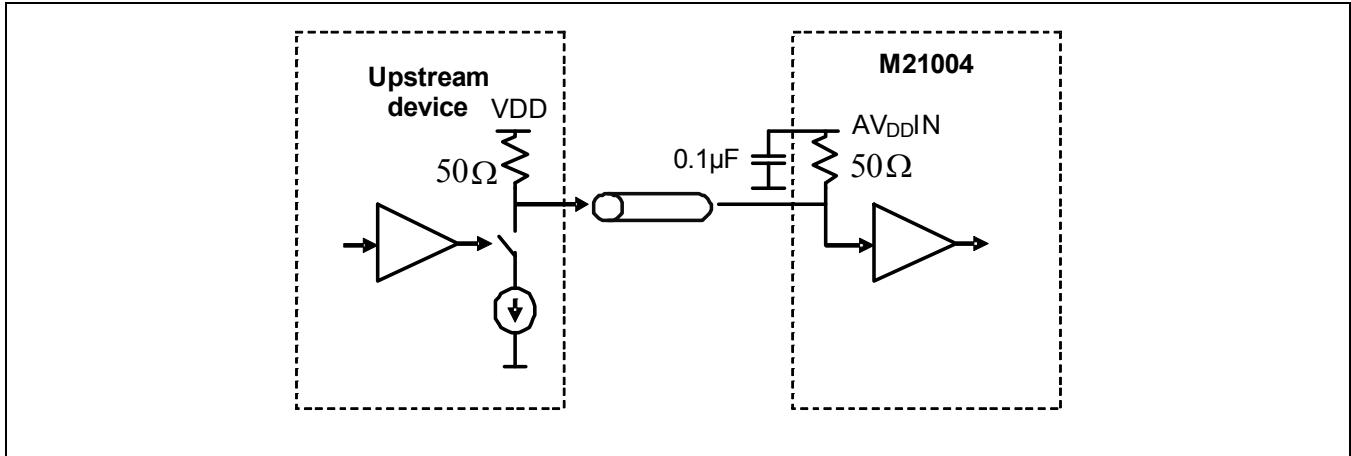
In most SDI applications, it is important to avoid AC coupled data interfaces between devices wherever possible. In addition to reducing the number of components, DC coupling will result in more system jitter margin. In order to accommodate DC coupling with the upstream device, the AV<sub>DD</sub>IN power domain of the M21004 is electrically independent from all other power domains allowing it to be tied to the VDD of the upstream device. This is demonstrated in Figure 4-2 below.

**Figure 4-2. M21004 AV<sub>DD</sub>IN Connected to the VDD of the Upstream Device**



Alternatively and provided that the internal regulators are not used, the M21004 allows for the input to be self biased, eliminating the need for an electrical connection between the supply voltages of the upstream device and M21004. This configuration offers the benefit of keeping the supply of the previous device and the power domain(s) of the M21004 completely isolated, while still allowing DC coupling. This self biasing scheme is demonstrated in Figure 4-3 below.

Figure 4-3. Self Biasing the Input of M21004



In this configuration, the minimum input common mode that can be tolerated is 600 mV. If AC coupling is desired or necessary, Because of the low frequency content of 3G level B pathological patterns, the coupling capacitor should be at least 10  $\mu$ F, when used for SDI applications.

## 4.2 Input Circuit Power

Due to the unique architecture of the M21004 front end, its current draw is dependent on the input configuration, as well as swing and common mode voltages.

