

### Applications

- SFP and SFF Modules
- 1G/2G/4G Fibre Channel modules
- Short reach and Metro SONET/SDH
- CPRI: 614.4, 1228.8, 2457.6, 3072.0, 4915.2 and 6144.0 Mbit/s
- OLT Laser Driver for GPON and EPON

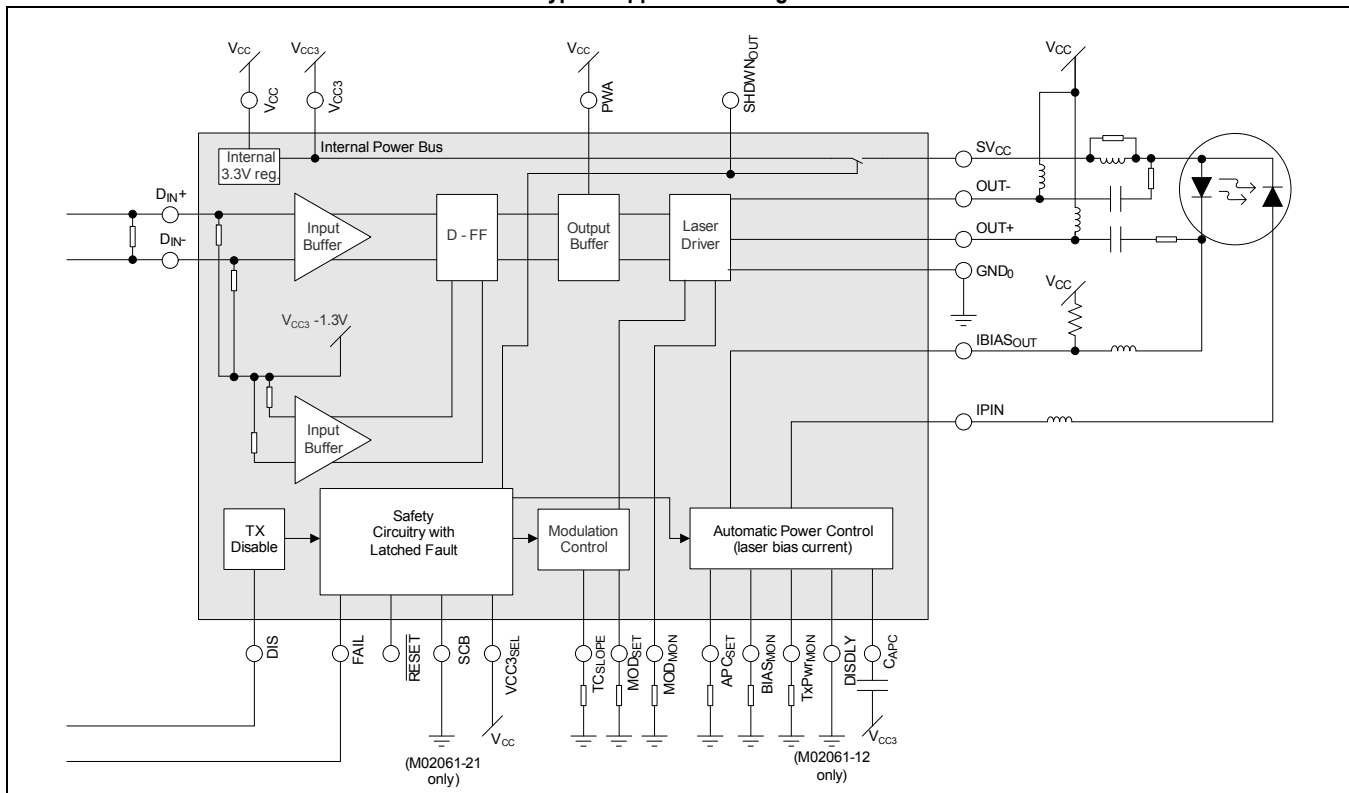
### Features

- High-speed operation; suitable for SFP/SFF applications. Typical rise/fall times of 55 ps.
- Programmable temperature compensation. Modulation output and bias output can be controlled using a few discrete resistors.
- Supports DDMI (SFF-8472) diagnostics
- DC or AC coupled modulation drive. Up to 100 mA modulation current available when AC coupled.
- Low overshoot allows high extinction ratio with low jitter
- Automatic laser power control, with “Slow-Start”
- PECL and CML compatible differential data inputs
- Complies with major MSAs (GBIC, SFF, SFF-8472, SFP) including timing requirements
- Packaged in a QFN24
- 3.3 V or 5 V operation
- Pulse width adjustment

The M02061 is a highly integrated, programmable laser driver intended for SFP/SFF modules. Using differential PECL data inputs, the M02061 supplies the bias and modulation current required to drive an edge-emitting laser. The modulation output can be DC coupled to the laser diode.

The M02061 includes automatic power control to maintain a constant average laser output power over temperature and life. In addition, the modulation current can be temperature compensated to minimize variation in extinction ratio over temperature.

Typical Applications Diagram



1

### Ordering Information

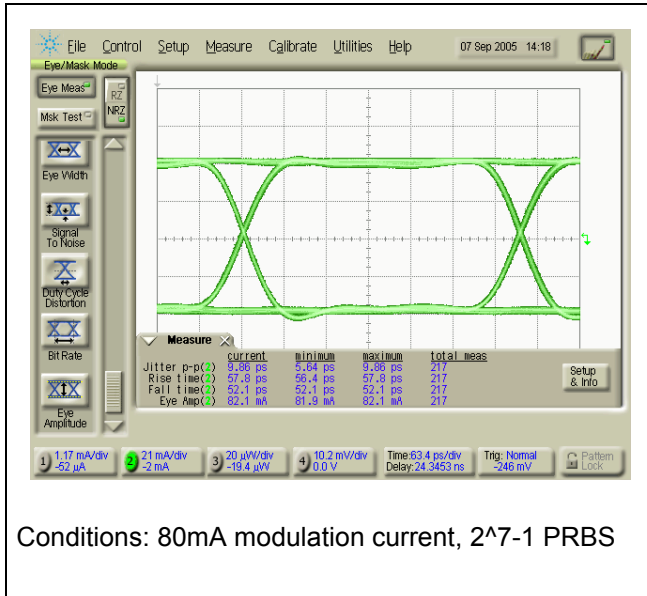
Part Number	Package	Pin Configuration
M02061-12	QFN24	DISDLY function on pin 8
M02061G-12*	QFN24 (RoHS Compliant)	DISDLY function on pin 8
M02061-21	QFN24	SCB function on pin 8
M02061G-21*	QFN24 (RoHS Compliant)	SCB function on pin 8
M02061-EVM	Combination Optical and Electrical Evaluation board	DISDLY function on pin 8

\*The G in the part number indicates that this is an RoHS compliant package.

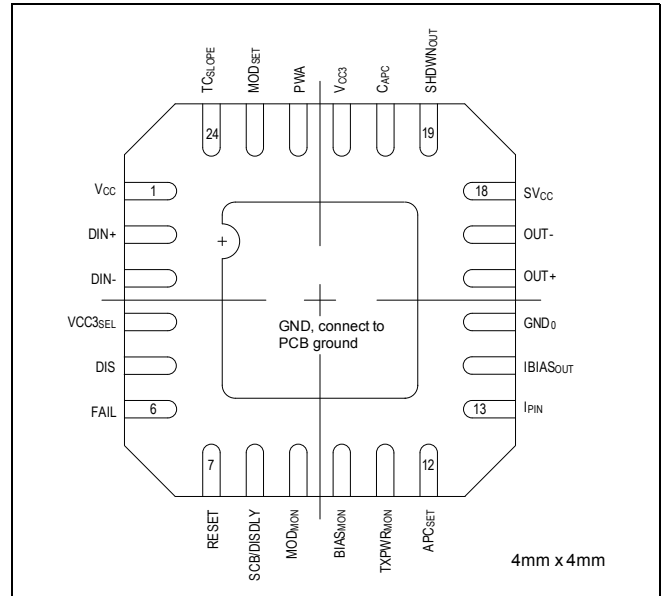
### Revision History

Revision	Level	Date	Description
V6	Release	May 2015	Updated logos and page layout. No content changes.
V5	Release	November 2014	Add <a href="#">Figure 4-2</a> . Recommended component values for AC coupling Laser. Change data sheet marking and numbering to the MACOM system.
G (V4)	Release	July 2012	Added recommendation for 100 k $\Omega$ pull-up resistor at pin 14 IBIASout when pin 18 SVCC is used to control laser current.
F (V3)	Release	February 2010	Added CPRI data rates to front page. Added $T_J$ specification and added equation for $T_A$ specification in <a href="#">Table 1-1</a> .
E (V2)	Release	February 2006	Added information for the M02061-21 with pin SCB instead of pin DISDLY.
D (V1)	Release	September 2005	New format. Remove 32 pin package information. Changes to Absolute Maximum Specifications - operating temperature, output voltage. Changes to Recommended Operating Conditions - VCC, operating temperature. Changes to DC Characteristics - ICC, VMD, TxPWRmon, logic inputs and outputs, data inputs, safety logic thresholds. Changes to AC Characteristics - IMOD, Tr, Tf, jitter.
C	Preliminary	March 2004	Added eye diagram, ; corrected rise/fall times.

2.5Gbps Electrical Eye Diagram



QFN24 Pin Configuration



# 1.0 Product Specification

## 1.1 Absolute Maximum Ratings

**Table 1-1. Absolute Maximum Ratings**

Symbol	Parameter	Rating	Units								
$V_{CC}$	Power supply voltage	-0.4 to +6.0	V								
$T_J$	Junction temperature	-40 to +110 <sup>1</sup>	°C								
$V_{CC3}$	3.3V power supply voltage	-0.4 to +4.0	V								
$T_A$	Operating ambient temperature	-40 to +95 <sup>2</sup>	°C								
$T_{STG}$	Storage temperature	-65 to +150	°C								
$I_{BIASOUT (MAX)}$	Maximum bias output current	150	mA								
$I_{MOD (MAX)}$	Maximum modulation current	140	mA								
$D_{IN+/-}$	Data inputs	0 to $V_{CC3} + 0.4$	V								
DIS, SCB, $V_{CC3SEL}$	Mode control inputs	-0.4 to $V_{CC} + 0.4$	V								
$BIAS_{MON}$ , $MOD_{MON}$	Bias and modulation output current mirror compliance voltage	-0.4 to $V_{CC3} + 0.4$	V								
IPIN	Photodiode anode voltage	-0.4 to $V_{CC3} + 0.4$	V								
IPIN	Photo diode current	2	mA								
FAIL	Status flags	-0.4 to $V_{CC} + 0.4$	V								
PWA, $APC_{SET}$ , $MOD_{SET}$	Set inputs	-0.4 to $V_{CC3} + 0.4$	V								
$TC_{START}$	Temperature compensation start temperature	-0.4 to 1.0	V								
$TC_{SLOPE}$	Temperature compensation slope	-0.4 to $V_{CC3} + 0.4$	V								
OUT+, OUT-	Output	-0.4 to $V_{CC} + 0.4$	V								
<p>1. QFN package:</p> <table> <tr> <td>Air Velocity</td> <td><math>\theta_{JA}</math></td> </tr> <tr> <td>0 m/s</td> <td>57 °C/W</td> </tr> <tr> <td>1 m/s</td> <td>50 °C/W</td> </tr> <tr> <td>2.5 m/s</td> <td>45 °C/W</td> </tr> </table> <p>The above thermal resistance is based on a 4-layer JEDEC standard board (76.2 x 114.3 mm).</p> <p>2. The maximum operating ambient temperature is the lesser of 95 °C or <math>T_A \leq T_J(Max) - (\theta_{JA} (Max) \times Q)</math> where Q is the power dissipated in the M02061.</p>				Air Velocity	$\theta_{JA}$	0 m/s	57 °C/W	1 m/s	50 °C/W	2.5 m/s	45 °C/W
Air Velocity	$\theta_{JA}$										
0 m/s	57 °C/W										
1 m/s	50 °C/W										
2.5 m/s	45 °C/W										

These are the absolute maximum ratings at or beyond which the IC can be expected to fail or be damaged. Reliable operation at these extremes for any length of time is not implied.

## 1.2 Recommended Operating Conditions

**Table 1-2. Recommended Operating Conditions**

Parameter	Rating	Units
Power supply ( $V_{CC-GND}$ )	3.3 $\pm$ 7.5% or 5.0 + 8%, -5%	V
Operating ambient	-40 to +95	$^{\circ}$ C

## 1.3 DC Characteristics

( $V_{CC} = +3.05V$  to  $+3.55V$  or  $4.75V$  to  $5.4V$ ,  $T_A = -40^{\circ}$ C to  $+95^{\circ}$ C, unless otherwise noted)

Typical values are at  $V_{CC} = 3.3V$ ,  $I_{BIASOUT} = 30mA$ ,  $I_{MOD} = 30mA$ ,  $T_A = 25^{\circ}$ C, unless otherwise noted.

**Table 1-3. DC Characteristics (1 of 3)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
$I_{CC}$	Supply current excluding $I_{MOD}$ and $I_{BIAS}$	PWA high (no pulse width adjust)	-	35	61	mA
		additional current when PWA used	-	1.5	-	
		additional current when operating from a 5V supply	-	1.5	-	
$I_{BIAS}$	Bias current adjust range	$V(I_{BIASOUT}) > 0.7V$ For 3.3V operation with an AC coupled laser For 5.0V operation with a DC coupled laser.	1		100	mA
			1		60	
$I_{BIAS(OFF)}$	Bias current with optical output disabled	DIS = high $V(I_{BIASOUT}) > V_{CC} - 1V$	-	-	300	$\mu$ A
	Ratio of $I_{BIAS}$ current to $BIAS_{MON}$ current		-	100	-	A/A
$V_{MD}$	Monitor diode reverse bias voltage	$V_{CC} = 3.3V$	1.5	-	-	V
$I_{MD}$	Monitor diode current adjustment range	Adjusted with $R_{APCSET}$	10	-	1500	$\mu$ A
	Ratio of TxPwr <sub>MON</sub> current to monitor photodiode current		0.95	1	1.25	A/A

**Table 1-3. DC Characteristics (2 of 3)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
C <sub>MD_MAX</sub>	Maximum monitor photodiode capacitance for APC loop stability. Includes all associated parasitic capacitances.				100	pF
	TTL/CMOS input high voltage (DIS)		2.0	–	5.4	V
	TTL/CMOS input low voltage (DIS)		–	–	0.8	V
	CMOS input high voltage (V <sub>CC3SEL</sub> , SCB)			2.4		V
	CMOS input low voltage (V <sub>CC3SEL</sub> , SCB)			1.2		V
	Logic output high voltage (FAIL)	With external 10kΩ pull-up to V <sub>CC</sub> .	V <sub>CC</sub> - 0.5	–	–	V
	Logic output low voltage (FAIL)	For 6.8k to 10k Ω resistor when pulled up to 5V. For 4.7k to 10k Ω resistor when pulled up to 3.3V.	–	–	0.4	V
R <sub>IN</sub>	Differential input impedance	Data inputs	–	6800	–	Ω
V <sub>SELF</sub>	Self-biased common-mode input voltage		–	V <sub>CC3</sub> - 1.3	–	V
V <sub>INCM</sub>	Common-mode input compliance voltage	Data inputs	V <sub>CC3</sub> - 1.45	–	V <sub>CC3</sub> - [V <sub>IN(DIFF)</sub> ]/4	V
V <sub>IN(DIFF)</sub>	Differential input voltage	= 2 x (D <sub>IN+HIGH</sub> - D <sub>IN+LOW</sub> )	200	–	2400	mVpp
V <sub>CC3THL</sub> <sup>(1)</sup>	3.3V supply detection, lower threshold		2.5	2.8	3.0	V
V <sub>CC3THH</sub> <sup>(1)</sup>	3.3V supply detection, upper threshold		3.65	3.9	4.25	V
V <sub>CC5THL</sub>	5V supply detection, lower threshold		3.9	4.25	4.65	V
V <sub>CC5THH</sub>	5V supply detection, upper threshold		5.4	5.8	6.1	V
V <sub>REF1</sub>	Reference voltage for MOD <sub>SET</sub>		1.18	1.3	1.4	V
V <sub>APCSET</sub>	Reference voltage for APC <sub>SET</sub>			1.3		V
V <sub>BL</sub>	Bias_OK lower voltage threshold		0.88	1.0	1.05	V

**Table 1-3. DC Characteristics (3 of 3)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
$V_{BH}$	Bias_OK upper voltage threshold		1.45	1.6	1.7	V
$V_{FAULTL}$	Lower voltage threshold for fault inputs I <sub>BIASOUT</sub> , OUT+, C <sub>APC</sub> , AND MOD <sub>SET</sub>	FAIL asserts if any of these signals fall below this value.		300	400	mV
$V_{OUT\_DIS}$	Self bias voltage for I <sub>BIASOUT</sub> and OUT+	DIS = high	0.5	1.65	2.2	V
$V_{SHDWNL}$	SHDWN <sub>OUT</sub> output low voltage	DIS = low, I <sub>SHDWNOUT</sub> ≤ 100uA			$V_{CC} - 4$	V
$V_{SHDWNH}$	SHDWN <sub>OUT</sub> output high voltage	DIS = low, I <sub>SHDWNOUT</sub> ≤ 10uA	$V_{CC} - 0.3V$			V

**NOTES:**

- When  $V_{CC} = 5V$ ,  $V_{CC3}$  "supply OK" circuitry monitors the internally regulated 3.3V supply. When  $V_{CC} = 3.3V$ ,  $V_{CC3}$  "supply OK" circuitry monitors  $V_{CC}$ .