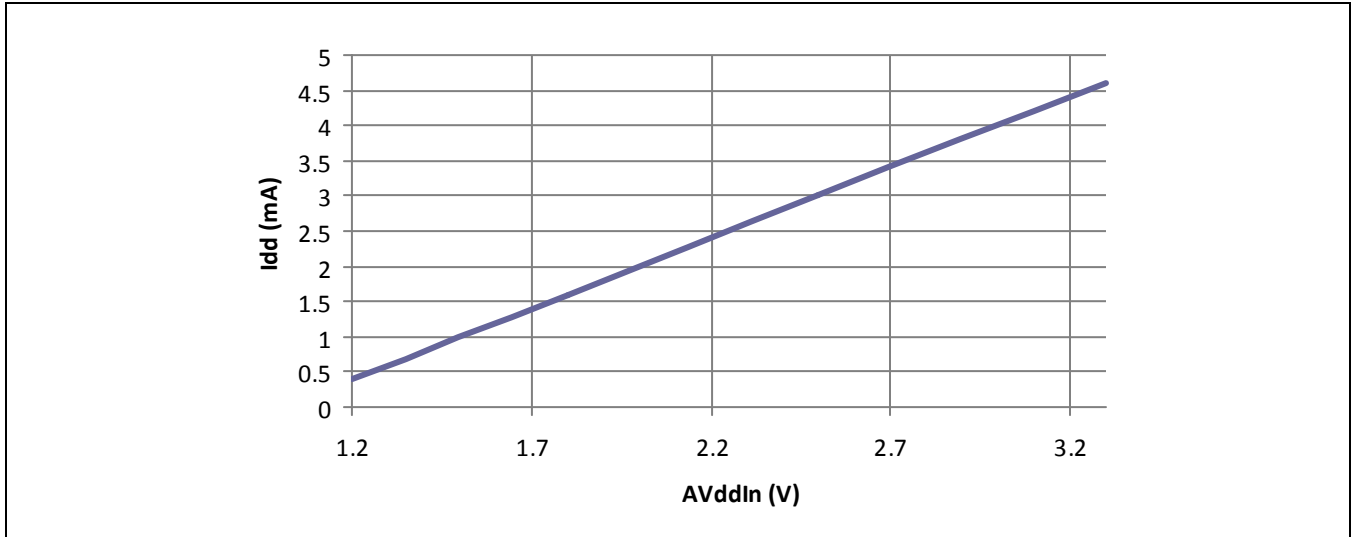
### 4.2.1 AC Coupled Configuration

In this configuration, the current is drawn from  $AV_{DDIN}$ :

$$I_{DC} = 2(AV_{DDIN} - 1)mA$$

$I_{DC}$  is the current drawn per differential input used, also see graph in [Figure 4-4](#).

Figure 4-4. Current drawn per differential input, when AC Coupled



## 4.2.2 DC Coupled Configuration

In this configuration, the current drawn is the sum of the AC Coupled case current drawn plus an additional bias current from the upstream driver. Note that the input common mode voltage  $V_{CM_{IN}}$  needs calculating first and this depends on the input signal swing:

$$V_{CM_{IN}} = (AV_{DD_{IN}} + V_{swing} / 4)V$$

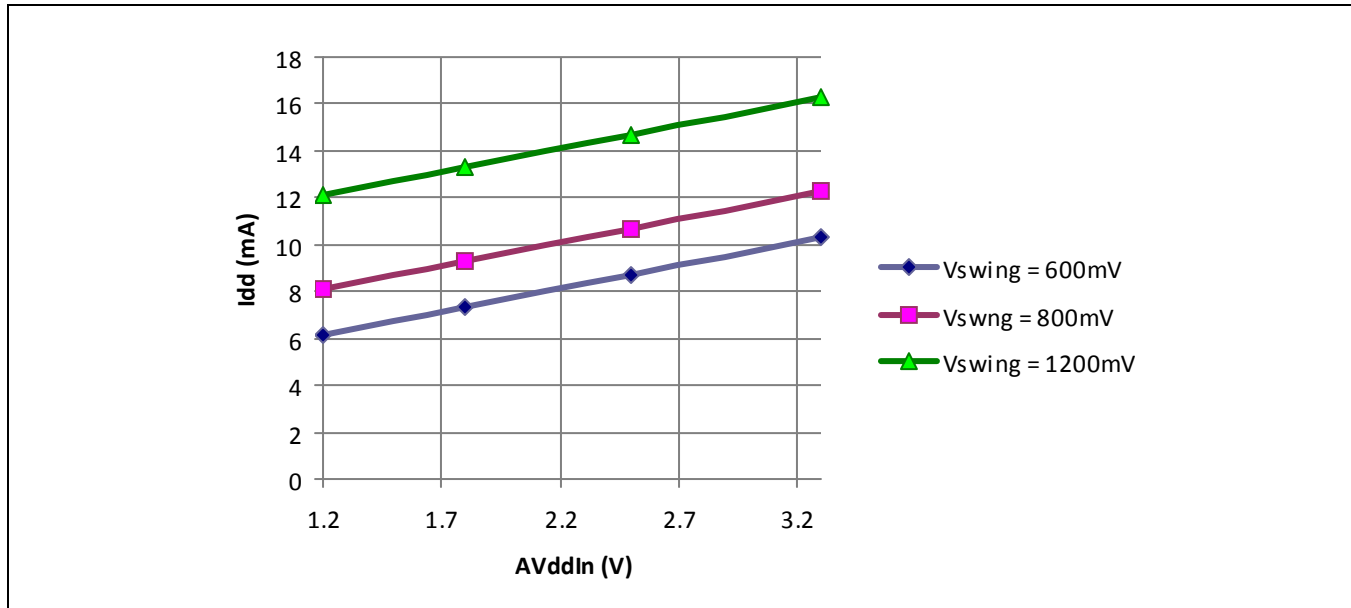
where  $V_{swing}$  is the differential input voltage peak to peak

This assumes that the upstream output driver has its  $50\Omega$  termination to the same voltage level as  $AV_{DD_{IN}}$ .

$$I_{DC} = 2(V_{CM_{IN}} - 1) + 10(V_{swing})mA$$

$I_{DC}$  is the current drawn per differential input used, also see graph in see [Figure 4-5](#).

Figure 4-5. Current drawn per differential input, when DC Coupled



### 4.2.3 Self Biased Configuration

In this configuration the DC current draw is the same as the AC coupled configuration. However, as no voltage is applied to  $A_{VDDIN}$ , the current is drawn from the VDD of the upstream driver device, see Figure 4-4.

## 4.3 High-Speed Output Description

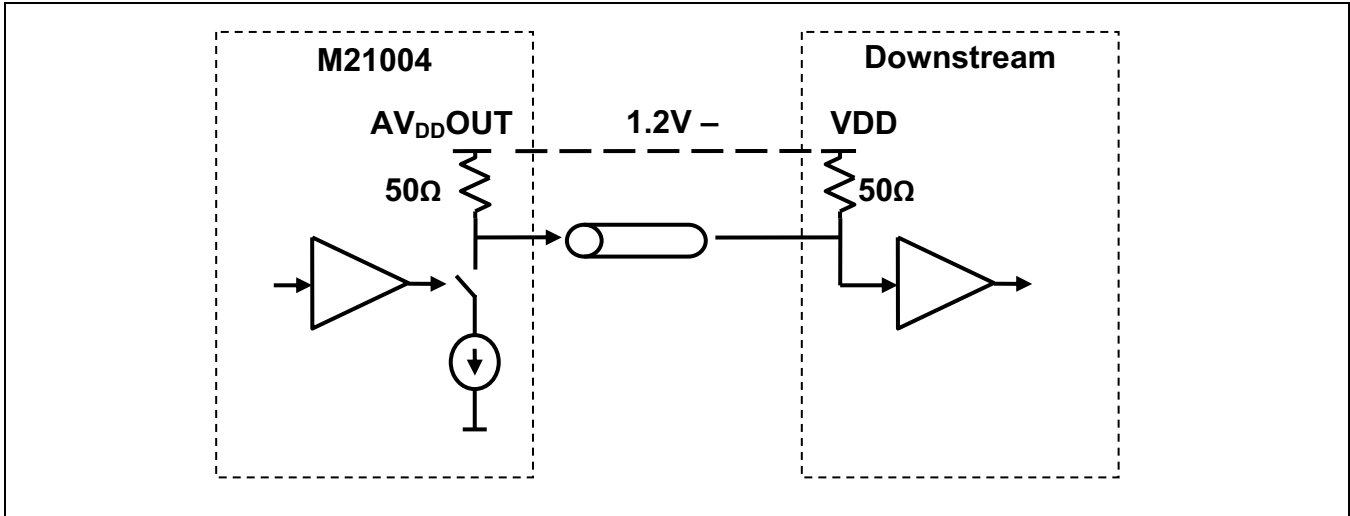
The M21004 features differential positive current mode logic (PCML) drivers with integrated  $50\Omega$  pull ups to  $A_{VDDOUT}$ .  $A_{VDDOUT}$  may be supplied from any voltage ranging from 1.2V to 3.3V.