## 1.4 AC Characteristics

( $V_{CC} = 3.05V$  to  $3.55V$  or  $4.75V$  to  $5.4V$ ,  $T_A = -40\text{ }^\circ\text{C}$  to  $+95\text{ }^\circ\text{C}$ , unless otherwise noted)

Typical values are at  $V_{CC} = 3.3V$ , I<sub>BIASOUT</sub> = 30 mA, I<sub>MOD</sub> = 30 mA, 25 ohm load and  $T_A = 25\text{ }^\circ\text{C}$ , unless otherwise noted.

**Table 1-4. AC Characteristics (1 of 2)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
I <sub>MOD</sub>	Modulation current range	3.3V operation, AC coupled, OUT+ and OUT- ≥ 1.6V  5V operation, DC coupled <sup>(1)</sup> into a 25Ω load to $V_{CC} - 1.2V$ . OUT+ and OUT- ≥ 1.15V	10	–	100	mA
I <sub>MOD(OFF)</sub>	Modulation current with output disabled	DIS = high	–	–	300	μA
	Ratio of modulation current to MOD <sub>MON</sub> current		–	100	–	A/A
I <sub>MOD-TC</sub>	Programmable range for modulation current temperature coefficient	Adjustable using TC <sub>SLOPE</sub> <sup>(2)</sup>	0	–	10 <sup>4</sup>	ppm/°C

**Table 1-4. AC Characteristics (2 of 2)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
tr	Modulation output rise time (2)	20% to 80% into 25 Ω. Measured using 11110000 pattern at 2.5Gbps	–	55	75	ps
tf	Modulation output fall time		–	55	75	ps
OS	Overshoot of modulation output current in the off direction.	into 25 Ω load	--	1	–	%
RJ	Random jitter		–	0.8	–	ps <sub>rms</sub>
DJ	Deterministic jitter	Measured into 25Ω load, 2 <sup>31</sup> - 1 PRBS at 2.7 Gbps  K28.5 pattern at 4.25 Gbps  (includes pulse width distortion <sup>3</sup> )		10  10	25  30	ps <sub>pp</sub>

**NOTES:**

1. Guaranteed by design and characterization.
2. DC coupled operation at 3.3V is not supported. AC coupled operation at 5V is possible provided the outputs never exceed 6V.
3. Pulse width distortion is measured single-ended.

## 1.5 Safety Logic Timing

(SCB pin low, V<sub>CC</sub> = 3.05 V to 3.55V or 4.7V to 5.4V, T<sub>A</sub> = -40 °C to +95 °C, unless otherwise noted)

**Table 1-5. Safety Logic Timing (1 of 2)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
t <sub>off</sub>	DIS assert time	Rising edge of DIS to fall of output signal below 10% of nominal <sup>(1)</sup>			10	μs
t <sub>on</sub>	DIS negate time	Falling edge of DIS to rise of output signal above 90% of nominal <sup>(1)</sup>			1	ms
t <sub>init</sub>	Time to initialize <sup>(2)</sup>	Includes reset of FAIL; from power on after Supply_OK or from negation of DIS during reset of FAIL condition	2	3	5	ms



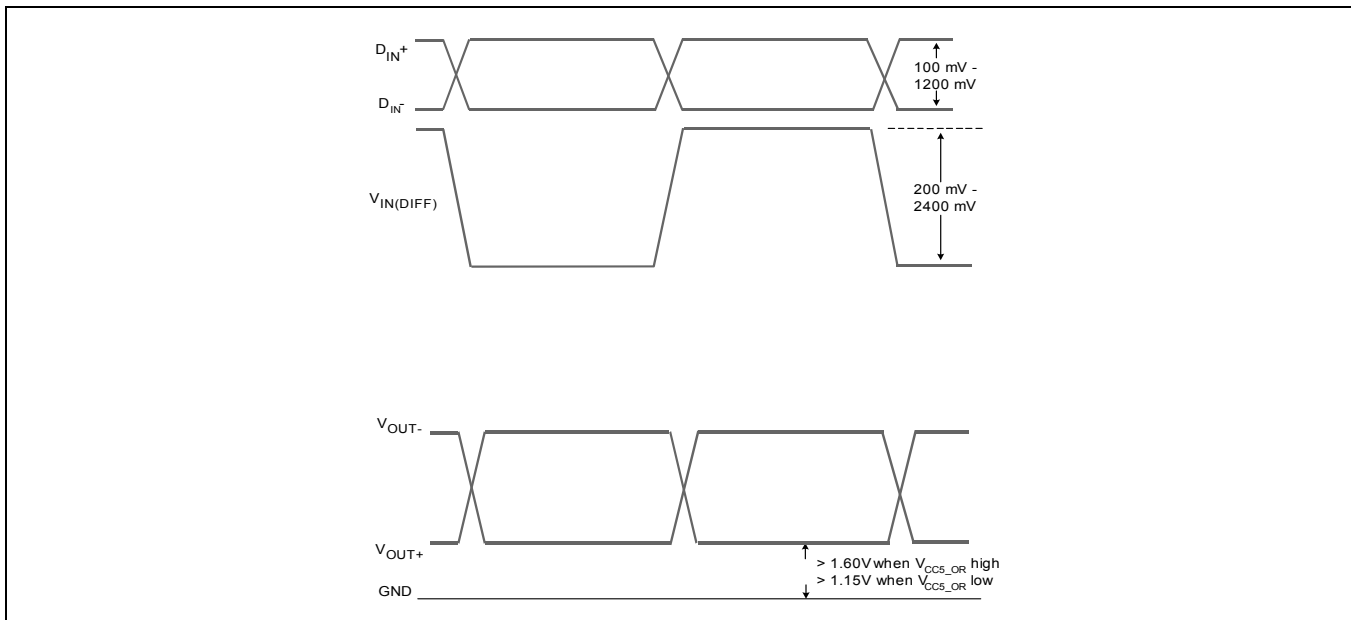
**Table 1-5. Safety Logic Timing (2 of 2)**

Symbol	Parameter	Conditions	Minimum	Typical	Maximum	Units
t <sub>fault</sub>	Laser fault time - from fault condition to assertion of FAIL	From occurrence of fault condition or when Supply_OK is beyond specified range			100	μs
t <sub>reset</sub>	DIS time to start reset	DIS pulse width required to initialize safety circuitry or reset a latched fault			10 <sup>(3)</sup>	μs
t <sub>VCC_OK</sub>	Supply_OK delay time	Delay between Supply_OK condition and when outputs are enabled	10	20		μs
t <sub>onBM</sub>	DIS negate (turn-on) time during burst-mode operation	I <sub>MOD</sub> > 20mA; outputs DC coupled (5V operation) <sup>(4)</sup>		300	500	ns
t <sub>offBM</sub>	DIS assert (turn-off) time during burst-mode operation	I <sub>MOD</sub> > 20mA; outputs DC coupled (5V operation)		200	500	ns

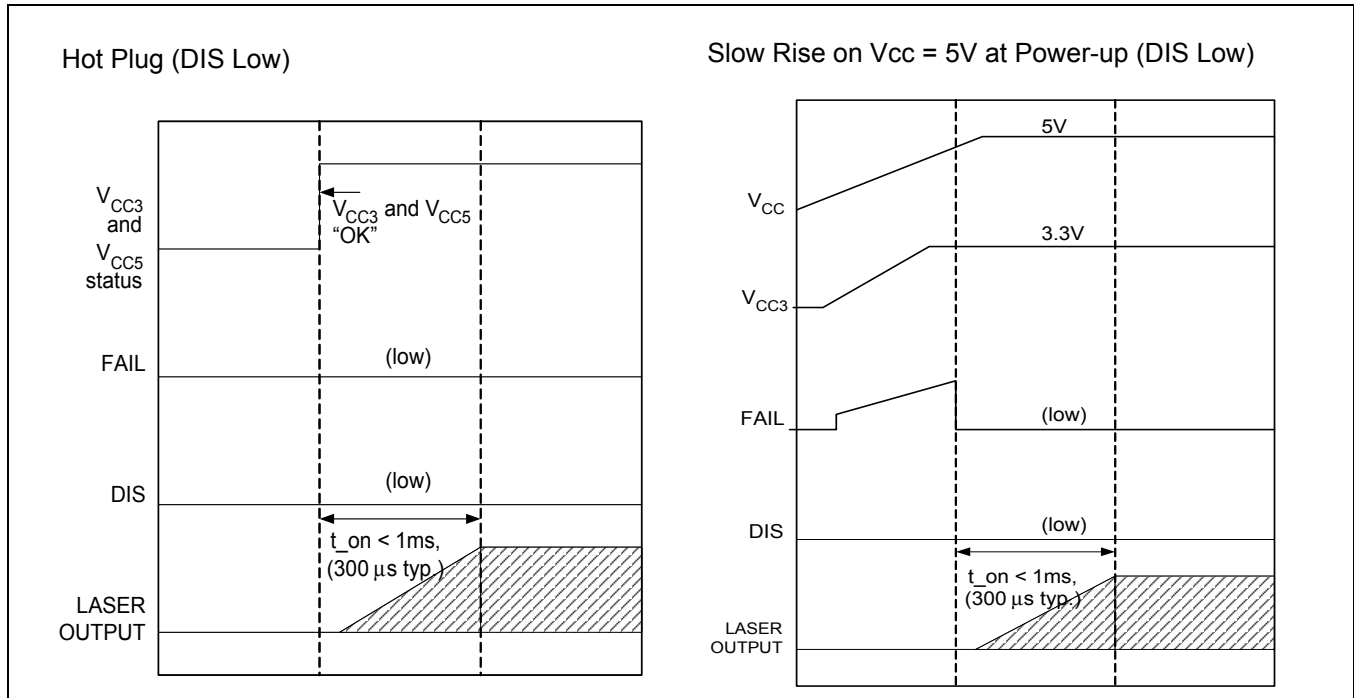
**NOTES:**

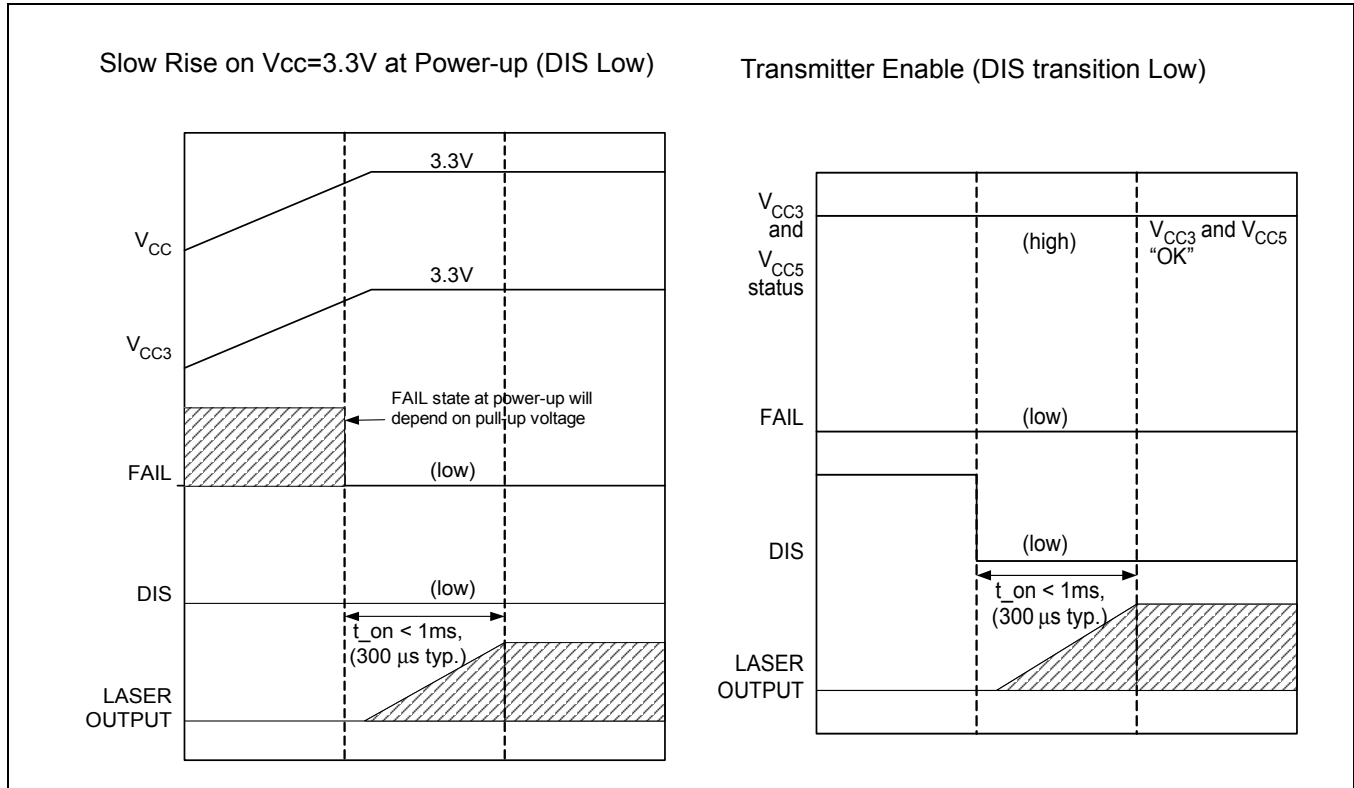
1. With CAPC ≤ 2.2nF
2. User-adjustable. Specifications reflect timing with no external RESET capacitor.
3. With ≤ 1nF capacitor from RESET pin to ground.
4. I<sub>mod</sub> > 12mA

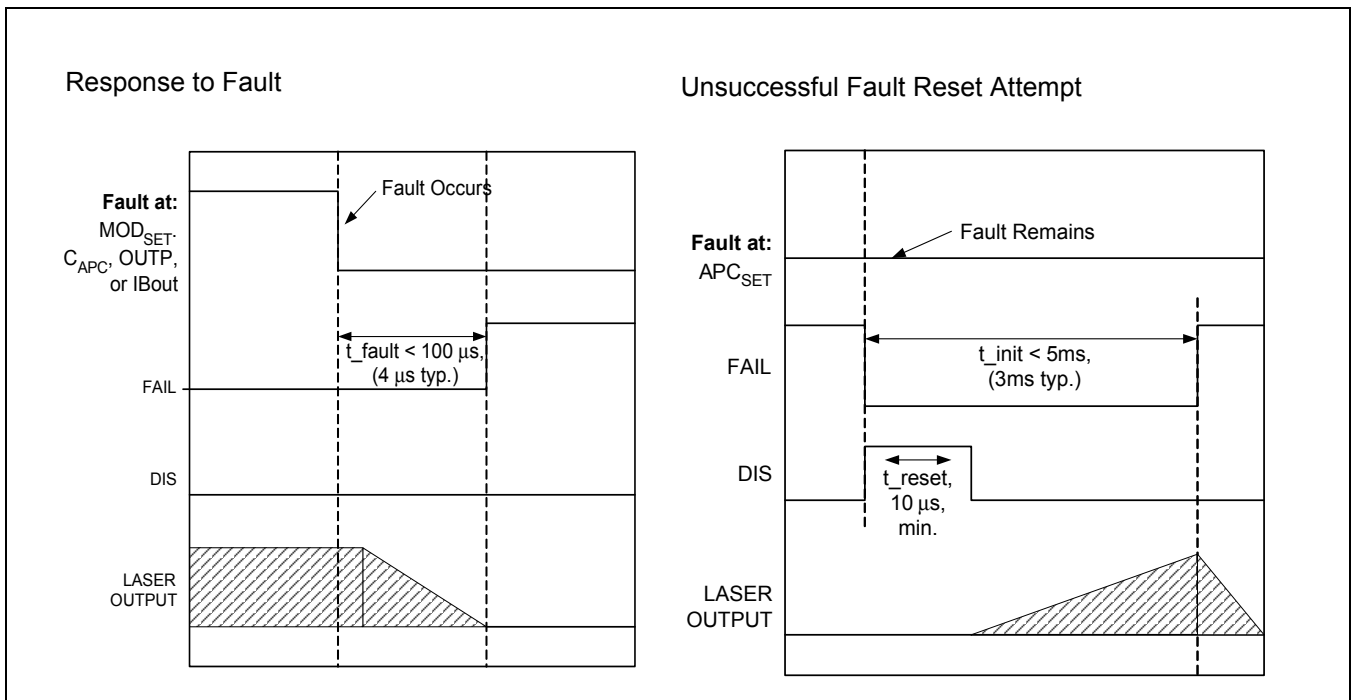
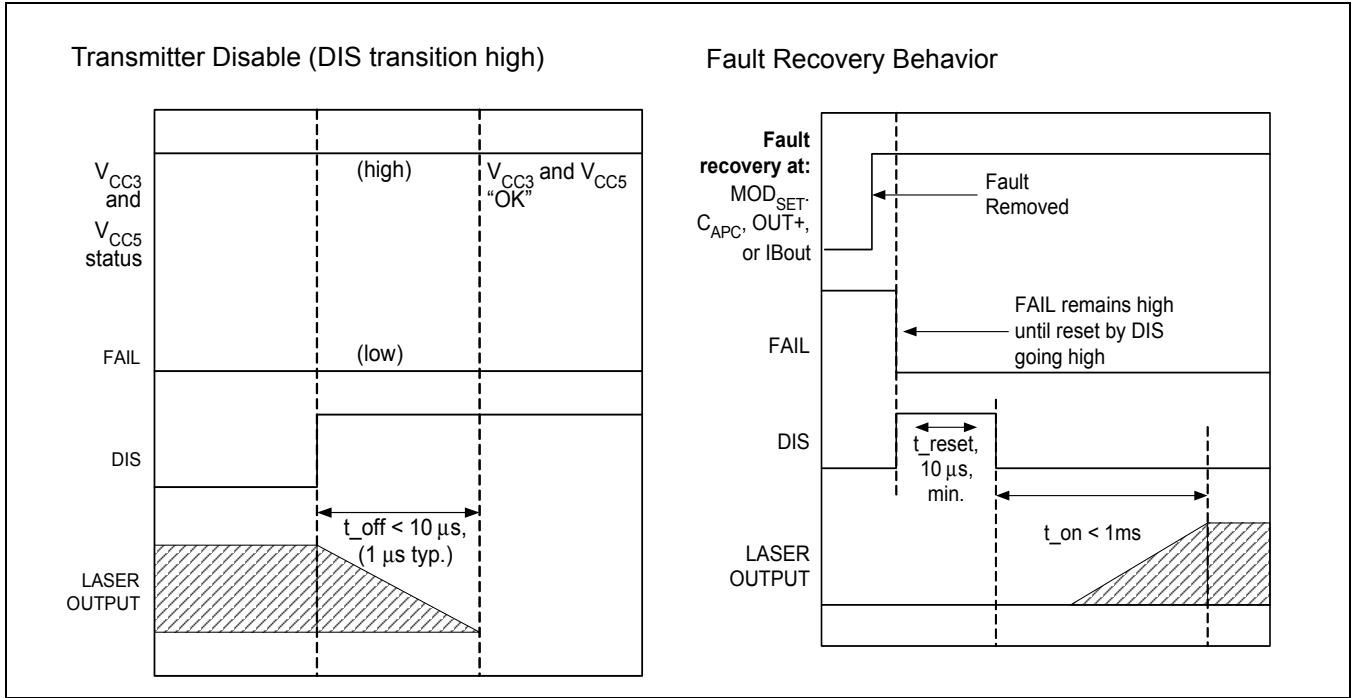
**Figure 1-1. Relationship between Data Inputs and Modulation Outputs**



**Figure 1-2. Safety Logic Timing Characteristics, SCB pin low**







## 2.0 Pin Definitions

Table 2-1 lists pin type definitions and descriptions for the M02061 device.

**Table 2-1. M02061 Pin Definitions and Descriptions (1 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
1	$V_{CC}$		Power supply
2	$D_{IN+}$	<p>The diagram shows the equivalent load for the <math>D_{IN+}</math> pin. It features a diode with its anode connected to <math>V_{CC}</math> and its cathode to the pin. A 4 kΩ resistor is connected between the pin and <math>V_{TT}</math>. Below the pin, there is a transistor circuit where the base is connected to the pin, the emitter is connected to ground, and the collector is connected to <math>V_{CC3}</math>. A current source symbol with a downward arrow is also connected to the collector and ground.</p>	<p>Positive data input. Self biased. Compatible with AC coupled PECL, AC coupled CML, and DC-coupled PECL (<math>V_{CC} = 3.3V</math>).</p> <p>When <math>D_{IN+}</math> is high, <math>OUT+</math> sinks current.</p>
3	$D_{IN-}$	See $D_{IN+}$ drawing	Negative data input. Self biased. Compatible with AC coupled PECL, AC coupled CML, and DC-coupled PECL ( $V_{CC} = 3.3V$ ).

**Table 2-1. M02061 Pin Definitions and Descriptions (2 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
4	VCC3 <sub>SEL</sub>		<p>3.3V V<sub>CC</sub> Select.            Connect to V<sub>CC3</sub> for V<sub>CC</sub> = 3.3V operation.            Connect to GND for V<sub>CC</sub> = 5V operation.</p>
5	DIS		<p>Bias and modulation output disable (TTL/CMOS).</p>

**Table 2-1. M02061 Pin Definitions and Descriptions (3 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
6	FAIL		<p>Safety circuit control failure output (TTL/CMOS). Goes high when a safety logic fault is detected. This output will be low when DIS is high.</p>
7	RESET		<p>Safety circuit reset. Leave open for normal operation or add a capacitor to ground to extend the reset time. Connect to GND to disable window comparators at APC<sub>SET</sub></p>

**Table 2-1. M02061 Pin Definitions and Descriptions (4 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
8 (M02061-21 only)	SCB		Safety Circuit Bypass. Connect to GND or leave open for normal operation. Connect to VCC to allow the bias and modulation outputs to operate even if the safety circuitry indicates a fault.
8 (M02061-12 only)	DISDLY		Disable delay control. Connect to ground for normal operation. In burst mode operation add a capacitor from this pin to ground to set the maximum disable time. Disable times greater than this maximum will engage the "slow-start" circuitry.



**Table 2-1. M02061 Pin Definitions and Descriptions (5 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
9	MOD <sub>MON</sub>	<p>The diagram shows a current mirror circuit. The MOD<sub>MON</sub> pin is connected to the gates of two NMOS transistors. The top NMOS transistor has its source connected to V<sub>CC</sub> and its drain connected to the MOD<sub>MON</sub> pin. The bottom NMOS transistor has its source connected to ground and its drain connected to the MOD<sub>MON</sub> pin. A 190 Ω resistor is connected between the drains of the two transistors. The gates of both transistors are connected to the MOD<sub>MON</sub> pin. The gates of the two PMOS transistors are connected to V<sub>CC3</sub>.</p>	<p>Modulation Current Monitor. Connect directly through a resistor to GND (MON<sub>POL</sub> high) or to V<sub>CC3</sub> (MON<sub>POL</sub> low). The current through this pin is approximately 1/100th of the MODULATION current to the laser.</p> <p>This pin may be left open if the feature is not needed and the M02061 current consumption will be reduced by 0.5mA typically.</p>
10	BIAS <sub>MON</sub>	See MOD <sub>MON</sub> drawing	<p>Bias Current Monitor. Connect directly through a resistor to GND (MON<sub>POL</sub> high) or to V<sub>CC3</sub> (MON<sub>POL</sub> low). The current through this pin is approximately 1/100th of the BIAS current to the laser.</p> <p>This pin may be left open if the feature is not needed and the M02061 current consumption will be reduced by 0.5mA typically.</p>
11	TxPwr <sub>MON</sub>	<p>The diagram shows a current mirror circuit. The TxPwr<sub>MON</sub> pin is connected to the gates of two NMOS transistors. The top NMOS transistor has its source connected to V<sub>CC</sub> and its drain connected to the TxPwr<sub>MON</sub> pin. The bottom NMOS transistor has its source connected to ground and its drain connected to the TxPwr<sub>MON</sub> pin. A 33 Ω resistor is connected between the drains of the two transistors. The gates of both transistors are connected to the TxPwr<sub>MON</sub> pin. The gates of the two PMOS transistors are connected to V<sub>CC3</sub>.</p>	<p>Transmit Power Monitor. Connect directly through a resistor to GND (MON<sub>POL</sub> high) or to V<sub>CC3</sub> (MON<sub>POL</sub> low). The current through this pin is approximately the same as the photo diode current into IPIN.</p> <p>This pin may be left open if the feature is not needed and the M02061 current consumption will be reduced by the IPIN current.</p>

**Table 2-1. M02061 Pin Definitions and Descriptions (6 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
12	APC <sub>SET</sub>		<p>Average Power Control, laser bias current adjustment. Connect a resistor between this pin and ground to set the bias current to the laser.</p> <p>The APC loop will control the laser bias current to maintain a voltage of approximately 1.3V at this pin. The current through this pin is approximately the same as the current into I<sub>PIN</sub>.</p>
13	I <sub>PIN</sub>		<p>Current input from monitor photodiode anode.</p> <p>The APC loop will adjust the laser bias current to maintain a voltage at APC<sub>SET</sub> of approximately 1.3V and at this pin of approximately one V<sub>GS</sub>. The voltage at this pin will not exceed 1.6V in normal operation</p>
14	IBIAS <sub>OUT</sub>		<p>Laser bias current output.</p> <p>Connect directly to laser cathode or at higher bit rates through a ferrite or a resistor to isolate the capacitance of this pin from the modulation drive, (~2pF).</p> <p>Maintain a voltage ≥ 0.7V at this pin.</p> <p>Connect 100 kΩ pullup resistor to VCC if pin 18 (SVCC) is used.</p>

**Table 2-1. M02061 Pin Definitions and Descriptions (7 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
15	GND <sub>0</sub>		Ground for output stage. May be connected directly to ground. At high bit rates (>2Gb/s) an optional inductor or ferrite may be added to reduce switching transients.
16	OUT+		Positive modulation current output. Sinks current when D <sub>IN+</sub> is HIGH. Maintain a voltage $\geq 1.6V$ at this pin when VCC3 <sub>SEL</sub> is high. Maintain a voltage $\geq 1.15V$ at this pin when VCC3 <sub>SEL</sub> is low.
17	OUT-	See OUT+ drawing	Negative modulation current output. Sinks current when D <sub>IN-</sub> is HIGH Maintain a voltage $\geq 1.6V$ at this pin when VCC3 <sub>SEL</sub> is high. Maintain a voltage $\geq 1.15V$ at this pin when VCC3 <sub>SEL</sub> is low.
18	SV <sub>CC</sub>		Switched V <sub>CC</sub> . 3.3V applications - Connect to laser anode. Safety circuitry will open the switch when a fault is detected and no current will flow through the laser. No capacitance is needed on this node. If capacitance to ground is added, do not exceed 100 pF. 5V applications - Disabled, leave open.