The differential, peak-to-peak output swing for each PCML driver is selectable and may be set to low, medium, or high through the SWING\_CTRL pin. Please note that the high output swing setting is only available when  $A_{VDDOUT}$  is supplied from a voltage of 1.8V or greater.

In order to improve signal integrity when used in large systems, each output also comes equipped with programmable de-emphasis (DE) for FR4 trace. There are three settings for output de-emphasis: 0 dB (or no DE), 4 dB, and 6 dB. The de-emphasis level for each output is set through the DE\_CTRL0 and DE\_CTRL1 pins.

In most SDI applications, it is important to avoid AC coupled data interfaces between devices wherever possible. In addition to reducing the number of components, DC coupling will result in more system jitter margin. In order to accommodate DC coupling with the downstream device, the  $A_{VDDOUT}$  power domain of the M21004 is electrically independent from all other power domains, therefore allowing it to be tied to the VDD of the downstream device. This is demonstrated in Figure 4-6 below.

Figure 4-6. M21004 AV<sub>DD</sub>OUT Connected to the VDD of the Downstream Device



If AC coupling is desired or necessary, then the capacitor should be at least 10  $\mu$ F.



### 4.4 Power Supply Description

The device core is designed to operate from a nominal 1.2V supply. However, if a 1.2V supply is not available locally then the internal regulator can be used to create a 1.2V domain from AV<sub>DD</sub>OUT.

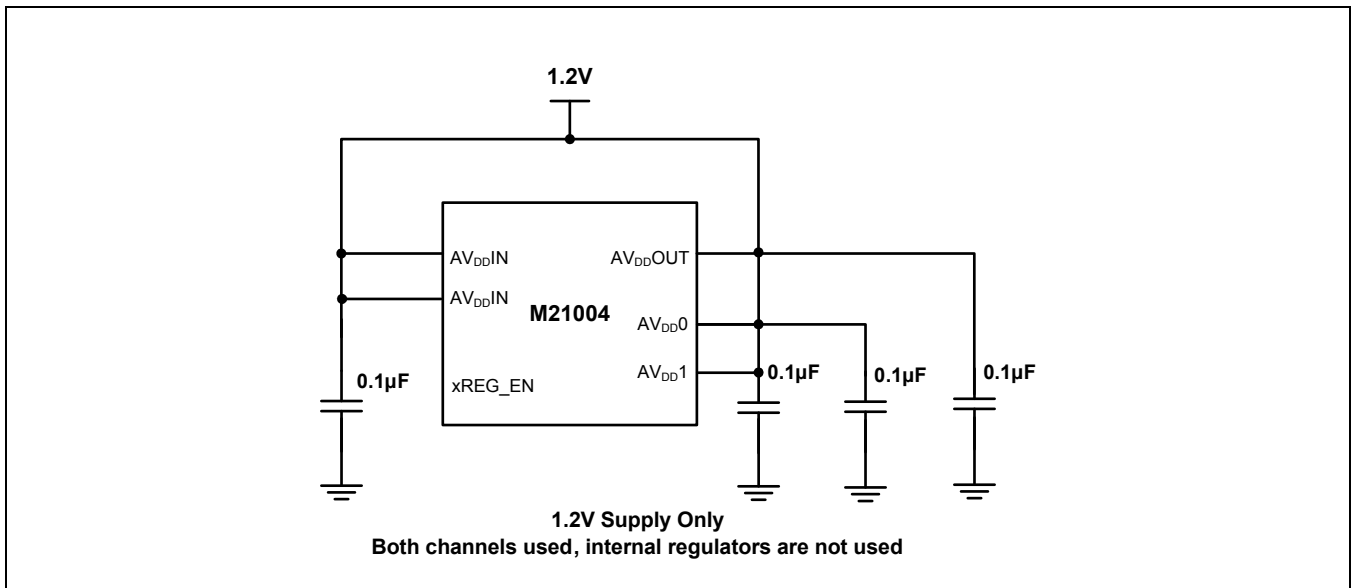
Note that as AV<sub>DD</sub>IN is electrically independent, it must always be supplied with a voltage within the specified range regardless of whether the regulator is enabled or not.