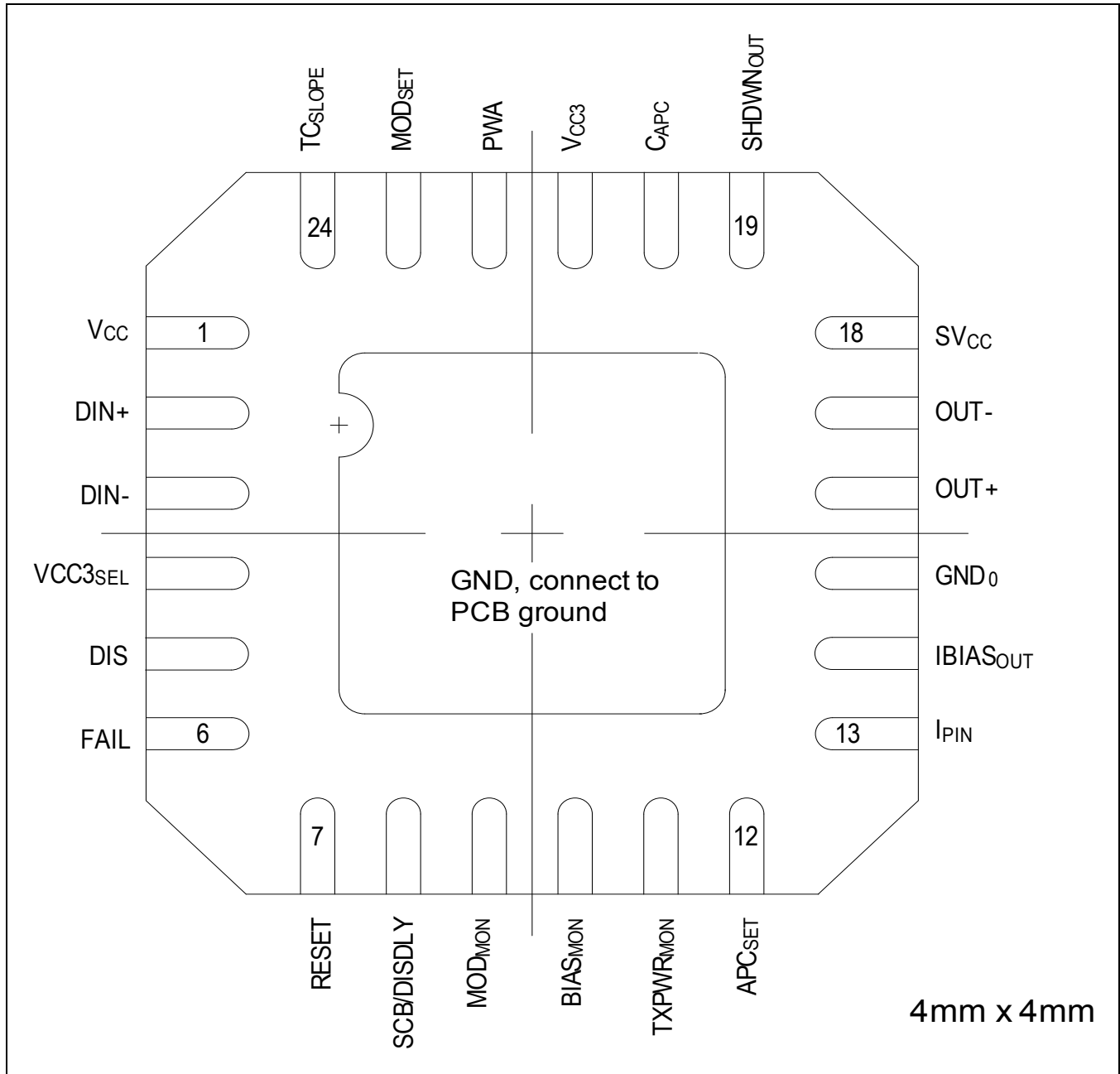
**Table 2-1. M02061 Pin Definitions and Descriptions (8 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
19	SHDWN <sub>OUT</sub>		External switched $V_{CC}$ control signal. Use in 5V applications to create an external $SV_{CC}$ .
20	$C_{APC}$		Automatic power control loop dominant pole capacitor. (Connect a capacitor between this pin and $V_{CC3}$ .) A 2.2 nF capacitor will give less than 1ms enable time and a loop bandwidth < 30kHz.
21	$V_{CC3}$		3.3V applications - Power supply input. Connect to $V_{CC}$ . 5V applications - Internally generated 3.3V. Power supply output. Do not attach to non-M02061 circuitry. For 5 V applications add 12 ohms in series with 100 nF to ground at this pin.

**Table 2-1. M02061 Pin Definitions and Descriptions (9 of 9)**

4x4 mm QFN24 Pin Number	Pin Name	Pin equivalent load	Function
22	PWA	<p>The diagram shows the internal circuit for the PWA pin. It features a 190 Ω resistor connected between the PWA pin and the V<sub>CC</sub> supply. A 1.28V reference voltage source is also connected to the PWA pin. The circuit includes a PMOS transistor controlled by V<sub>CC3</sub> and an NMOS transistor controlled by a 1.28V reference voltage. The PWA pin is also connected to ground through a diode.</p>	Pulse Width Adjust. Connect a resistor to GND to enable, (between 1kΩ and 20kΩ). Connect to V <sub>CC3</sub> to disable.
23	MOD <sub>SET</sub>	See PWA drawing	Modulation current control. Connect a resistor to ground to set the modulation current.
24	TC <sub>SLOPE</sub>	See PWA drawing	Modulation current temperature compensation coefficient adjustment. Connect a resistor to ground to set the temperature compensation coefficient. Leave open to disable the temperature compensation. A 51 kΩ resistor will result in a temperature compensation slope of approximately 0.5%/°C.
CENTER PAD	GND		Connect to GND.

Figure 2-1. QFN24 Pinout Information



## 3.0 Functional Description

---

### 3.1 Overview

The M02061 is a highly integrated, programmable laser driver intended for SFP/SFF module with data rates up to 4.25 Gbps. Using differential PECL data inputs, the M02061 supplies the bias and modulation current required to drive an edge-emitting laser.

Monitor outputs and internal safety logic in the M02061 combined with the M02088 will support designs requiring DDMI compliance.

The M02061 includes automatic power control to maintain a constant average laser output power over temperature and life. In addition, the modulation current can be temperature compensated to minimize variation in extinction ratio over temperature.

Many features are user-adjustable, including the APC (automatic power control) loop bias control (via a monitor photo diode), modulation current, temperature compensation control of modulation current, and pulse-width adjustment. The part may be operated from a 3.3V or 5V supply.

The driver modulation output can be AC, DC, or Differentially coupled to the laser.

Safety circuitry is also included to provide a latched shut-down of laser bias and modulation current if a fault condition occurs. An internal  $V_{CC}$  switch provides redundant shutdown when operating the device from a 3.3V supply. Control is provided to allow for a redundant external switch when operating with a 5V supply, if desired.

[Figure 3-1](#) details the functional blocks and pin signals for the M02061 device.

Figure 3-1. M02061 Block Diagram

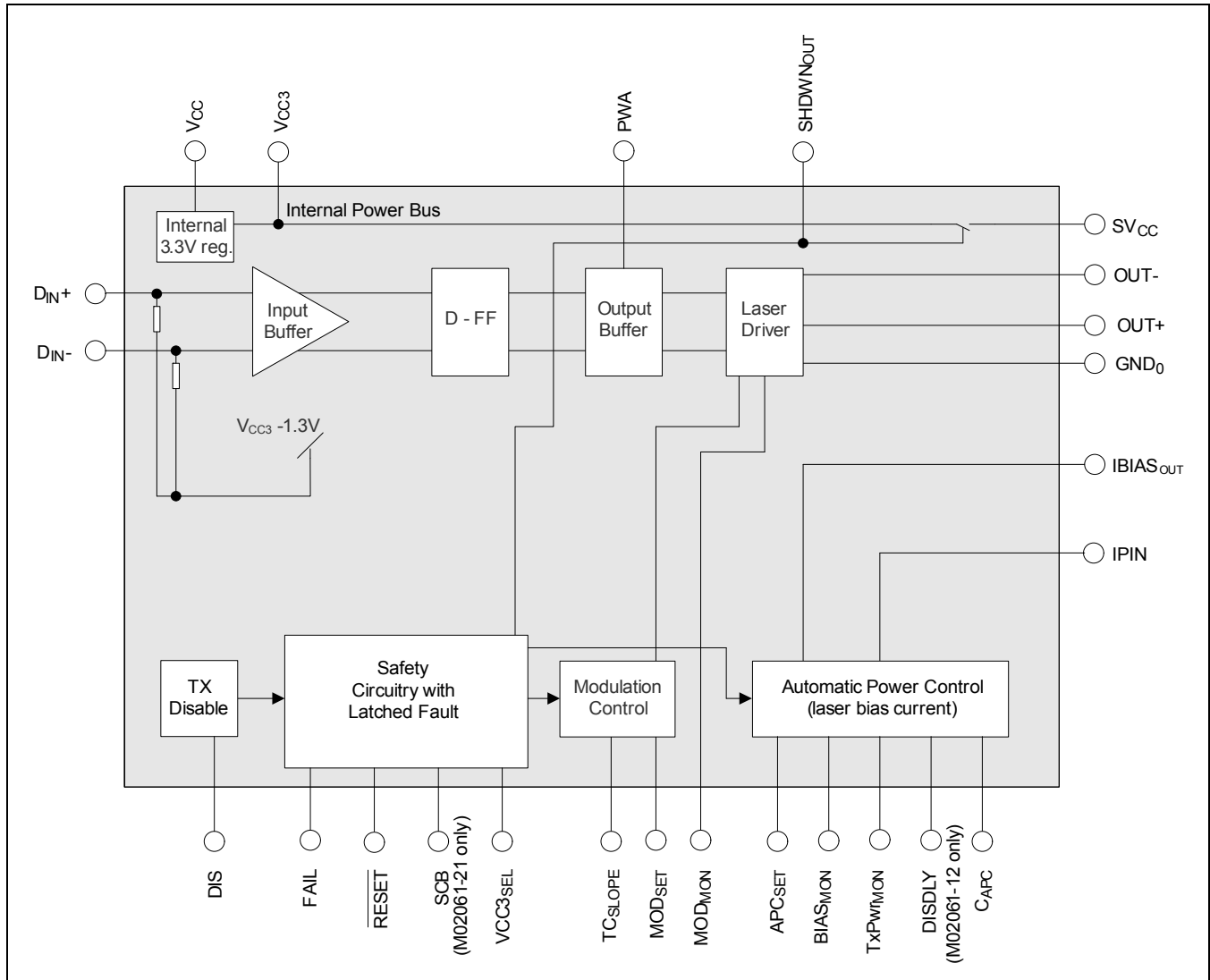
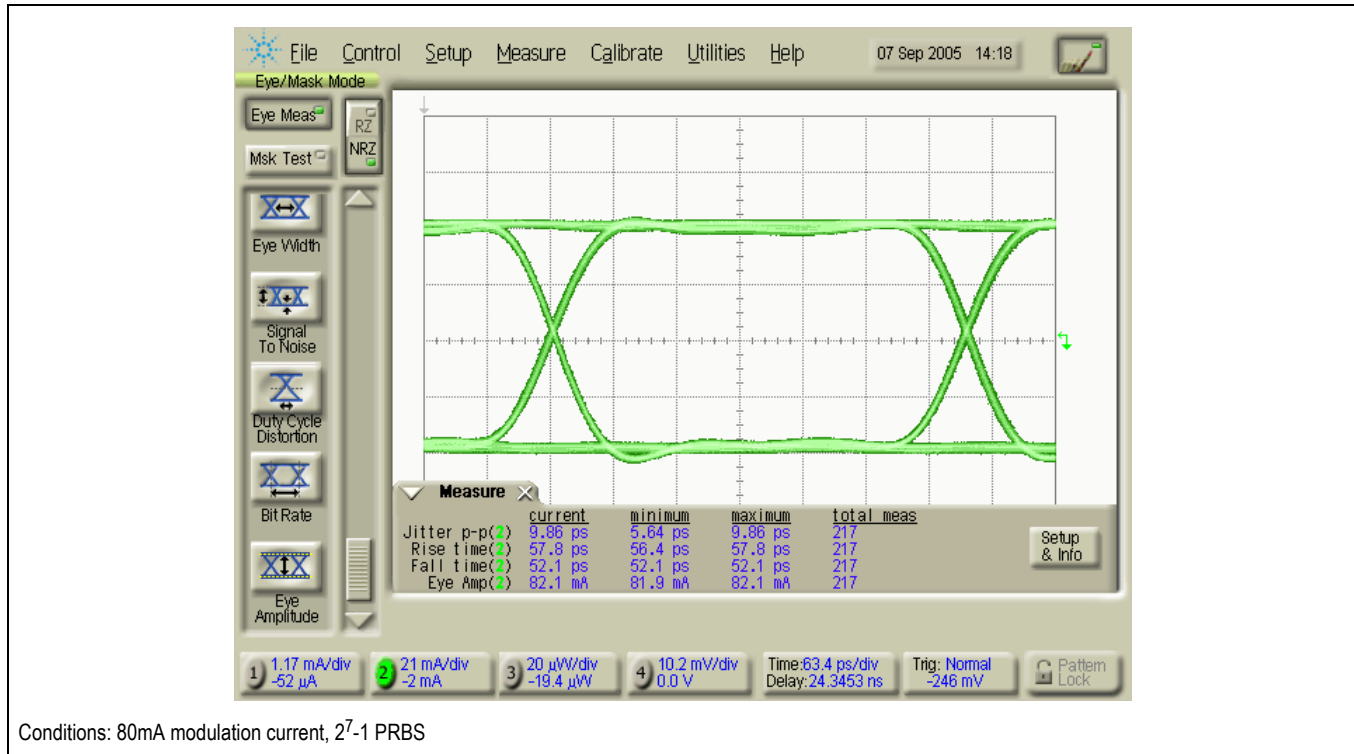




Figure 3-2. 2.5Gbps Electrical Eye Diagram



**Figure 3-3. 2.5Gbps Filtered Optical Eye Diagram with NEC NX7315UA Laser**

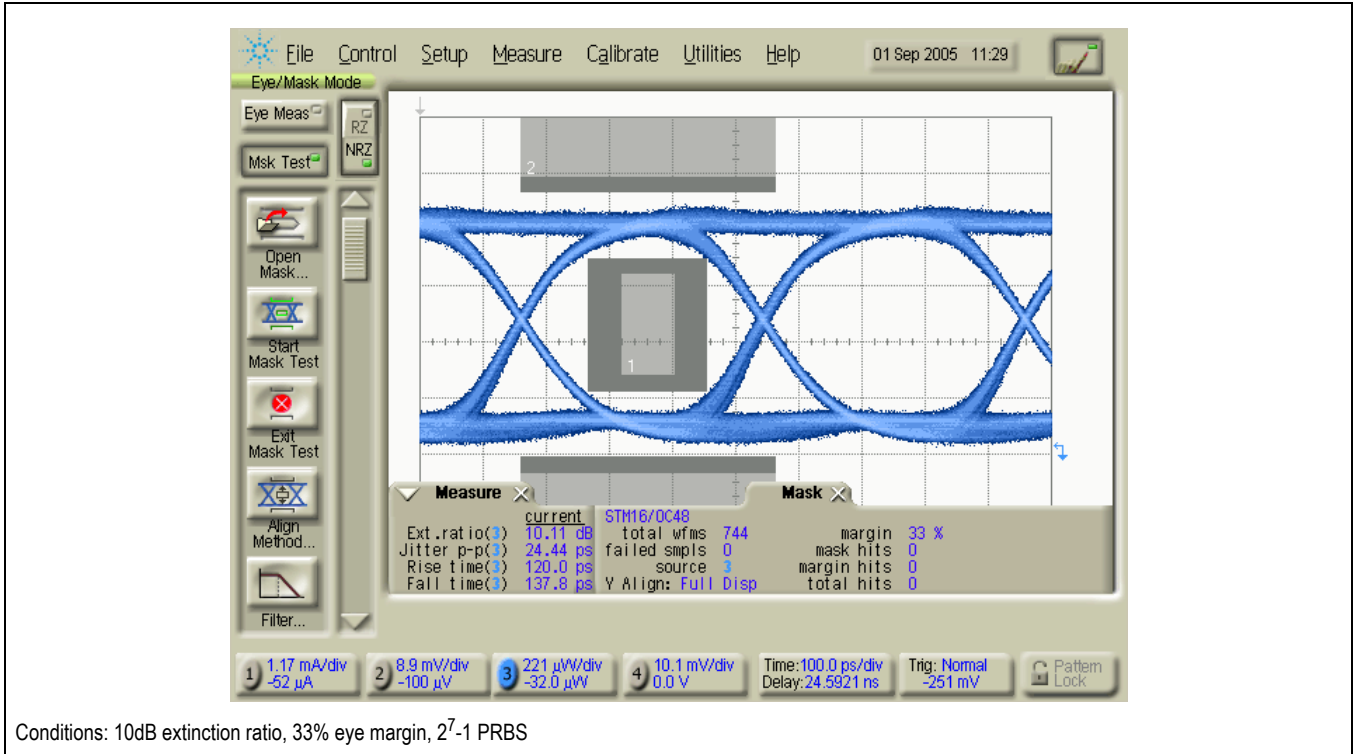
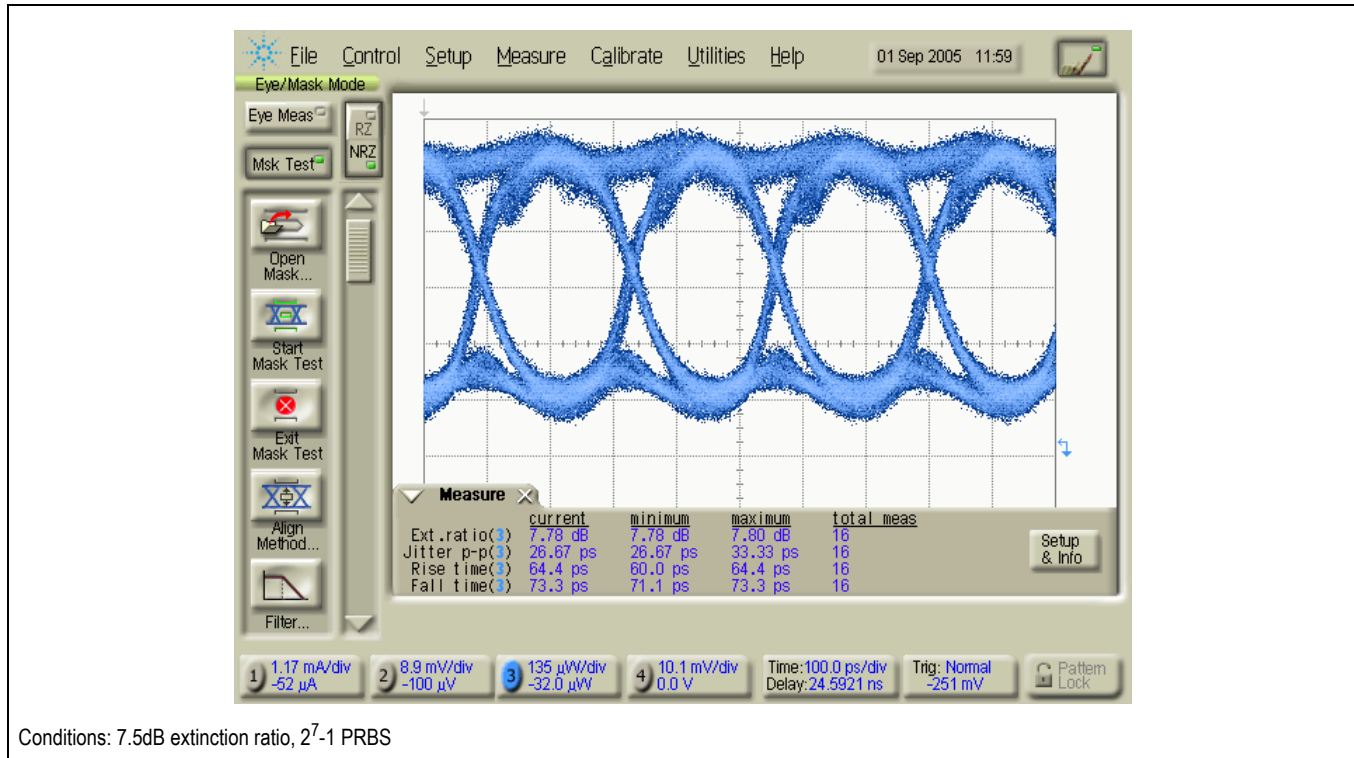


Figure 3-4. 4.25Gbps Unfiltered Optical Eye Diagram with Archcom AC3460 Laser



## 3.2 Features

- High speed operation; suitable for SFP/SFF applications from 155Mbps to 4.25 Gbps. Typical rise/fall times of 55 ps.
- Programmable temperature compensation. Modulation output and bias output can be controlled using a few discrete resistors.
- Supports DDMI (SFF-8472) diagnostics when combined with the M02088.
- DC or AC coupled modulation drive. Up to 100mA modulation current available when AC coupled.
- Low overshoot allows high extinction ratio with low jitter.
- Automatic Laser Power Control, with “Slow-Start”.
- Differential data inputs to minimize pattern dependent jitter, PECL and CML compatible.
- Packaged in a QFN24
- 3.3V or 5V operation

## 3.3 General Description

The M02061 is a highly integrated, programmable laser driver intended for SFP/SFF module with data rates up to 4.25 Gbps. Using differential PECL data inputs, the M02061 supplies the bias and modulation current required to drive an edge-emitting laser. Monitor outputs and internal safety logic support the DDMI requirements.

The M02061 laser driver consists of the following circuitry: an internal regulator, bias current generator and automatic power control, data inputs, buffer with pulse width adjust, modulation current control, modulator output, laser fail indication, disable control, and monitor outputs for the bias current, modulation current, and transmitted power.

### 3.3.1 Internal Regulator

The M02061 contains an internal 3.3V regulator so high bit rate performance can be achieved with 5V or 3.3V power supply.

When operating from a 5V supply ( $V_{CC}$  is connected to +5V), an internal regulator provides a voltage of approximately 3.3V to the majority of the on-chip circuitry. The on-chip regulator is internally compensated, requiring no external components. However, for 5V operation with high modulation currents, it may be necessary to add 12 ohms in series with 100nF to ground at VCC3 or the internal power supply may dip and cause a fault condition. When a 3.3V supply is used ( $V_{CC}$  and  $V_{CC3}$  connected to 3.3V) the regulator is switched off and the internal circuitry is powered directly through the  $V_{CC3}$  supply pin. The decision as to whether or not the internal regulator is required is made via the  $VCC3_{SEL}$  pin, which also determines whether the safety circuitry needs to monitor for proper +5V supply voltage.

For 3.3V applications,  $SV_{CC}$  is sourced from  $V_{CC3}$  through a switch (leave  $SV_{CC}$  open for 5V applications).  $SV_{CC}$  is to be used to power the anode of the laser diode and the cathode of the photo diode, any resistive or ferrite pull-ups on the OUT+ and OUT- outputs should be connected directly to VCC. When a fault condition is present, FAIL will assert and the switch sourcing  $SV_{CC}$  will open so no current can pass through the laser.  $SV_{CC}$  does not need any external capacitance, if capacitance to ground is added at  $SV_{CC}$  it should be  $\leq 100$  pF. When  $SV_{CC}$  is used, add a 100k  $\Omega$  pullup resistor to VCC at pin 14 (IBIASout).

For 5V operation, an analog switch controlled by SHTDWNOUT can be used to source 5V to the laser anode. In the case of a fault condition, SHTDWNOUT will go high and open the analog switch which will result in an open circuit at the laser. SHTDWNOUT is designed to drive a CMOS logic input. An FET transistor may have excessive Miller capacitance and a fault may be signalled if it turns on too slow.

$V_{CC}$  and  $V_{CC3}$  status are internally monitored by the M02061 during power-up and normal operation. During power-up the “slow-start” circuitry requires that  $V_{CC}$  and  $V_{CC3}$  each reach an acceptable level before enabling bias or modulation current.

Table 3-1. Pin Connection for 3.3V and 5V V<sub>CC</sub>

		Pin Connection For:	
		V <sub>CC</sub> = 3.3V	V <sub>CC</sub> = 5V
Pins Dependent on V <sub>CC</sub> Voltage	V <sub>CC3</sub>	Connect to V <sub>CC</sub>	Reference for C <sub>APC</sub> and PWA
	SV <sub>CC</sub>	Laser Anode	OPEN
	SHDWN <sub>OUT</sub>	OPEN	External safety control switch
	C <sub>APC</sub>	Capacitor between C <sub>APC</sub> and V <sub>CC3</sub> or V <sub>CC</sub>	Capacitor between C <sub>APC</sub> and V <sub>CC3</sub> (not V <sub>CC</sub> )
	PWA	Connect to V <sub>CC3</sub> or V <sub>CC</sub> to disable	Connect to V <sub>CC3</sub> to disable (not V <sub>CC</sub> )
	VCC3 <sub>SEL</sub>	Connect to V <sub>CC3</sub> or V <sub>CC</sub>	Connect to GND

### 3.3.2 Bias Current Generator and Automatic Power Control

To maintain constant average optical power, the M02061 incorporates a control loop to compensate for the changes in laser threshold current over temperature and lifetime. The bias current will be determined by the value of the external resistor R<sub>APCSET</sub> and the transfer efficiency between the laser and monitor photo diode.

The photo current from the monitor photo diode mounted in the laser package is sunk at I<sub>PIN</sub>. This photo current is mirrored and an equivalent current is sourced from pins TxPwr<sub>MON</sub> and APC<sub>SET</sub>. The APC loop adjusts the laser bias current (hence the monitor diode photo current) to maintain a voltage at APC<sub>SET</sub> of 1 band-gap voltage or ~1.3V.

$$R_{APCSET} * I_{PIN} = 1.3 \text{ V}$$

The APC loop has a time constant determined by C<sub>APC</sub>, R<sub>APCSET</sub> and the transfer efficiency between the laser and monitor photo diode. The larger the C<sub>APC</sub> capacitor the lower the bandwidth of the loop and the larger R<sub>APCSET</sub> the lower the loop BW.

In general, it is recommended that at least 2.2 nF of external capacitance be added externally between C<sub>APC</sub> and V<sub>CC3</sub>. With use of a 2.2 nF capacitor, the bias current can reach 90% of its final value within 1ms, i.e., bias current rise-time is less than 1ms and the APC loop bandwidth is less than 30 kHz, which should be adequate for bit rates of 155Mbps. (and all higher bit rates).

The bias generator also includes a bias current monitor mirror (BIAS<sub>MON</sub>), whose output current is typically 1/100th of the bias current. This pin can be connected directly through a resistor to ground. If this function is not needed this pin can be left open.

### 3.3.3 Data Inputs

Both CML and PECL inputs signals can be AC coupled to the M02061. These inputs are internally biased to approximately V<sub>CC3</sub> - 1.3V. In most applications the data inputs are AC coupled with controlled impedance pcb traces which will need to be terminated externally with a 100Ω or 150Ω resistor between the + and - inputs.

PECL and CML signals may be DC coupled to the M02061 data inputs when both the M02061 and the source of the input signals are operating from 3.3V supplies. If the M02061 is operating from a 5V supply, PECL and CML signals may be DC coupled as long as the source of the input signals is operating at a 3.3V supply and the signals are referenced to VCC3 at the M02061.

### 3.3.4 Pulse Width Adjust

The data output buffer incorporates pulse-width adjustment control to compensate for laser pulse width distortion. A potentiometer can be connected between the PWA input and GND for adjustment (programming resistance should be between 1kΩ and 20kΩ). By adjusting the potentiometer, the pulse-width can be adjusted over a range of approximately ±40 ps. Pulse width control can be disabled by connecting PWA to VCC3, resulting in roughly a 50% crossing point at the output and reducing supply current by approximately 1.5mA.

### 3.3.5 Modulation Control

There are programmable control lines for controlling the modulation current and its temperature compensation. These inputs can be programmed simply with a resistor to ground.

The modulation current amplitude is controlled by the MOD<sub>SET</sub> input pin. The modulation current is temperature compensated by the TC<sub>SLOPE</sub> inputs. The temperature compensation is independent of the setting.

If the temperature compensation at TCSLOPE is disabled, the modulation output current is simply:

$$I_{OUT} = 100 \times (1.3V / R_{MODSET})$$

Where R<sub>MODSET</sub> is the resistance from pin MOD<sub>SET</sub> to ground.

Figure 3-5 is the most accurate method for selecting RTCslope.

However, you can also select R<sub>TCSLOPE</sub> using the following relationship:

$R_{TCSLOPE} = 19.5 \times (TC)^{-1.5}$ , where TC is the desired slope of the modulation current from 25°C to 85°C in%/°C and R<sub>TCSLOPE</sub> is in kΩ. If no temperature compensation is desired, leave R<sub>TCSLOPE</sub> open.

In any case, R<sub>TCSLOPE</sub> will have negligible effect at M02061 case temperatures below 10°C.

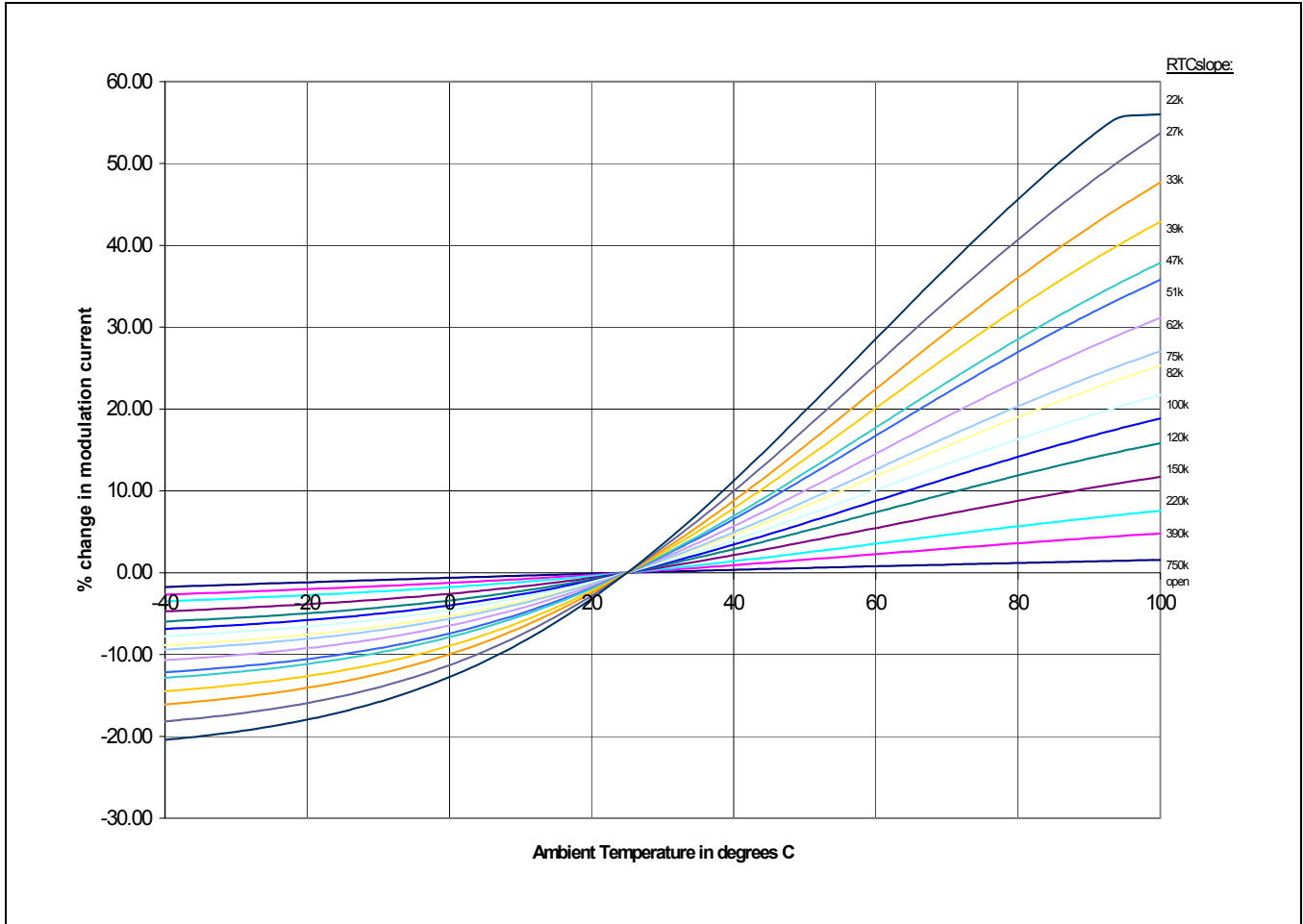
For example:

Given a laser with a desired modulation current at low temperatures of 30mA and a temperature coefficient of -0.5%/°C at high temperatures (which will require a laser driver temperature coefficient of +0.5%).

$$\text{Choose } R_{MODSET} = 100 \times (1.3V / 30mA) = 4.3k\Omega$$

$$\text{Choose } R_{TCSLOPE} = 19.5 \times (0.5)^{-1.5} k\Omega = 56k\Omega$$

**Figure 3-5. TC Slope Compensation Behavior**



### 3.3.6 Modulator Output

The output stage is designed to drive a 25Ω output load over a wide range of currents and circuit architectures. The laser may be AC, DC, or Differentially coupled depending on the supply voltage.