The regulator is controlled through the active low xREG\_EN pin. Setting the pin LOW by connecting it to AV<sub>SS</sub> enables the regulator. The xREG\_EN signal is referenced to AV<sub>DD</sub>OUT, so in order to set it HIGH it must be connected to that supply rail. However, the pin features an integrated pull-up resistor, so it may be left floating if the regulator is not used.

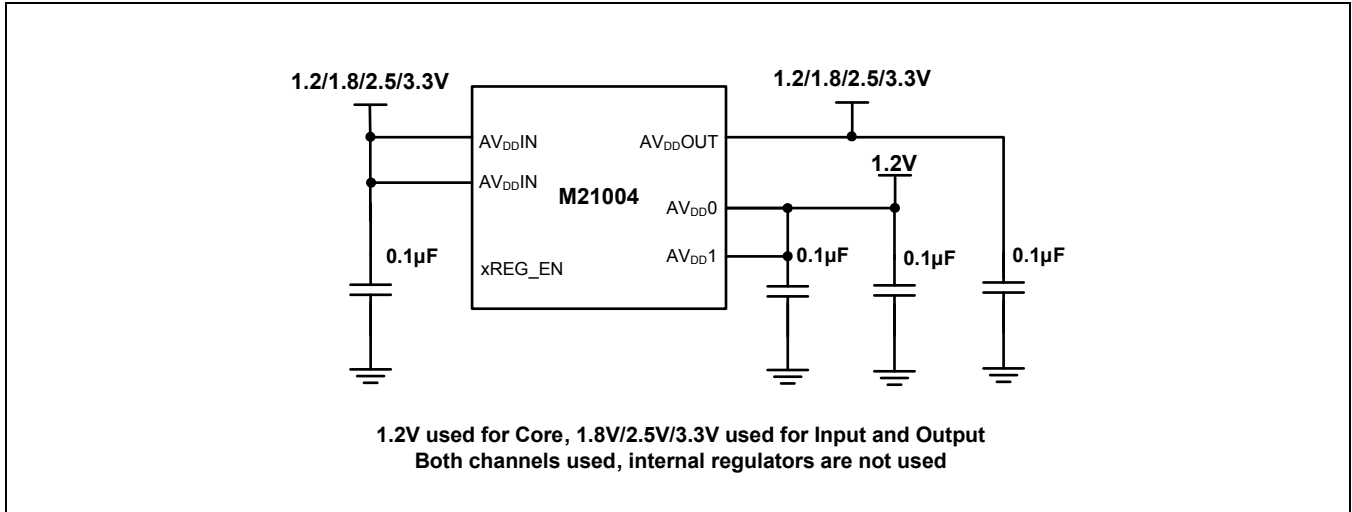
When using the internal regulator the total power consumption will increase, the amount of increase depends on supply voltage used. This occurs because the voltage dropped across the regulator (supply voltage - 1.8) is dissipated within the M21004.