**Table 3-2. Modulation Current Maximums (1 of 2)**

	Max Modulation Current	Max Bias Current
V <sub>CC</sub> =5V, Laser DC coupled	80	60
V <sub>CC</sub> =5V, Laser AC coupled	80 <sup>(1)</sup>	60

**Table 3-2. Modulation Current Maximums (2 of 2)**

	Max Modulation Current	Max Bias Current
$V_{CC}=3.3V$ , Laser DC coupled	100 <sup>(2)</sup>	100
$V_{CC}=3.3V$ , Laser AC coupled	100	100
When differentially coupling, the maximum modulation and bias current is determined by either the AC or DC coupling of the OUT+ or OUT- output, whichever has the minimum rating.		
1. When AC coupling the output should never be allowed to swing above the absolute voltage rating of the part, which is 6V.		
2. When $V_{CC}=3.3V$ , the OUT+ and OUT- should not be driven below 1.6V. In most 3.3V applications, this will make DC coupling impractical.		

When DC coupled, OUT+ should be connected through a series resistor to the laser such that the total impedance seen at the output is 25 ohms. This will result in the optimum pulse response while allowing the maximum modulation current (see [Figure 4-3](#)).

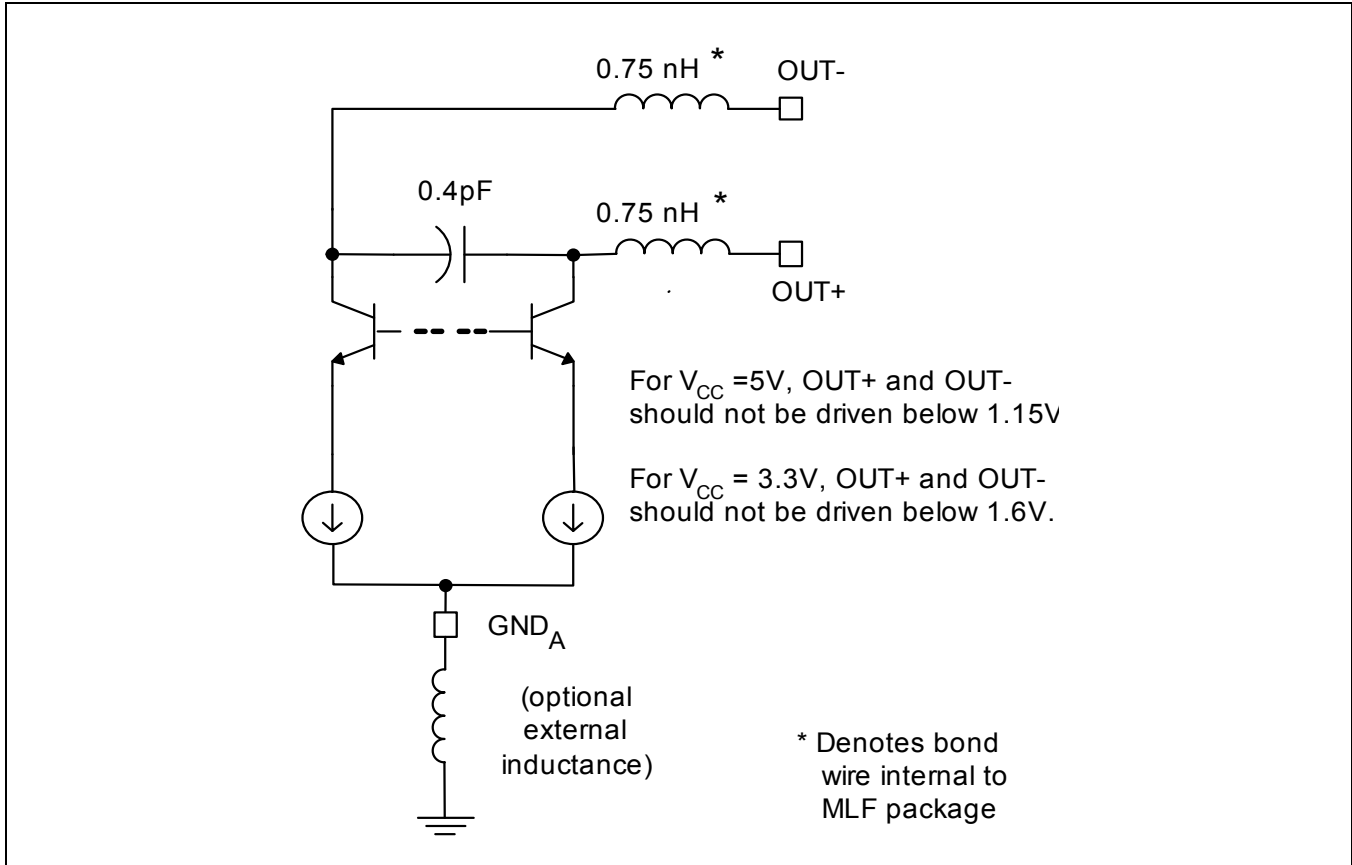
The output can also be AC coupled to the laser. This is the required operating mode when using a 3.3V supply (unless the laser has a small forward voltage and OUT+ will not go below 1.6V). When AC coupled the dynamic resistance seen by OUT+ should still be 25 ohms. In addition to a resistor in series with the laser, a capacitor is added in series and a ferrite is used to pull up the collector at OUT+ to  $V_{CC}$ .

When the laser is AC coupled, the OUT- pin is usually tied to the laser anode through an AC coupled series resistor which matches the impedance seen by the OUT+ pad (see [Figure 4-1](#)).

The output stage also has a separate current path to GND labelled  $GND_0$ . This isolates the output switching currents from the rest of the system.



Figure 3-6. Modulator Output



### 3.3.7 Fail Output

The M02061 has a FAIL alarm output which is compatible with the TX\_FAULT signalling requirements of common pluggable module standards.

The ESD protection on this pin provides a true open collector output that can withstand significant variation in  $V_{CC}$  when signalling between circuit boards. Also, if the M02061 loses power the pull-up will signal a fail condition. In a simple static protection scheme used by other ICs the protection diodes would clamp the FAIL signal to ground when the chip loses power.

### 3.3.8 TX Disable and Disable Delay Control

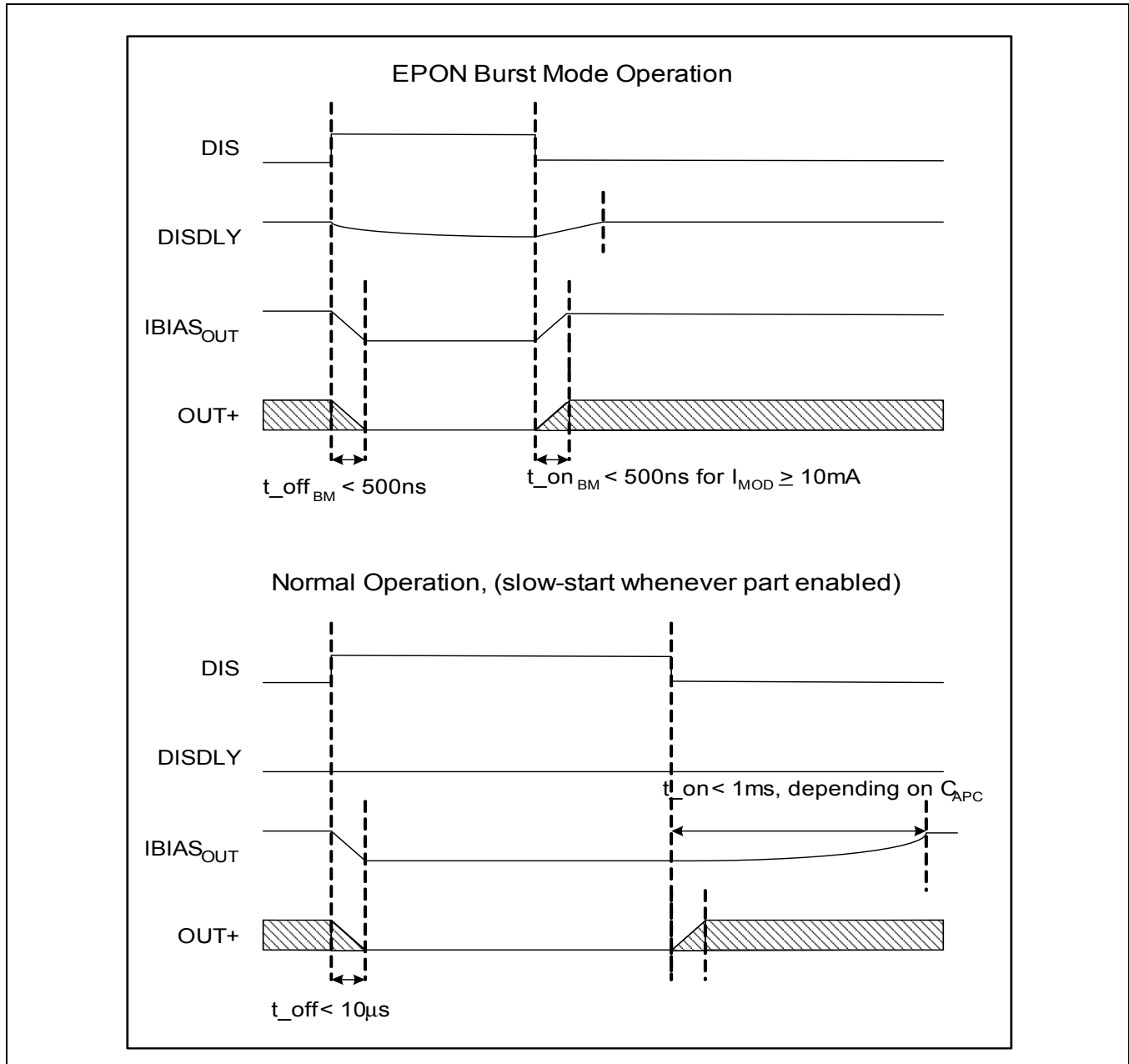
The DIS pin is used to disable the transmit signal (both the modulation and bias current are disabled when DIS = high).

The DIS input is compatible with TTL levels regardless of whether  $V_{CC} = 5V$  or  $V_{CC} = 3.3V$ . The external 4.7k $\Omega$  and 10k $\Omega$  pull-up resistor required by most interface standards is not needed because this pin has an internal 7k $\Omega$  resistor to  $V_{CC}$ .

The DISDLY pin is used in conjunction with the DIS pin to control bias current enable time. In normal operation the DISDLY pin should be connected to ground. In this case, each time DIS transitions from high to low the bias current will be enabled by the “slow-start” circuitry (enable time of less than 1 ms with a  $C_{APC} = 2.2$  nF).

For burst mode operation a capacitor C is added to the DISDLY pin, the slow-start circuitry is disabled for approximately  $T = 3 * 10^6$  (sec/F) \* C (F) following the DIS high transition (see figure 8). If the part is enabled (DIS transitions low) during this time the bias and modulation current will quickly return to within 90% of their final value (in less than 500ns). If DIS transitions low after the DISDLY time the slow-start circuitry will engage and the bias current will not return to its final value for approximately 1ms (depending on the  $C_{APC}$  capacitor value).

Figure 3-7. DIS and DISDLY Timing



### 3.3.9 TX Disable Control

The DIS pin is used to disable the transmit signal (both the modulation and bias current are disabled when DIS = high).

The DIS input is compatible with TTL levels regardless of whether  $V_{CC} = 5V$  or  $V_{CC} = 3.3V$ . The external  $4.7k\Omega$  and  $10k\Omega$  pull-up resistor required by most interface standards is not needed because this pin has an internal  $7k\Omega$  resistor to  $V_{CC}$ .

### 3.3.10 Monitor Outputs

To facilitate complying with laser safety and DDMI<sup>1</sup> requirements, output monitors are provided for transmit power ( $TxPwr_{MON}$ ) bias ( $BIAS_{MON}$ ) and modulation current ( $MOD_{MON}$ ).

These outputs will source current proportional to the emitted optical power ( $TxPwr_{MON}$ ) the bias current ( $BIAS_{MON}$ ) and modulation current ( $MOD_{MON}$ ). These pins should be terminated with a resistor to ground that sets the desired full-scale voltage (not to exceed  $V_{CC3}-1V$ ). Using a monitor polarity selection ( $MON_{POL}$ ) these monitors can be set to sink current instead of source current. They will then need to be terminated with a resistor to  $V_{CC3}$  and the induced voltage should not exceed 2.5V.

If the outputs of these monitors are not needed,  $MON_{POL}$ ,  $TxPwr_{MON}$ ,  $BIAS_{MON}$ , and  $MOD_{MON}$  can all be left floating and the chip current consumption will be reduced by the value of the monitor currents.

## 3.4 Laser Eye Safety

Using this laser driver in the manner described herein does not ensure that the resulting laser transmitter complies with established standards such as IEC 825. Users must take the necessary precautions to ensure that eye safety and other applicable standards are met. Note that determining and implementing the level of fault tolerance required by the applications that this part is going into is the responsibility of the transmitter designer and manufacturer since the application of this device cannot be controlled by MACOM.

### 3.4.1 Safety Circuitry

On the M02061-12 with DISDLY on pin 8, SCB is internally bonded to ground so SCB is always in a logic low state. When SCB is high the OUTP, BIASout and SVCC outputs will not be disabled when FAIL asserts (FAIL goes high). The outputs are only disabled by making DIS high.

The FAIL output will also ignore much of the safety sensing circuitry when SCB is high. However, it will monitor the state of the window comparators at pin  $APC_{SET}$ . The bias current is controlled to nominally maintain the voltage at  $APC_{SET}$  to 1.3V. The threshold levels at the window comparators around  $APC_{SET}$  are specified by the parameters  $V_{BH}$  and  $V_{BL}$  as shown in the table below. This provides the same level of eye safety protection as our previous generation of laser drivers. The current sourced out of pin  $APC_{SET}$  is equal to the current into pin  $I_{PIN}$  sourced from the laser monitor photo diode. If the laser is emitting excess power this will be reflected in the  $I_{PIN}$  current and the voltage at  $APC_{SET}$  will go high and the FAIL pin will assert. If  $I_{PIN}$  is not connected to the laser monitor photo diode then the voltage at  $APC_{SET}$  will fall and FAIL will assert.