Figure 4-7 to Figure 4-10 illustrate the connection for four different supply configurations. Note that the decoupling capacitors must be 0.1 μF or greater.

**Figure 4-7. Supply Configuration Example #1**



**Figure 4-8. Supply Configuration Example #2**



**Figure 4-9. Supply Configuration Example #3**

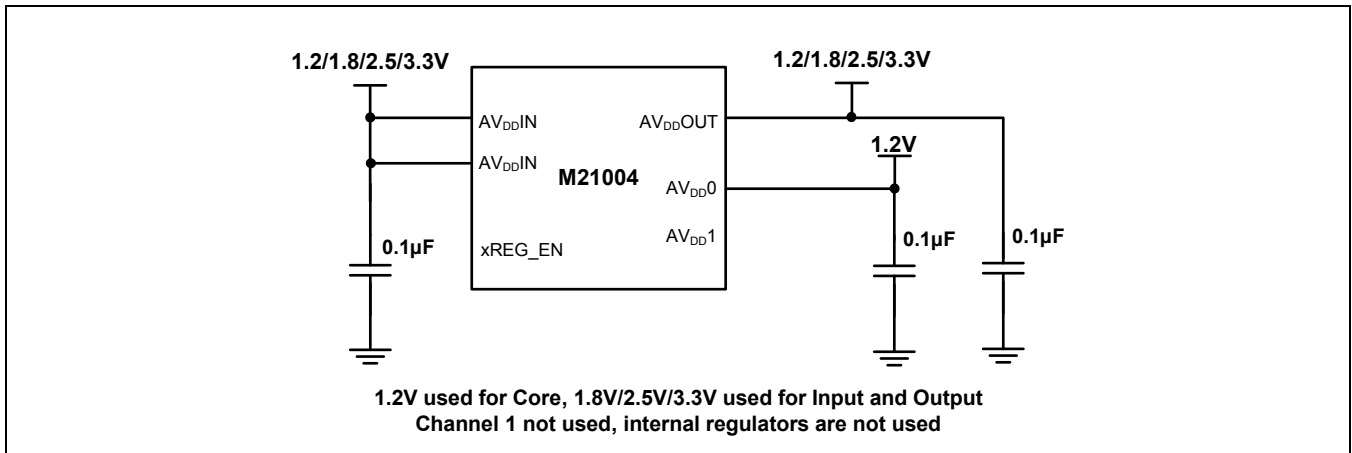
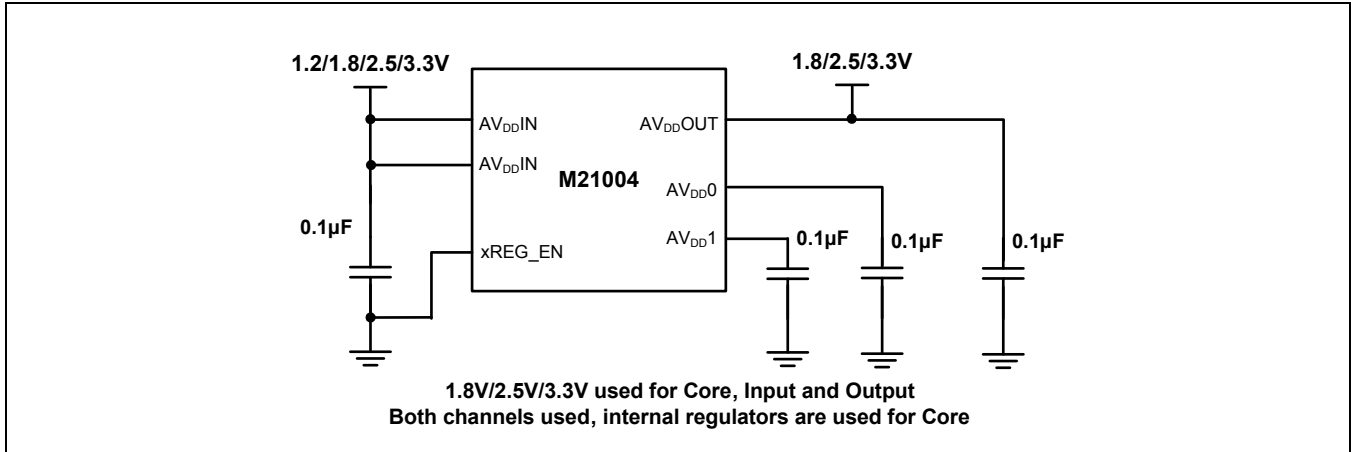