When SCB is low, safety circuitry in the M02061 will disable the modulation and bias current and assert the FAIL output immediately upon detecting a fault condition. In addition, the supply voltage that sources the laser ( $SV_{CC}$  or

an external switch controlled by SHDWN<sub>OUT</sub>) will immediately go open circuit and prevent any current from passing through the laser.

Fault conditions checked by the M02061 include shorts to ground or V<sub>CC</sub> of all pins which can increase the laser modulation or bias current.

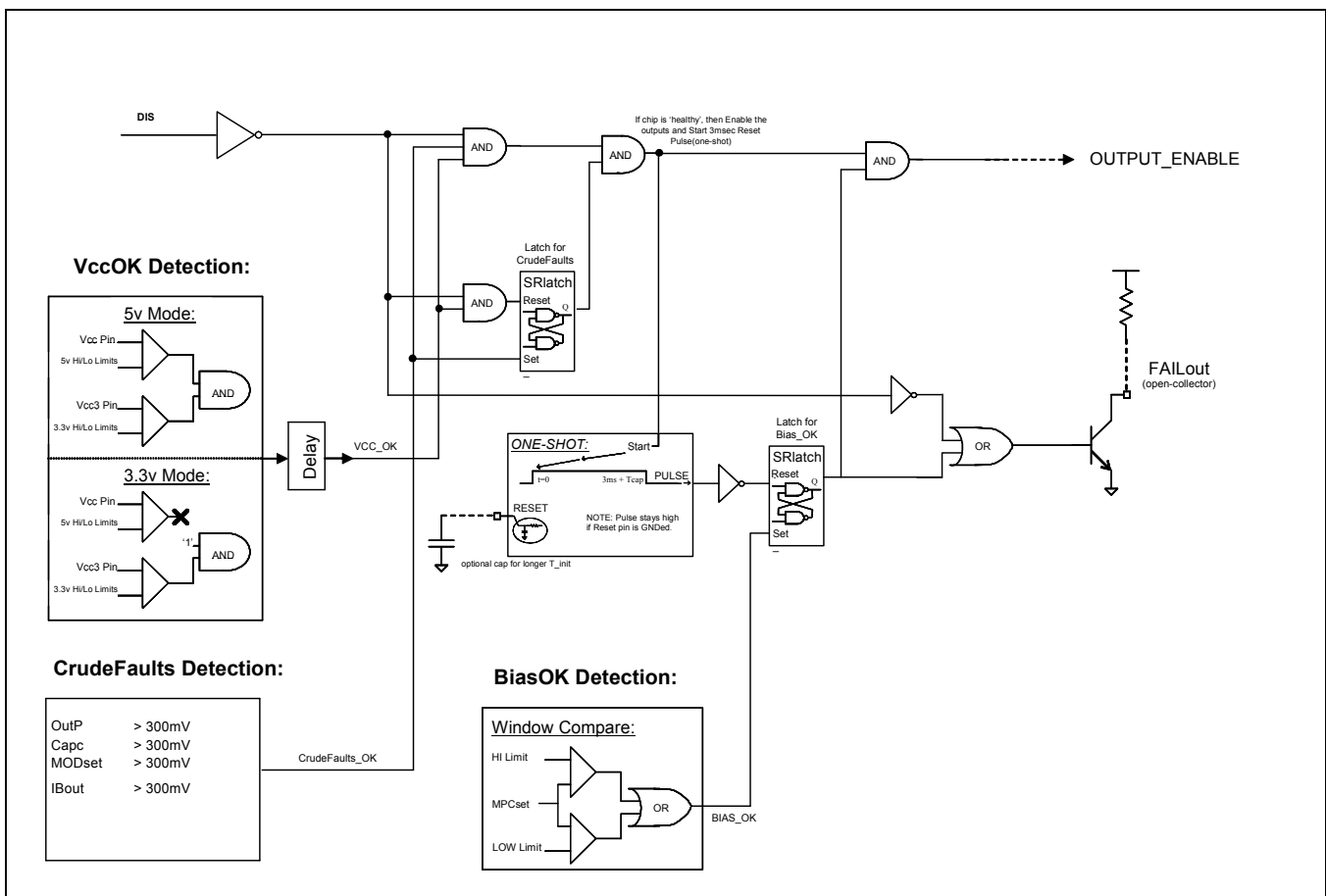
For an initialization sequence to be successful, all the fault detection monitors must signal that the chip is “healthy”. When DIS goes low, pins are checked for shorts to ground or V<sub>CC</sub> and a FAIL condition is latched if there is a fault.

If the state of the pins is OK, a one-shot at the reset pin begins a countdown which will latch a FAIL condition if the bias current has not stabilized to an acceptable level during the one-shot time. The one-shot can be extended with an external capacitor connected from the RESET pin to ground.

The one-shot<sup>1</sup> width is approximately:

$$T_{\text{ONE-SHOT}} = 3 \text{ ms} + (0.3 \text{ ms/pF}) \times (\text{external capacitance}).$$

**Figure 3-8. Safety Circuit Block Diagram, for SCB Pin Low**



1. The one-shot is actually comprised of an oscillator and 10-bit counter.

### 3.5 Fault Conditions when SCB is Low

This section describes the M02061 operating modes during fault conditions. Over voltage, under voltage, pins shorted to  $V_{CC}$  and pins shorted to ground are included in the fault table.

**Table 3-3. Circuit Response to Single-point Fault Conditions, when SCB is Low<sup>1 2</sup> (1 of 2)**

Pin Name	Circuit Response to Over-voltage Condition or Short to $V_{CC}$	Circuit Response to Under-Voltage Condition or Short to Ground
$V_{CC}$	Bias and modulation outputs are disabled once $V_{CC}$ rises above the supply detection (high voltage) threshold	Bias and modulation outputs are disabled once $V_{CC}$ drops below the supply detection (low voltage) threshold
DIN+, DIN-	The APC loop will attempt to compensate for the change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1, 2)</sup>	The APC loop will attempt to compensate for the change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1, 2)</sup>
$V_{CC3_{SEL}}$	Does not affect laser power.	Does not affect laser power.
DIS	Bias and modulation outputs are disabled. 3.3V operation - $SV_{CC}$ is opened. 5V operation - $SHDWN_{OUT}$ goes high.	Does not affect laser power (normal condition for circuit operation).
FAIL	Does not affect laser power.	Does not affect laser power.
RESET	Does not affect laser power.	Does not affect laser power.
$MOD_{MON}$	Does not affect laser power.	Does not affect laser power.
$BIAS_{MON}$	Does not affect laser power.	Does not affect laser power.
$TxPWR_{MON}$	Does not affect laser power.	Does not affect laser power.
$APC_{SET}$	A fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
$I_{PIN}$	A fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
$IBIAS_{OUT}$	The laser will be turned off, then a fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
$OUT_P$	Laser modulation is prevented; the APC loop will increase the bias current to compensate for the drop in laser power if it is DC coupled. If the set output power can not be obtained, a fault state occurs. <sup>(1, 2)</sup>	A fault state occurs. <sup>(1)</sup>
$OUT_N$	Does not affect laser power.	Does not affect laser power.
$SV_{CC}$	Does not affect laser power.	Laser bias current will be shut off and a fault state occurs. <sup>(1)</sup>
$C_{APC}$	Laser bias current will be shut off, then a fault state occurs. <sup>(1)</sup>	A fault state occurs. <sup>(1)</sup>
$V_{CC3}$	Bias and modulation outputs are disabled once $V_{CC3}$ rises above the supply detection (high voltage) threshold	Bias and modulation outputs are disabled once $V_{CC3}$ drops below the supply detection (low voltage) threshold
PWA	Does not affect laser power.	Does not affect laser power
$SHDWN_{OUT}$	Does not affect laser power. if this pin is used to control an external switch, laser current is disabled and fault state occurs. <sup>(1)</sup>	Does not affect laser power.
$MOD_{SET}$	The APC loop will attempt to compensate for the change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1, 2)</sup>	A fault state occurs. <sup>(1)</sup>

**Table 3-3. Circuit Response to Single-point Fault Conditions, when SCB is Low<sup>1 2</sup> (2 of 2)**

Pin Name	Circuit Response to Over-voltage Condition or Short to V <sub>CC</sub>	Circuit Response to Under-Voltage Condition or Short to Ground
TC <sub>SLOPE</sub>	Does not affect laser power.	May affect laser power. If this is the case, the APC loop will attempt to compensate for the change in output power. If the APC loop can not maintain the set average power, a fault state occurs. <sup>(1, 2)</sup>
SCB	Does not affect laser power.	Does not affect laser power.
DISDLY	Does not affect laser power.	Does not affect laser power.

**NOTES:**

1. A fault state will assert the FAIL output, disable bias and modulation outputs and will either open the switch at SV<sub>CC</sub> (3.3V operation) or SHDWN<sub>OUT</sub> will go high (5V operation).
2. Does not affect laser power when the output is AC coupled to the laser.

# 4.0 Applications Information

## 4.1 General

- SFP and SFF Modules
- 1G/2G/4G Fibre Channel modules
- Short reach and Metro SONET/SDH

Figure 4-1 and Figure 4-3 illustrate typical applications for 3.3V/AC coupled and 5V/DC coupled laser.

**Figure 4-1. Application Diagram, VCC = 3.3V Laser AC Coupled Example**

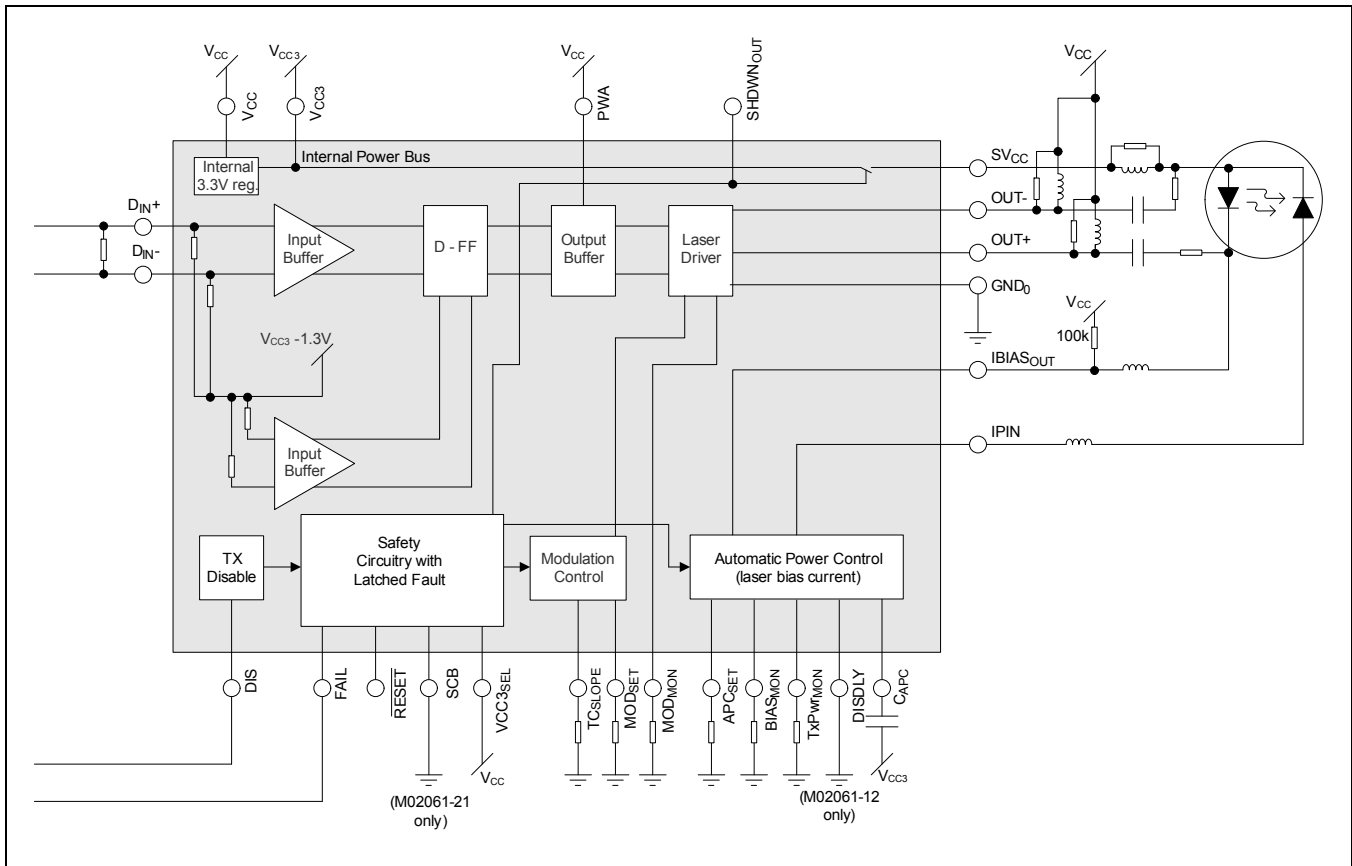
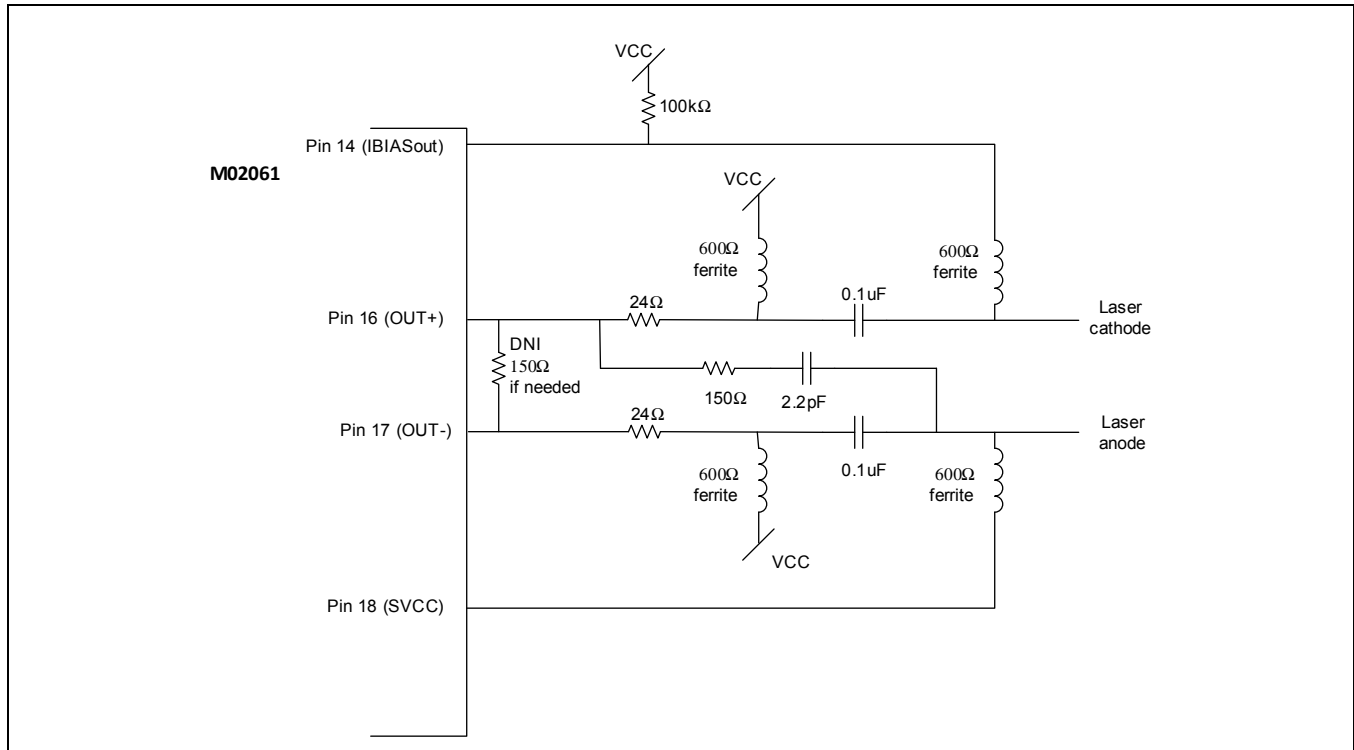
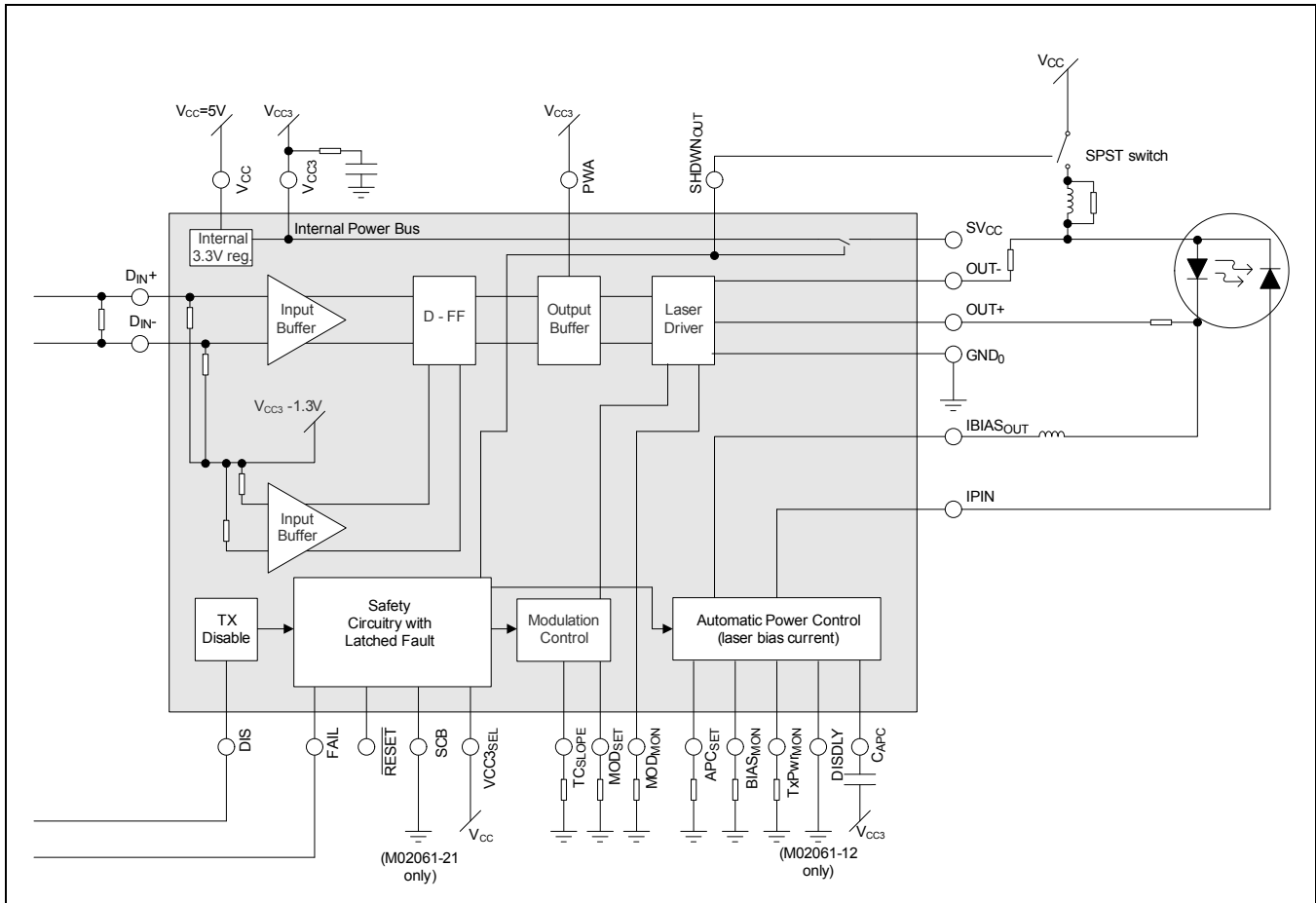