Figure 4-10. Supply Configuration Example #4



## 4.5 Power Up Sequence

For most applications the supply power up sequence does not matter. However if the supply to AV<sub>DD</sub>OUT is low impedance when powered down then current can be diverted from the core supply to AV<sub>DD</sub>OUT. To prevent these current surges from AV<sub>DD</sub>0,1 to the output stage, the power up sequence should be as follows:

AV<sub>DD</sub>OUT first followed by AV<sub>DD</sub>IN and AV<sub>DD</sub>0, 1.

## 4.6 Logic Control Signals

The M21004 may be configured through several digital control pins. In order to allow interfacing to logic levels other than the 1.2V core voltage, the digital control signals are referenced to AV<sub>DD</sub>OUT.

Some digital control pins have three states: HIGH (H), LOW (L), or FLOATING (F). In order to assert the F state, the pin must be left unconnected or undriven.

### 4.6.1 Input Equalizer Control

The IE\_CTRL pins in the M21004 set the equalizer level for the corresponding inputs.

Table 4-1. Operation of IE\_CTRL Pins (Input Equalizer)

Pin	Level	Function
IE_CTRL0, IE_CTRL1	L	Input equalization disabled
	F	Medium equalization (default)
	H	High equalization

### 4.6.2 Output De-emphasis Control

The DE\_CTRL pins in the M21004 set the de-emphasis level for the corresponding outputs.

**Table 4-2. Operation of DE\_CTRL Pins (De-emphasis)**

Pin	Level	Function
DE_CTRL0, DE_CTRL1	L	De-emphasis disabled
	F	Medium de-emphasis (default)
	H	High de-emphasis

### 4.6.3 Output Swing Control

The SWING\_CTRL pin in the M21004 sets the PCML swing level for the corresponding output.

**Table 4-3. Operation of SWING\_CTRL Pin**

Pin	Level	Function
SWING_CTRL0, SWING_CTRL1	L	Output swing set to low
	F	Output swing set to medium (default)
	H	Output swing set to high

### 4.6.4 LOS/Mute Control

The LOS/Mute pins are dual purpose:

#### LOS Output

If left floating the pin is a loss of signal detect output, when the input signal goes below the LOS assert level the output will go high, when the input signal goes above the de-assert level the output will go low. Note that the impedance of the load attached to LOS/MUTE should be > 50 kΩ to either ground or Vdd, this will prevent false activation of Mute Control.

In this state the M21004 automatically squelches the relevant channels data outputs when LOS is asserted.

#### Mute Control

When LOS\_MUTE is forced high externally, the squelch circuit is overridden and high speed data output for that channel will be forced to mute.