Figure 4-2. Application Diagram, VCC = 3.3V Laser AC Coupled with Recommended Component Values



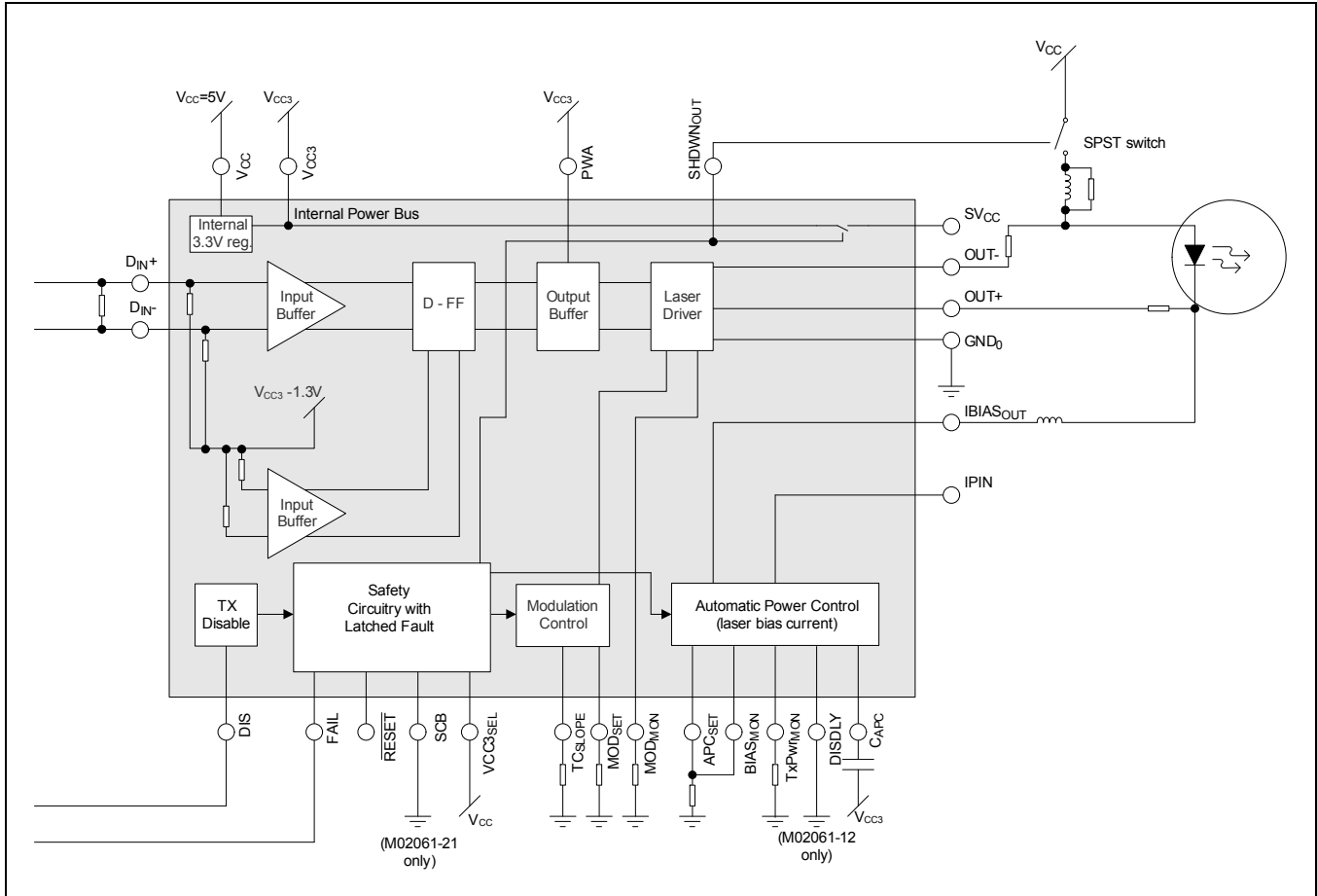
**Figure 4-3. Application Diagram, VCC = 5V Laser DC Coupled Example**



## 4.2 Video Operation

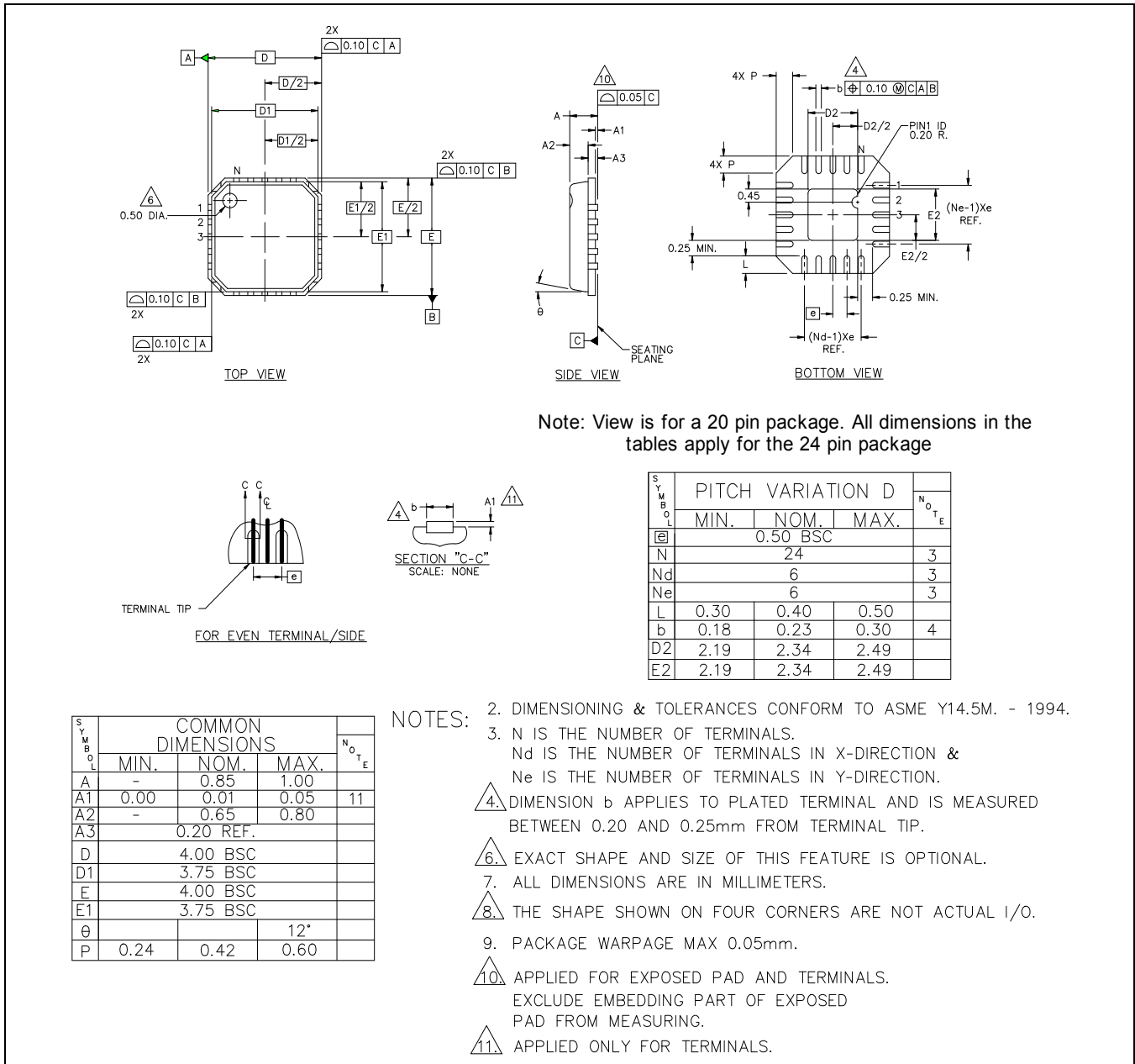
The M02061 can be used to transmit digital video optical data even in the presence of the pathological signal. This is done by fully DC coupling the signal from the input to the laser output. In most data communications applications, AC coupling occurs at 3 points in a laser driver schematic: the data inputs, the APC control, and coupling the modulation current to the laser. In the M02061 DC coupling can be used at all 3 of these points. The data inputs can be DC coupled using PECL or CML levels (see Section 3.3.3, “Data Inputs”). LVDS signals can be DC coupled with level shifting. The APC of the bias current is controlled by feedback from the monitor photo diode in the laser package in most communications applications. In video applications this monitor photo diode should not be used if the pathological pattern may occur. Instead, the APC should be controlled in an “open loop” configuration. (Open loop simply means a monitor photo diode is not used). In the open loop configuration the APC is controlled by a resistor or a thermistor network or a look-up table. This removes AC time constants from the bias current. In Figure 4-4 the BIASmon pin is connected to the APCset pin. In this case the bias current is  $I_{BIAS} = 100 \times (1.3V / R_{APCset})$ . The modulation current output OUT+ can be DC coupled to the laser as shown in Figure 4-4. There are no AC time constants in the modulation current amplitude in this configuration.

Figure 4-4. Video Application Block Diagram



# 5.0 Package Specification

Figure 5-1. QFN24 Package Information



M/A-COM Technology Solutions Inc. All rights reserved.

Information in this document is provided in connection with M/A-COM Technology Solutions Inc ("MACOM") products. These materials are provided by MACOM as a service to its customers and may be used for informational purposes only. Except as provided in MACOM's Terms and Conditions of Sale for such products or in any separate agreement related to this document, MACOM assumes no liability whatsoever. MACOM assumes no responsibility for errors or omissions in these materials. MACOM may make changes to specifications and product descriptions at any time, without notice. MACOM makes no commitment to update the information and shall have no responsibility whatsoever for conflicts or incompatibilities arising from future changes to its specifications and product descriptions. No license, express or implied, by estoppel or otherwise, to any intellectual property rights is granted by this document.

THESE MATERIALS ARE PROVIDED "AS IS" WITHOUT WARRANTY OF ANY KIND, EITHER EXPRESS OR IMPLIED, RELATING TO SALE AND/OR USE OF MACOM PRODUCTS INCLUDING LIABILITY OR WARRANTIES RELATING TO FITNESS FOR A PARTICULAR PURPOSE, CONSEQUENTIAL OR INCIDENTAL DAMAGES, MERCHANTABILITY, OR INFRINGEMENT OF ANY PATENT, COPYRIGHT OR OTHER INTELLECTUAL PROPERTY RIGHT. MACOM FURTHER DOES NOT WARRANT THE ACCURACY OR COMPLETENESS OF THE INFORMATION, TEXT, GRAPHICS OR OTHER ITEMS CONTAINED WITHIN THESE MATERIALS. MACOM SHALL NOT BE LIABLE FOR ANY SPECIAL, INDIRECT, INCIDENTAL, OR CONSEQUENTIAL DAMAGES, INCLUDING WITHOUT LIMITATION, LOST REVENUES OR LOST PROFITS, WHICH MAY RESULT FROM THE USE OF THESE MATERIALS.

MACOM products are not intended for use in medical, lifesaving or life sustaining applications. MACOM customers using or selling MACOM products for use in such applications do so at their own risk and agree to fully indemnify MACOM for any damages resulting from such improper use or sale.