**Table 4-4. Operation of DE\_CTRL pin**

Pin	Level	Function
LOS/MUTE0, LOS/MUTE1	L	Output is never muted
	F	LOS output indicator with Squelch enabled
	H	Output muted

### 4.6.5 XPT Control

The XPT\_CTRL1:0 pins control the internal 2x2 crosspoint switch, this is situated between the input IE block and the output DE block, see [Figure 4-1](#). All four modes; feed-through, crossover and broadcast from input 0 or input 1 are supported as shown in [Table 4-5](#) below:

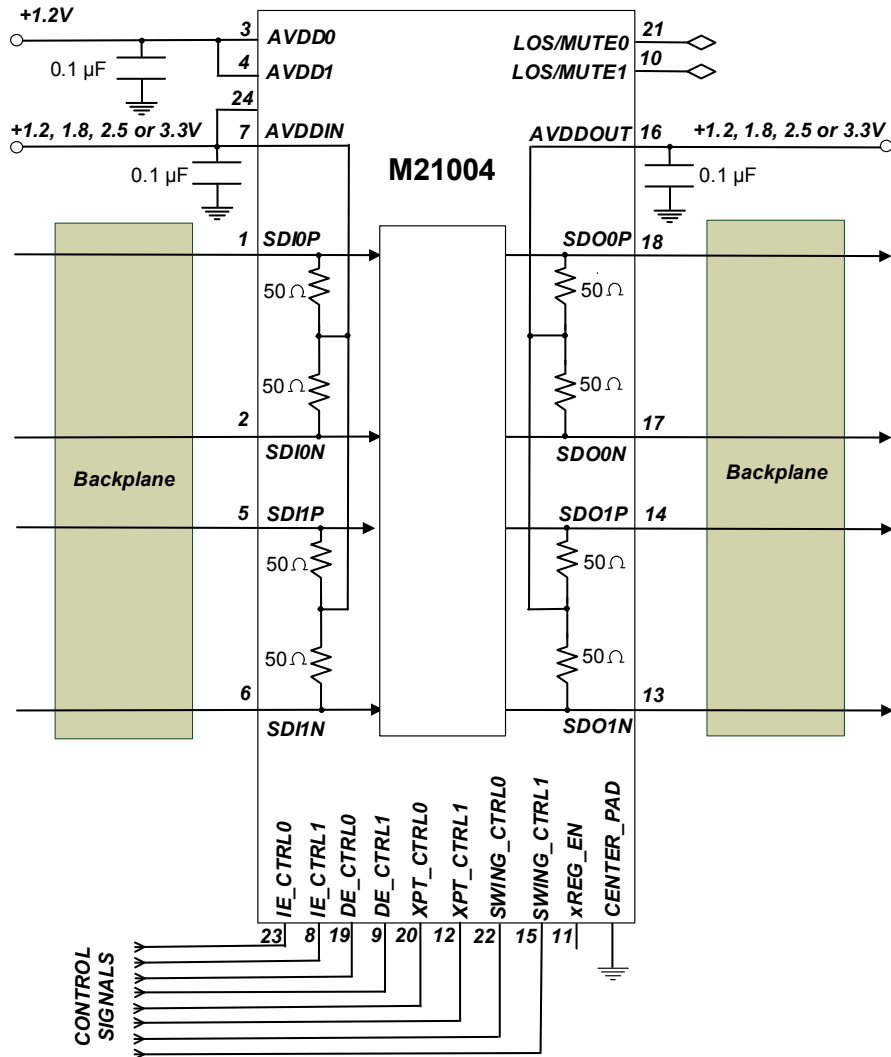
**Table 4-5. XPT Control**

XPT_CTRL1	XPT_CTRL0	Connections	Function
LOW (L)	LOW (L)	SDI0 to SDO0, SDI0 to SDO1	Broadcast 0
LOW (L)	HIGH (H)	SDI1 to SDO0, SDI0 to SDO1	Crossover
HIGH (H)	LOW (L)	SDI0 to SDO0, SDI1 to SDO1 (default)	Feed-through
HIGH (H)	HIGH (H)	SDI1 to SDO0, SDI1 to SDO1	Broadcast 1

### 4.6.6 Regulator Enable

Setting the xREG\_EN pin low enables the internal regulator, see [Section 4.4](#) for description.

### 4.7 Typical Application Circuit



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